

Stable isomorphisms of f.d. Hilb spaces.

By a stable isom between K and L we mean an isom $K \oplus W \xrightarrow{\sim} W \oplus L$ for some

~~W~~ I think we can ^{then} define a scattering operator $g(z) : K \rightarrow L$ ~~that~~ rational function z

analytic for $|z| \leq 1$ and unitary valued for $|z| = 1$. But this does not determine W ~~unless~~ when there are bound states. We need to understand the algebra. There are lots of things to prove such as the existence of the scattering. Recall how done.

$$\begin{array}{ccccccc}
 \oplus U^{-2}K & \oplus U^{-1}K & \oplus W & \oplus L & \oplus UL & \oplus U^2L \\
 \downarrow u & \downarrow u & \downarrow u & \downarrow u & \downarrow u & \downarrow u \\
 \oplus U^{-2}K & \oplus UK & \oplus K & \oplus W & \oplus UL & \oplus U^2L
 \end{array}$$

obtain

So we ~~define~~ Hilbert space rep of $C(S^1)$.

Now can look for eigenvectors in the Gelfand sense.

Ultimately you get out of this a coherent analytic sheaf over S^2 . This you must be precise about, because you need a precise K -module. Concentrate on the sheaf F .

The rank will be $\dim(K) = \dim(L)$; call this r .

~~Recall that~~ V should be $K \oplus W = W \oplus L$. Recall

that if $0 \rightarrow \mathcal{O}(-1) \otimes W \rightarrow \mathcal{O} \otimes V \rightarrow F \rightarrow 0$, then $V = H^0(F)$
 has $\dim = \deg(F) + \text{rank}(F)$. $\therefore \dim(W) = \deg(F)$.

You see the coherent sheaf as follows. You see ~~the~~ its Cech complex, namely, the above Hilb. space gives the sections over S^1 . ~~Recall that~~ Then inside this are sections extending to either disk. These will be, ~~incoming + outgoing~~ subspaces, and they intersect to give $V = H^0(F)$.

500 What can you hope for? You might be able to put metrics on Joyce's modules.

For the moment you need to be able to calculate. First show that bound states split off.

What happens if $W=0$. Then we give an isomorphism of $K \xrightarrow{\sim} L$. Sheaf is \mathcal{O}^r $r = \dim(K) = \dim(L)$
want $\mathbb{Z}/3$ to act. Need a stable isom. $\mathbb{C}^3 \oplus W \xrightarrow{\sim} W \oplus \mathbb{C}^3$
Will have $S(z) \in M_3(\mathbb{C}) \quad \exists \quad S(z) \in U(3) \quad \text{for } |z| \in S^1$

Jan 21. ~~What's going on~~

try to understand whether you can get anything interesting on the $\Gamma = SL_2(\mathbb{Z})$ tree, or whether any ~~invariant~~ Γ invariant circuit over the tree necessarily has ~~the~~ continuous spectrum. So let's ~~not~~ look at ~~the question~~ how to describe ~~something~~ ^{equivariantly} over the tree. ~~So consider what begin by looking at~~

begin where? finite LC circuit 2-port over the edges
finite LC circuit 3-port for the vertices. Former must have $\mathbb{Z}/2$ symmetry, latter $\mathbb{Z}/3$ symmetry. First look at the real case. (LC). So where do I start? The ~~method~~ will be just as before to use the impedance function on the edges, except now you have to understand ~~how to~~ find how ^{to} handle the vertices. So let us see what to do. identical

to get started let us connect n 3-ports together to see what happens



first look at 2 ~~ports~~ ports.
First you need to handle the group \mathbb{Z} and maybe next $\mathbb{Z}/2 \times \mathbb{Z}/2$ before proceeding to the general case. ~~Next~~
~~What~~ You have description of a general 2-port, namely a ^{positive and} quadratic form on \mathbb{R}^2 depending on s in a specific rational way. ~~With luck it will be clear~~

501 So you have this tree acted on by $PSL_2(\mathbb{Z})$.
 $= \mathbb{Z}/2 * \mathbb{Z}/3$, and you want ~~to construct~~ some
idea of what ~~we~~ can construct by surgery. Look
at the simpler case of $\mathbb{Z}/2 * \mathbb{Z}/2$, or even \mathbb{Z}
which might be simpler.

Digress to make notes on Top. paper refereed
by D. Salamon. Novikov Morse theory: manifold
 X with a "Morse function" $X \rightarrow S^1$. In this
situation Novikov, ~~in~~ analogy ~~with~~ Witten,
constructed a chain complex out of critical
point data, ~~but~~ but the complex consists of f.g.
free $\mathbb{Z}[t, t^{-1}]$ -modules. Presumably the complex is
supposed to calculate the homology of $\tilde{X} = X \times_{S^1} \mathbb{R}$.
Choose on X a generic metric, let Σ be the
inverse image of a generic point of S^1 say 0,
and ~~look~~ look at the gradient flow arising
from the Morse function. Watch the trajectory
of a point of Σ . If ~~one doesn't hit a~~
the trajectory doesn't get swallowed by a critical
point, then it comes back to Σ . So you
get a ~~map~~ partially defined $f: \Sigma \rightarrow \Sigma$, such
that the fixpts of f^k for any k yield closed trajectories
of the gradient flow. ~~These fixpts~~ Apparently
there's a zeta function arising from these fixpts;
D. Fried (BU) studied this for a ~~flow~~ nonsingular
and transversal Σ . In the present paper a
connection is made with K-forsim. One assumes

502 The chain complex of finite free $\mathbb{Z}[t, t^{-1}]$ -modules is acyclic, so then the torsion is defined and should be something like an element of $\mathbb{Z}[t, t^{-1}]$.

The ~~three~~ main theorem I think says that the two ~~two~~ ^{natural} expressions for ζ differ multiplicatively by the torsion. Ideas in the proof include the Lefschetz Fixed theorem, extending the partial operator f on homology somehow.

Return to tree acted on by $PSL_2(\mathbb{Z}) = \mathbb{Z}/2 * \mathbb{Z}/3$.

Actually, take simpler tree, namely \mathbb{R} acted on by \mathbb{Z} .

~~Each edge is a 2-port~~ Each edge is a 2-port and there is ~~coupling~~ coupling at each vertex. Now

I need to describe LC 2-ports, a general LC 2-port. Real ^{polarized} vector space $V = V^+ \oplus V^-$ equipped with a

~~quadratic form~~ quadratic form which is a rational function of s simple poles ^{at $s = \pm i\omega$} non-negative residues etc.

V^+, V^- same dimensions. Coupling should be an isomorphism $V^+ \simeq V^-$. Idea: can you replace

the quadratic form on $V = V^+ \oplus V^-$ by some

sort of map $V^+ \rightarrow V^-$ which can be combined

with the ^{coupling} isom $V^+ \xrightarrow{\sim} V^-$ to be iterated ~~by the~~

~~like a~~ ^{like a} transfer matrix. ~~Can you replace~~

simple puzzle: Quadratic form q_s on $V^+ \oplus V^-$ together with isomorphism $V^+ \simeq V^-$ yields operator $V^+ \xrightarrow{A} V^+$ depending on s , which can be iterated

k times then trace taken. These traces can be

assembled into
$$-\log \det(1 - \lambda A) = \sum_{k=1}^{\infty} \frac{\lambda^k}{k} \text{tr}(A^k)$$

503 So what next?

Let's try to understand the coupling.

$g_s: V \rightarrow V^*$ symmetric positive.

$\Gamma_{g_s} \subset V \oplus V^*$ Lagrangian subspace

significance of splitting

$V = V^+ \oplus V^-$
 $V^* = V^{+*} \oplus V^{-*}$

have $\begin{cases} \theta: V^+ \rightarrow V^- \\ \theta^t: V^{-*} \rightarrow V^{+*} \end{cases}$

With appropriate sign $\Gamma_\theta + \Gamma_{\theta^t}$ might be Lagrang.

This is interesting. How do I explain coupling?

Take θ and pass to a quotient. You have the quadratic form on $V = V^+ \oplus V^-$ and coupling amounts to restriction to the graph of $\theta: V^+ \rightarrow V^-$.

I'm thinking of V^+ as the potentials, so ~~you~~ you are setting voltages equal (up to θ). You still don't have a map $V^+ \rightarrow V^+$.

Would it be clearer if you ~~could~~ represented V as a subquotient of a polarized Hilbert space.

Consider then a quadratic form on $\mathbb{R}^2 = V^+ \oplus V^-$

$g_s = \sum_{\omega \in S} a_\omega \frac{s(1+\omega^2)}{s^2+\omega^2}$ $a_\omega \geq 0$ $\sum a_\omega > 0$

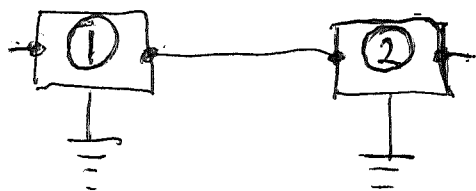
A quadratic form has 3 coeff $\begin{pmatrix} a & b \\ b & c \end{pmatrix}$ $\begin{cases} a > 0 \\ ac - b^2 > 0 \end{cases}$

~~Restrict to~~ Suppose given $\theta: V^+ \rightarrow V^-$. I propose to restrict g_s to the graph of θ . ~~is that~~ Maybe all that's important is that we have a line in \mathbb{R}^2 to restrict to. No - you want to ~~be~~ ~~able to~~ get an operator on V^+ .

504

Problem: Given q_s on $\mathbb{R}^2 = \mathbb{R}^+ \oplus \mathbb{R}^-$

You haven't gotten anywhere. Try something simpler. Given V_1, V_2, V_3 and q_s pos. def. q forms on $V_1 \oplus V_2$ and $V_2 \oplus V_3$ explain how to get a ~~quadratic~~ quadratic form on $V_1 \oplus V_3$. To avoid disconnected graphs, assume always connected with basept. ~~Then~~ Then we have 2 circuits.



I am thinking of series connections

① gives $(E_1, E_2) \in V_1 \oplus V_2 \xrightarrow{q_s} (I_1, I_2) \in V_1^* \oplus V_2^*$

$$(E, q_s(E)) = \text{power } E_1 I_1 + E_2 I_2$$

So what does coupling mean. We consider

(E_1, E_2, I_1, I_2) graph of $q_s^{(1)}$ (E_3, E_4, I_3, I_4) graph of $q_s^{(2)}$

4 ind. variables

~~Must~~ impose conditions $E_2 = E_3, I_2 = -I_3$
imposing $E_2 = E_3$ is restricting to a subspace of the voltage space, but then you ignore $E_2 = E_3$ which means passing to a quotient space.

So you have a subquotient of the direct sum.

Can you get this any better?? First of all concentrate on the geometry. You have two ~~circuits~~ networks and a common subquotient. ~~Then~~

505 Puzzle What is the significance of a real
 v.s. V equipped with quad form $g_s(\sigma)$ rational in s etc.
 You would like to say that there is some Hilbert
 space around with a partial ~~quad~~ self adjoint? op.
~~You need something again.~~ You are missing this
 picture. Try coupling ^{positive} quadratic forms on \mathbb{R}^2 .

First point. Given a ^{pos. def.} quad. form on $V_1 \oplus V_2$
 and a pos. def quad form on $V_2 \oplus V_3$ there is
 an induced quadratic form on $V_1 \oplus V_3$, namely

$$\begin{array}{ccc}
 V_1 \oplus V_2 \oplus V_3 & \xrightarrow{1 \times \Delta} & (V_1 \oplus V_2) \oplus (V_2 \oplus V_3) \\
 \downarrow & & \begin{array}{cc} P & Q \end{array} \\
 V_1 \oplus V_3 & &
 \end{array}$$

$$\cancel{r} r(x, z) = \inf_{y \in V_2} (p(x, y) + q(y, z))$$

$$p(x, y) = \begin{pmatrix} x \\ y \end{pmatrix}^t \begin{pmatrix} a & b \\ b & c \end{pmatrix} \begin{pmatrix} x \\ y \end{pmatrix} \quad q(y, z) = \begin{pmatrix} y \\ z \end{pmatrix}^t \begin{pmatrix} d & e \\ e & f \end{pmatrix} \begin{pmatrix} y \\ z \end{pmatrix}$$

$$ax^2 + 2bxy + cy^2 + dy^2 + 2eyz + fz^2$$

$$\underbrace{(c+d)y^2 + (2bx + 2ez)y}_{\text{complete the square}}$$

$$(c+d) \left[y^2 + 2 \frac{(bx+ez)}{c+d} y + \left(\frac{bx+ez}{c+d} \right)^2 \right] - \frac{(bx+ez)^2}{c+d}$$

$$ax^2 - \frac{b^2x^2 + 2bexz + e^2z^2}{c+d} + fz^2$$

$$\left(a - \frac{b^2}{c+d} \right) x^2 - 2 \frac{be}{c+d} xz + \left(f - \frac{e^2}{c+d} \right) z^2$$

506 Is there some way to compose things, like replacing quadratic forms by scattering matrices.

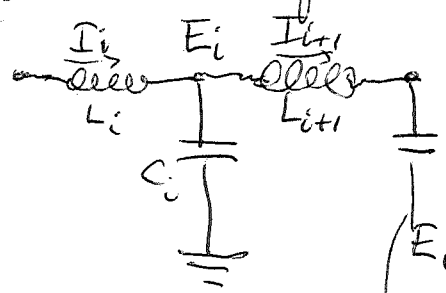
$$\begin{pmatrix} a & b \\ b & d \end{pmatrix}, \begin{pmatrix} d & e \\ e & f \end{pmatrix} \mapsto \begin{pmatrix} a & b & 0 \\ b & c+d & e \\ 0 & e & f \end{pmatrix}$$

restrict to $y = -\frac{bx+ez}{c+d}$

$$\begin{pmatrix} 1 & -\frac{b}{c+d} & 0 \\ 0 & -\frac{e}{c+d} & 1 \end{pmatrix} \begin{pmatrix} a & b \\ b & c+d & e \\ e & f \end{pmatrix} \begin{pmatrix} 1 & 0 \\ -\frac{b}{c+d} & -\frac{e}{c+d} \\ 0 & 1 \end{pmatrix}$$

$$\begin{pmatrix} 1 & -\frac{b}{c+d} & 0 \\ 0 & -\frac{e}{c+d} & 1 \end{pmatrix} \begin{pmatrix} a - \frac{b^2}{c+d} & -\frac{be}{c+d} \\ b - b & -e \\ -\frac{be}{c+d} & -\frac{e^2}{c+d} + f \end{pmatrix} = \begin{pmatrix} a - \frac{b^2}{c+d} & -\frac{be}{c+d} \\ -\frac{be}{c+d} & f - \frac{e^2}{c+d} \end{pmatrix}$$

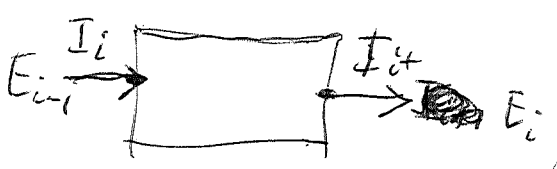
There's a clear analogy with Jacobi matrices. You want to find something that composes nicely.



$$\begin{aligned} C_{iS} E_i &= I_i - I_{i+1} & I_{i+1} &= I_i + C_{iS} E_i \\ L_{iS} I_i &= E_{i-1} - E_i & E_{i-1} &= E_i + L_{iS} I_i \end{aligned}$$

$$\begin{pmatrix} E_{i-1} \\ I_i \end{pmatrix} = \begin{pmatrix} 1 & L_{iS} \\ 0 & 1 \end{pmatrix} \begin{pmatrix} E_i \\ I_i \end{pmatrix} \quad \begin{pmatrix} E_i \\ I_i \end{pmatrix} = \begin{pmatrix} 1 & 0 \\ C_{iS} & 1 \end{pmatrix} \begin{pmatrix} E_{i+1} \\ I_{i+1} \end{pmatrix}$$

$$\begin{pmatrix} E_{i-1} \\ I_i \end{pmatrix} = \begin{pmatrix} 1 & L_{iS} \\ 0 & 1 \end{pmatrix} \begin{pmatrix} 1 & 0 \\ C_{iS} & 1 \end{pmatrix} \begin{pmatrix} E_i \\ I_{i+1} \end{pmatrix}$$



so maybe the point is not to look at $E_1, E_2 \mapsto I_1, I_2$ but rather the IVP $E_1/I_1 \rightsquigarrow E_2/I_2$

508. ~~Symplectic space~~ Take a symplectic isom.

$$g = \begin{pmatrix} a & b \\ c & d \end{pmatrix} : \begin{matrix} V \\ \oplus \\ V^* \end{matrix} \xrightarrow{\sim} \begin{matrix} W \\ \oplus \\ W^* \end{matrix} \quad \text{Look at its graph } \subset \begin{matrix} V \oplus W \\ \oplus \\ (V \oplus W)^* \end{matrix} = \begin{matrix} X \\ \oplus \\ X^* \end{matrix}$$

~~What~~ What does symplectic mean.

$$g^* \begin{pmatrix} 0 & 1 \\ -1 & 0 \end{pmatrix} g = \begin{pmatrix} 0 & 1 \\ -1 & 0 \end{pmatrix}$$

$$\begin{pmatrix} a & b \\ c & d \end{pmatrix}^{-1} = \begin{pmatrix} 0 & -1 \\ 1 & 0 \end{pmatrix} \begin{pmatrix} a^t & c^t \\ b^t & d^t \end{pmatrix} \begin{pmatrix} 0 & 1 \\ -1 & 0 \end{pmatrix} = \begin{pmatrix} d^t & -b^t \\ -c^t & a^t \end{pmatrix}$$

Look at Γ_g

$$\begin{matrix} W \\ \oplus \\ W^* \end{matrix} \begin{pmatrix} a & b \\ c & d \end{pmatrix} \begin{matrix} V \\ \oplus \\ V^* \end{matrix}$$

$$\begin{pmatrix} 1 & 0 \\ a & b \\ 0 & 1 \\ c & d \end{pmatrix} \begin{pmatrix} V \\ \oplus \\ V^* \end{pmatrix}$$

$$\subset \begin{pmatrix} V \\ \oplus \\ W \\ \oplus \\ V^* \\ \oplus \\ W^* \end{pmatrix}$$

$$\begin{pmatrix} 1 & a & 0 & c \\ 0 & b & 1 & d \end{pmatrix} \begin{pmatrix} 1 & & & \\ & 1 & & \\ & & -1 & \\ & & & -1 \end{pmatrix} \begin{pmatrix} 1 & 0 \\ a & b \\ 0 & 1 \\ c & d \end{pmatrix} = \begin{pmatrix} 1 & a & 0 & c \\ 0 & b & 1 & d \\ & & & \\ & & & \end{pmatrix} \begin{pmatrix} 0 & 1 \\ -c & -d \\ -1 & 0 \\ +a & +b \end{pmatrix}$$

$$= \begin{pmatrix} -ac+ca & 1-ad+bc \\ -bc-1+da & -b+db \end{pmatrix} = 0$$

Look at $\begin{pmatrix} 0 & 1 \\ c & d \end{pmatrix} \begin{pmatrix} 1 & 0 \\ a & b \end{pmatrix}^{-1} = \begin{pmatrix} 0 & 1 \\ c & d \end{pmatrix} \begin{pmatrix} 1 & 0 \\ -b^t a & b^t \end{pmatrix} = \begin{pmatrix} -b^t a & b^t \\ c-db^t a & db^t \end{pmatrix}$

$$\begin{pmatrix} 1 & a & 0 & c \\ 0 & b & 1 & d \end{pmatrix} \begin{pmatrix} 1 & & & \\ & 1 & & \\ & & -1 & \\ & & & -1 \end{pmatrix} \begin{pmatrix} 1 & 0 \\ a & b \\ 0 & 1 \\ c & d \end{pmatrix} = \begin{pmatrix} 1 & a & 0 & c \\ 0 & b & 1 & d \\ & & & \\ & & & \end{pmatrix} \begin{pmatrix} 0 & -1 \\ c & d \\ -1 & 0 \\ -a & -b \end{pmatrix} = 0$$

$$\begin{pmatrix} 0 & -1 \\ c & d \end{pmatrix} \begin{pmatrix} 1 & 0 \\ a & b \end{pmatrix}^{-1} = \begin{pmatrix} 0 & -1 \\ c & d \end{pmatrix} \begin{pmatrix} 1 & 0 \\ -b^t a & b^t \end{pmatrix} = \begin{pmatrix} b^t a & -b^t \\ c-db^t a & db^t \\ -b^t & \end{pmatrix}$$

507 So I ask about the transfer matrix instead.

So given $\begin{pmatrix} I_1 \\ -I_2 \end{pmatrix} = \begin{pmatrix} a & b \\ b & c \end{pmatrix} \begin{pmatrix} E_1 \\ E_2 \end{pmatrix}$ you solve for E_1, I_1 in terms of E_2, I_2

$$\begin{aligned} I_1 &= aE_1 + bE_2 & I_1 &= ab^{-1}(-I_2 - cE_2) + bE_2 \\ -I_2 &= bE_1 + cE_2 & E_1 &= b^{-1}(-I_2 - cE_2) \end{aligned}$$

$$\begin{aligned} I_1 &= -\frac{a}{b}I_2 + \left(b - \frac{ac}{b}\right)E_2 \\ E_1 &= -\frac{1}{b}I_2 - \frac{c}{b}E_2 \end{aligned} \quad \begin{pmatrix} -\frac{a}{b} & b - \frac{ac}{b} \\ -\frac{1}{b} & -\frac{c}{b} \end{pmatrix}$$

$$\det = +\frac{ac}{b^2} + \left(b - \frac{ac}{b}\right)\frac{1}{b} = 1.$$

as long as ~~the trans. coeff~~ the trans. coeff

Basically, what happening is that you have

idea here: a quadratic form on $V_1 \oplus V_2$ is ~~the graph~~ a kind of Lagrangian subspace of $(V_1 \oplus V_2) \oplus (V_1 \oplus V_2)^*$, and and generically should be the graph of a symplectic transf. $V_1 \oplus V_1^* \rightarrow V_2 \oplus V_2^*$

Jan 22
Complex version: ~~identify a part of~~ identify a part of unitary with ^{max} isotropic subspace of $H^+ \oplus H^-$ wrt $\|h_+\|^2 - \|h_-\|^2$. Now suppose you have $H^\pm = H_1^\pm \oplus H_2^\pm$. So the indef H-space H is the direct sum of H_1 and H_2 . ~~generically~~ The graph of a isometric isom. $T: H_1 \rightarrow H_2$ should be isotropic, except you should reverse signs on H_2

$$(H, \varepsilon) = (H_1, \varepsilon_1) \oplus (H_2, -\varepsilon_2) \quad \text{consider } \begin{pmatrix} 1 \\ T \end{pmatrix}: H_1 \rightarrow \begin{pmatrix} H_1 \\ H_2 \end{pmatrix}$$

$$\begin{pmatrix} h_1 \\ Th_1 \end{pmatrix}^* \begin{pmatrix} \varepsilon_1 & 0 \\ 0 & -\varepsilon_2 \end{pmatrix} \begin{pmatrix} h_1 \\ Th_1 \end{pmatrix} = h_1^* \varepsilon_1 h_1 - (Th_1)^* \varepsilon_2 (Th_1)$$

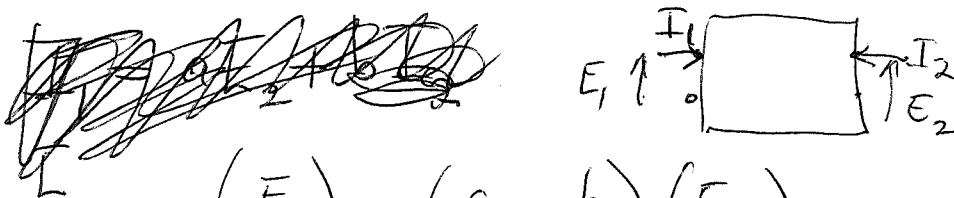
509 ~~How~~ To analyze 2-ports. First statement should concern the relation between $SL_2(\mathbb{R}) = Sp_2(\mathbb{R})$ and the space of pos. def quad forms on \mathbb{R}^2 . - The same as harmonic oscillators with phase space \mathbb{R}^2 . Both are 3 diml over \mathbb{R} and connected. $SL_2(\mathbb{R}) \sim SO(1)$ whereas pos def q f are contractible. Try ~~with~~ yesterday's notation:

$$\begin{pmatrix} E_1 \\ I_1 \end{pmatrix} = \begin{pmatrix} a & b \\ c & d \end{pmatrix} \begin{pmatrix} E_2 \\ I_2 \end{pmatrix}$$

~~Equation~~

$$I_1 = cE_2 + d(-cE_1 + aI_1)$$

$$(1 - da)I_1 = -dcE_1 + cE_2 \quad I_1 = \frac{+dE_1 + E_2}{b}$$



$$\begin{pmatrix} E_1 \\ I_1 \end{pmatrix} = \begin{pmatrix} a & b \\ c & d \end{pmatrix} \begin{pmatrix} E_2 \\ -I_2 \end{pmatrix}$$

$$ad - bc = 1.$$

$$E_1 = aE_2 - bI_2$$

$$-bI_2 = E_1 - aE_2$$

$$I_1 = cE_2 - dI_2$$

$$I_2 = \left(-\frac{1}{b}\right)E_1 + \frac{a}{b}E_2$$

$$I_1 = cE_2 - d\left(-\frac{1}{b}E_1 + \frac{a}{b}E_2\right)$$

$$= \frac{d}{b}E_1 + \left(c - \frac{da}{b}\right)E_2$$

$$\frac{cb - da}{b} = -\frac{1}{b}$$

$$\begin{aligned} I_1 &= \frac{d}{b}E_1 + \left(-\frac{1}{b}\right)E_2 \\ I_2 &= \left(-\frac{1}{b}\right)E_1 + \frac{a}{b}E_2 \end{aligned}$$

What does positive mean?

$$\frac{d}{b} > 0 \quad \frac{ad-1}{b^2} = \frac{bc}{b^2} = \frac{c}{b} > 0$$

$$a = \frac{1}{d}(1+bc)$$

so if $\frac{c}{b} > 0$ then a, d same sign

$$510 \quad \begin{pmatrix} a & b \\ c & d \end{pmatrix} \in SL_2(\mathbb{R}) \quad \mapsto \quad \begin{pmatrix} A & B \\ B & C \end{pmatrix} = \left(\begin{array}{c|c} \frac{d}{b} & -\frac{1}{b} \\ \hline -\frac{1}{b} & \frac{a}{b} \end{array} \right)$$

Conversely assume

$$I_1 = AE_1 + BE_2$$

$$I_2 = BE_1 + CE_2$$

$$BE_2 = -AE_1 + I_1 \quad E_2 = -\frac{A}{B}E_1 + \frac{1}{B}I_1$$

$$I_2 = BE_1 + C\left(-\frac{A}{B}E_1 + \frac{1}{B}I_1\right) \quad I_2 = \frac{B^2-AC}{B}E_1 + \frac{C}{B}I_1$$

$$BE_1 = I_2 - CE_2$$

$$I_1 = AE_1 + B\left(-\frac{C}{B}E_2 + \frac{1}{B}I_2\right) + BE_2$$

$$E_1 = \left(-\frac{C}{B}\right)E_2 + \frac{1}{B}I_2$$

$$I_1 = \left(\frac{B^2-AC}{B}\right)E_2 + \left(\frac{A}{B}\right)I_2$$

$$\frac{AC}{B^2} + \frac{B^2-AC}{B^2} = 1$$

$$\begin{pmatrix} E_1 \\ I_1 \end{pmatrix} = \begin{pmatrix} -\frac{C}{B} & -\frac{1}{B} \\ \frac{B^2-AC}{B} & -\frac{A}{B} \end{pmatrix} \begin{pmatrix} E_2 \\ -I_2 \end{pmatrix}$$

$$\begin{pmatrix} A & B \\ B & C \end{pmatrix} = \begin{pmatrix} \frac{d}{b} & -\frac{1}{b} \\ -\frac{1}{b} & \frac{a}{b} \end{pmatrix}$$

$$\begin{pmatrix} a & b \\ c & d \end{pmatrix} = \begin{pmatrix} -\frac{C}{B} & -\frac{1}{B} \\ \frac{B^2-AC}{B} & -\frac{A}{B} \end{pmatrix}$$

what role
does pos. play.

$$\begin{pmatrix} E_1 \\ -I_1 \end{pmatrix} = \begin{pmatrix} -\frac{c}{B} & +\frac{1}{B} \\ \frac{Ac-B^2}{B} & -\frac{A}{B} \end{pmatrix} \begin{pmatrix} E_2 \\ +I_2 \end{pmatrix}$$

So what is the program. In general ~~terms~~ you want to relate the Sp and the indef. U-theories. You want to recover the Hilbert space.

~~the~~ Principle: the response of a port does not see the bound states.

You should try to classify 2-ports. Suppose you start with $V = V_1 \oplus V_2$ equipped with Q_s . $Q_s = \begin{pmatrix} A & B \\ B & C \end{pmatrix}$. Assuming $B \neq 0$, one can replace Q_s by a symplectic transformation depending on s . What you want to do is to factor this symplectic transf. into simpler ones. Is this possible?

You think that a 1-port gives rise to a coherent sheaf of rank 1 over the Riemann sphere, so a 2 port should give rise to a coherent sheaf of rank 2, i.e. a rank 2 vector bundle which has the form $\mathcal{O}(k) \oplus \mathcal{O}(l)$ together with a torsion sheaf - this should be the bound state part. Anyway what happens?

Let's analyze ranked 1. Let's go over the general case V/W subquotient of $H = H^+ \oplus H^-$. On H consider q.f. $s \|h^+\|^2 + s^{-1} \|h^-\|^2$ and descend to the quotient space H/W .

What do you have to get straight first. Given a ^{nondeg} q.f. on a vector space how it descends to a subspace ~~at~~ ^{and} quotient space.

512 Let $W \subset V$ let $V \xrightarrow{g} V^*$ be a quadratic form whose restriction to W is n.d.

$$\begin{array}{ccc}
 W \hookrightarrow V & \xrightarrow{g} & V^* \\
 \downarrow \text{ } & & \downarrow \text{ } \\
 W & \xrightarrow{g^W} & W^*
 \end{array}$$

Claim: 1) g induces a n.d. quad form on V/W :
 $g^{V/W}: V/W \rightarrow (V/W)^*$. Specifically $g^V(\sigma+W)$

has a unique stationary point w_0 and $g^{V/W}(\sigma+W) = g^V(\sigma+w_0)$.
 $g^{V/W}$ is the stationary value of g^V on the coset $\sigma+W$.

$$2) (g^{V/W})^{-1} = P \bar{g}^{-1} P^t$$

Do I need to amplify 1) further? What else might I say? Splitting the filtration. Examine W^\perp w.r.t g .

$$W^\perp = \{ \sigma \mid g^t(\sigma) = 0 \} = g^{-1} \text{Im}(p^t)$$

Start with W

$$\begin{array}{ccc}
 W \hookrightarrow V & \hookrightarrow & W^\perp \\
 \downarrow \text{ } & & \downarrow \text{ } \\
 W^* & \xrightarrow{g^W} & W^*
 \end{array}$$

Define $W^\perp = \{ \sigma \mid \langle W, g\sigma \rangle = 0 \}$
 bilinear form $\langle \sigma_1, g\sigma_2 \rangle$
 Claim W^\perp comp to W in V .

$W \subset V$, $g: V \rightarrow V^*$ bilinear form $\langle \sigma_1, g\sigma_2 \rangle$

assume g restricted to W is n.d. - this means $g^W: W \rightarrow W^*$ is an isom.

then $W \oplus W^\perp = V$ $W^\perp = \{ \sigma \in V \mid \langle W, g\sigma \rangle = 0 \}$

Let $w \in W \cap W^\perp$ $\langle w, g w \rangle = 0 \quad \forall w$

$\Rightarrow w \in \text{Ker } g^W \Rightarrow w = 0$. Given $\sigma \in V \exists!$

w such that $\langle w, g w \rangle = \langle w, g \sigma \rangle \quad \forall w$

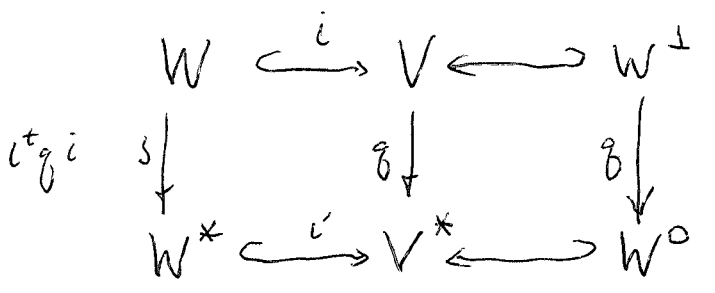
$\Rightarrow \sigma - w \in W^\perp \quad \therefore \sigma \in W + W^\perp$.

5/13 next we have $g|_{W^\perp} \subset (V/W)^* = \text{Ker } i^t$
 $W^\perp = g^{-1}(W^0)$. $\therefore g(W^\perp) \subset W^0$.

$$V = W \oplus W^\perp$$

First W^\perp is a complement to W

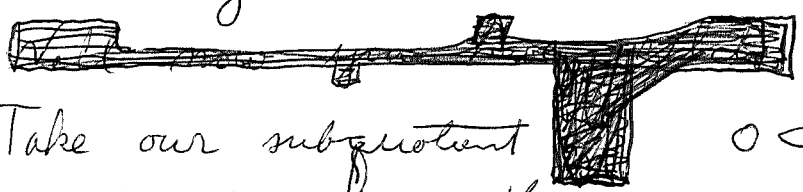
2nd $g|_W$ is isomorphism to $W^0 = (V/W)^*$



OKAY, what do we find. $\langle \sigma_1, g\sigma_2 \rangle$ bilinear form on V non-degenerate when rest. to W . Then

$$V = W \oplus W^\perp \quad \text{in fact } (V, g) = (W, g|_W) \oplus (W^\perp, g|_{W^\perp})$$

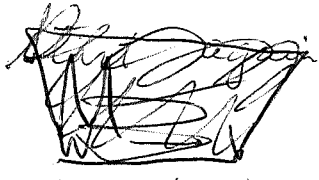
so g is ~~non-degenerate~~ non-deg $\iff g|_{W^\perp}$ is non-deg. so what. am I learning?



Take our subquotient $0 \subset W \subset V \subset H$

~~What~~ What happens is that you can do everything relative to $s=1$. In any case you have a splitting of this ~~split~~ ~~split~~ $H = W \oplus (V \ominus W) \oplus (H \ominus V)$ this splitting depends on s , but it can be referred to $s=1$. Really clean

You have on $H = H^+ \oplus H^-$ the quadratic form $s\|h^+\|^2 + s^{-1}\|h^-\|^2$, so what do you do? Gauss orthonormalization? so how can I analyze all this??
~~You start~~ You go over the idea of the ~~orth~~.



What sort of things do you want to accomplish? Clean up the result describing the quadratic form on V/W . Recover H from V/W and Q .

Recap. $Q(v_1, v_2)$ bilinear form on V

$W^\perp = \{v \mid Q(w, v) = 0 \ \forall w \in W\}$. Assume $Q(w_1, w_2)$ non-degenerate i.e. $\forall \lambda \in W^* \exists! w$ such that $\langle w_1, \lambda \rangle = Q(w_1, w) \ \forall w_1 \in W$.

$\langle w_1, Qw \rangle \quad Q: W \xrightarrow{\sim} W^*$

Then $W \oplus W^\perp \xrightarrow{\sim} V$, why?



If $w \in W \cap W^\perp$, then $Q(w_1, w) = 0 \ \forall w_1 \implies w = 0$.

Given $v \in V$ have $Q(w_1, v) \in W^*$ know $\exists! w \ni Q(w_1, v-w) = 0 \ \forall w_1 \implies v-w \in W^\perp$.

Then $Q(w_1 + z_1, w_2 + z_2) = Q(w_1, w_2) + Q(z_1, z_2) + Q(w_1, z_2) + Q(z_1, w_2)$
 as $z_2 \in W^\perp$

515 $W \subset V$ $Q(\sigma_1, \sigma_2)$ bilinear on V
 s.t. $Q(w_1, w_2)$ nondeg on W i.e.
 $w \mapsto Q(-, w)$ is an isom $W \rightarrow W^*$.

Let $W^\perp = \{v \in V \mid Q(w, v) = 0\}$

Claim $W \oplus W^\perp = V$; ~~$W \cap W^\perp = \{0\}$~~

~~$w \in W \cap W^\perp$~~ $w \in W \cap W^\perp \Rightarrow Q(w, w) = 0$
 $\Rightarrow w = 0$ since $Q(-, w)$ on W
 determines w

next given v have $Q(-, v)$ on W , know $\exists w$
 $Q(-, v) = Q(-, w) \Rightarrow Q(-, v-w) = 0 \Rightarrow v-w \in W^\perp$
 $\Rightarrow v \in W + W^\perp$. ~~$v \in W + W^\perp$~~

Critical question: Is $Q(W^\perp, W) = 0$?

~~$V = W \oplus W'$~~
 $V = W \oplus W'$
 \downarrow triangular
 $V^* = W^* \oplus W'^*$

$$Q\left(\begin{pmatrix} x \\ x' \end{pmatrix}, \begin{pmatrix} y \\ y' \end{pmatrix}\right) = \begin{pmatrix} x \\ x' \end{pmatrix}^t \begin{pmatrix} 1 & 0 \\ 0 & 1 \end{pmatrix} \begin{pmatrix} y \\ y' \end{pmatrix} = xy + x'y'$$

Let $W = \left\{ \begin{pmatrix} * \\ 0 \end{pmatrix} \right\}$

~~$Q\left(\begin{pmatrix} * \\ 0 \end{pmatrix}, \begin{pmatrix} y \\ y' \end{pmatrix}\right) = *$~~

$$Q\left(W, \begin{pmatrix} y \\ y' \end{pmatrix}\right) = Q\left(\begin{pmatrix} * \\ 0 \end{pmatrix}, \begin{pmatrix} y \\ y' \end{pmatrix}\right) = *y = 0 \quad \forall x \iff y=0$$

$\therefore W^\perp = \left\{ \begin{pmatrix} 0 \\ * \end{pmatrix} \right\}$

$$Q\left(\begin{pmatrix} 0 \\ x' \end{pmatrix}, \begin{pmatrix} y \\ 0 \end{pmatrix}\right) = x'y \neq 0.$$

$W^\perp \quad W$

5/6 Statement: $W \subset V$, Q ^(skew-) symmetric ~~form~~ ^{bilinear} form on V whose restriction to W is n.d. Then $W^\perp = \{v \mid Q(w, v) = 0, \forall w \in W\}$ is a complement to W , i.e. $V = W \oplus W^\perp$ and $Q = Q|_W \oplus Q|_{W^\perp}$.

Moreover Q is non-deg $\iff Q|_{W^\perp}$ is n.d.

What else do you need? Let's take $Q_s(h) = s \|h_+^+\|^2 + s^{-1} \|h_+^-\|^2$ on $H = H^+ \oplus H^-$

Start with $W \subset H$, you aim to compute Q_s induced on H/W .

First the restriction to W .
 Hence of a ^{canonical} splitting $H = \bigoplus_{\omega \in S} H_\omega$ compatible with $\begin{cases} H = H^+ \oplus H^- \\ H = W \oplus W^\perp \end{cases}$

such that if $V_\omega = H_\omega \cap V$ $0 < \omega < \infty$, then

V_ω ~~(is)~~ = graph of ωu , $u: H_\omega^+ \xrightarrow{\sim} H_\omega^-$ orth isom

~~...~~ $g = 1$ $F = \varepsilon = +1$ or -1 .

$$\begin{array}{l} V_0 = H_0^+ \\ V_\infty = H_\infty^- \end{array} \left(\begin{array}{l} (V_0^\perp)_0 = H_0^- \\ (V_\infty^\perp)_\infty = H_\infty^+ \end{array} \right)$$

I guess I should also consider V^\perp since I need both. ~~Go back to F~~

Maybe you can play the game with F, ε

517 Consider alternate approaches.

Go back to passing a q. form to a quotient space. Look at this on ~~an~~ a Euclidean space

You have V Euclidean with $(,)$ scalar product and $Q(v_1, v_2) = (v_1, Av_2)$ A symmetric > 0 . Next have subspace $W \subset V$, and you want induced form on W, W^\perp

$$V = W \oplus W^\perp \quad \text{orth for } (,)$$

$$i: W \hookrightarrow V \quad \text{get } i^*A i \text{ rep. } Q_W.$$

Next take $z \in W^\perp$ and choose w to minimize $(z+w, A(z+w)) =$

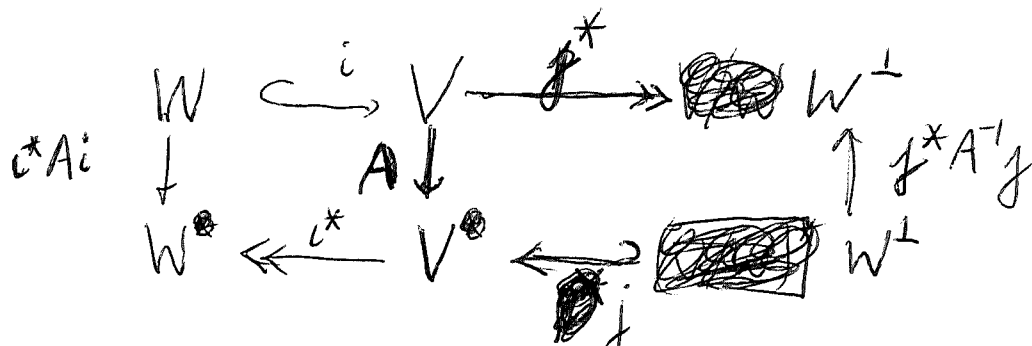
$$(\delta w, A(z+w)) + (z+w, A \delta w) = 2(\delta w, A(z+w))$$

i.e. $A(z+w) \in W^\perp$ or

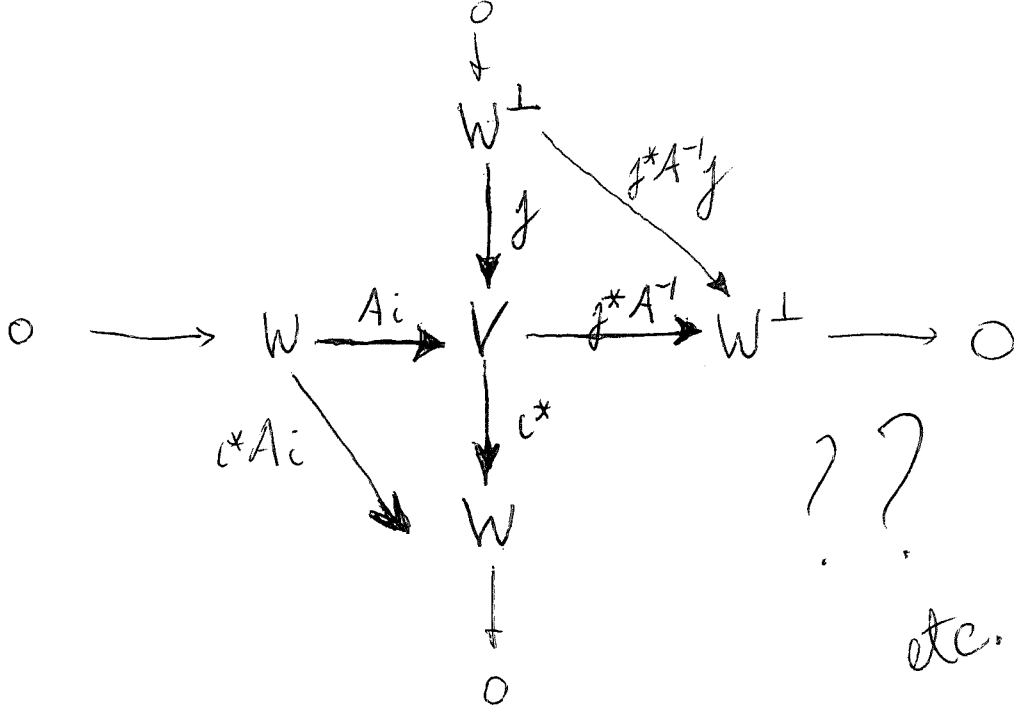
$$z+w \in A^{-1}W^\perp$$

Then you want the values of $(-, A-)$ on this element ~~of~~ $z+w$.

$$(z+w, A(z+w)) = (z, A(z+w))$$



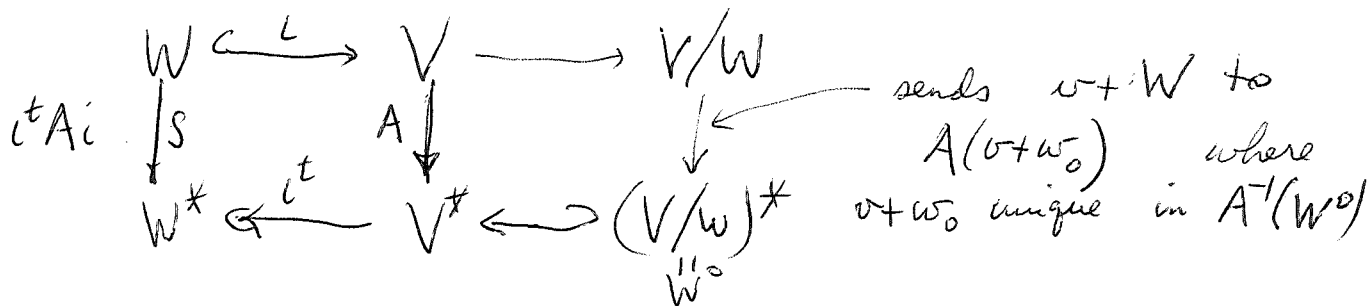
can I prove that $(j^*A^{-1}j)^{-1}$ is the quadratic form induced on V/W .



~~the stationary point~~

You have a g.f. $(\sigma, A\sigma)$ on V to be pushed down to $V/W \cong W^\perp$. $\min_{\sigma \in \sigma+W} (\sigma, A\sigma)$. Let σ_0 be the stationary point. Then $(W, A\sigma_0) = 0$ $A\sigma_0 \in W^\perp$ so $\sigma_0 \in A^{-1}W^\perp$.

Go back to $\sigma + w_0 \mapsto \sigma + W$



You take $\sigma + W \in V/W$ choose $\sigma + w_0$ to minimize the g.f. on this coset so $\langle W, A(\sigma + w_0) \rangle = 0$ $L^t A(\sigma + w_0) = 0$, so $A(\sigma + w_0) \in (V/W)^*$

So it seems you are using $V = W \oplus A^{-1}(W^\circ)$
 $\sigma = w_0 + A^{-1}(A(\sigma + w_0))$

519 So the result says that given $Q(v) = (v, Av)$ on V
 then restriction of $Q(v)$ to W is (w, i^*Aw)
 and push forward of $Q(v)$ to W^\perp is $(j^*(A^{-1})j)^T$.

$$(z + w_0, A(z + w_0))$$

min means exactly that $\underbrace{A(z + w_0)}_{z'} \in W^\perp$

$$(z, z') = ? \quad \text{where} \quad z' = A(z + w_0) \quad \& \quad z' \in W^\perp$$

$$\text{then } z + w_0 = A^{-1}(z')$$

$$\text{and } z = (j^*A^{-1}j)(z')$$

$$\therefore z' = (j^*A^{-1}j)^{-1}z$$

Jan 23 Situation: V Euclidean space with ^{pos. def.} quadratic form (v, Av) , W subspace. Have $V/W \cong W^\perp$. What is the orthogonal complement ^{of W} for bilinear form $B(v_1, v_2) = (v_1, Av_2)$? Clearly it is $A^{-1}(W^\perp)$ so we have $V = W \oplus A^{-1}(W^\perp)$.

To compute the ~~indeed~~ push forward quadratic form
 For $p: V \rightarrow V/W$ which we can identify with proj
 $j^*: V \rightarrow W^\perp$, $j: W^\perp \rightarrow V$ inclusion. Answer:
 given $z \in W^\perp$ you write $z = w + A^{-1}(z') \in W \oplus A^{-1}(W^\perp)$
 and take the values ~~(z, z)~~ $(A^{-1}z', A(A^{-1}z')) = (A^{-1}z', z')$

Thus we have ~~the same~~

Try different notation



$$z + w$$

$$z + w = A^{-1}(z')$$

$$\text{want } z \mapsto (z', A^{-1}z')$$

$$z \mapsto (A^{-1}z', A(A^{-1}z')) = (z', A^{-1}z')$$

$$\text{and you write this } (z, \bar{A}z)$$

$$\therefore (z, \bar{A}z) = (z', A^{-1}z')$$

get in trouble with

$$(z', A^{-1}z') = (z', z+w) = (z', z)$$

$$\therefore \bar{A}z = z' \quad \text{and} \quad \cancel{A} j^* A_j^{-1} (\bar{A}z)$$

$$= j^* A^{-1}(z') = j^*(z+w) = z$$

So it works, but it isn't clean yet.

Summarize: Problem of quad. forms versus operators ^{1.a.}

Look abstractly. ~~General~~ General problem -

Given V with ~~two~~ two positive quadratic forms $(-, -)$ and $(-, A-)$. Then you are given a filtration $0 \subset F_1 \subset \dots \subset F_k = V$. Each quadratic form gives a splitting. ~~There's a general~~ The A orthogonal splitting can be expressed in terms of the Euclidean one by a ~~quadratic~~ triangular matrix with positive diagonal entries. ~~What might be used~~

~~What might be used~~

What might be useful. A itself gives rise to a filtration by increasing λ E_λ projections from the spectral theorem. $A = \int \lambda dE_\lambda$. And you have the joint filtration ~~of~~ ^(Schreier combining) the two filtrations (F_p) and $(E_\lambda V)$.

Concentrate on cleaning this up as much as possible. The goal should be a dilation statement (something like GNS) for subquotients of a polarized Hilbert space. Key idea: A quadratic form on V modded on W yields a quadratic form on V/W as follows IDEA non symmetric
 V is the direct sum of W and W° . forms occur naturally with $\text{Ext} = R\text{Hom}$ on K^* Bondal, tilting autos of derived category

521 A quadratic form Q on V which is non degenerate on a subspace W yields a splitting $V = W \oplus W^\perp$ orthogonal wrt Q , hence a quadratic form Q'' on $V/W \cong W^\perp$.

$Q''(\sigma + W) =$ stationary value of $Q(\sigma + w)$ for $w \in W$.

(so you have two descriptions of the ~~splitting~~ induced form on the quotient space.

Do the above for symm. or skew-symm. bil. forms.

$B(\sigma_1, \sigma_2)$. $W^\perp = \{v \in V \mid B(w, v) = 0\}$.

$$\begin{array}{ccc} W \hookrightarrow V & \xrightarrow{k} & W^* \\ w \mapsto w & \mapsto & B(w, \cdot) \end{array} \quad \begin{array}{l} k \text{ is inv.} \\ \therefore V = W \oplus W^\perp \end{array}$$

~~old~~ $B(w_1 + x_1, w_2 + x_2) = B(w_1, w_2) + B(x_1, x_2)$

How can you describe B'' on V/W .

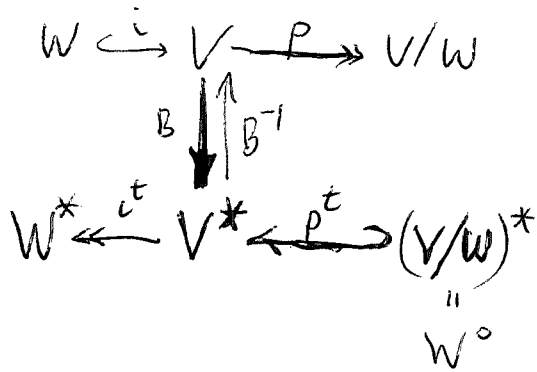
Given v_1, v_2 write $v_i = w_i + x_i$ and examine $B(\sigma_1, \sigma_2) = B(w_1, w_2) + B(x_1, x_2)$

this is a ^(quad) function of (w_1, w_2)

$$\delta B(w_1, w_2) = B(\delta w_1, w_2) + B(w_1, \delta w_2)$$

" vanishes for all $\delta w_1 \iff w_2 = 0$. \iff $\underbrace{B(w_1, \delta w_2)}_{\text{vanishes for all } \delta w_2} \iff w_1 = 0$

~~How suppose~~ $W \xrightarrow{i} V \xrightarrow{A} W^\perp$
 \downarrow
 ~~$W^* \xrightarrow{i^*} V^* \xrightarrow{A^*} W^*$~~
 $W^* \xleftarrow{i^*} V^*$



$$B^{-1}(W^0) = W^\perp$$

$$\begin{array}{l}
 W = W \oplus B^{-1}W^0 \\
 \downarrow \\
 V^* = BW \oplus W^0
 \end{array}$$

Let's leave and move to polarizations.
 But first I need a clean statement.

$$\begin{array}{l}
 V \text{ with } (\cdot, \cdot) \quad B(\sigma_1, \sigma_2) = (\sigma_1, A\sigma_2) \quad A \text{ symm.} \\
 W \subseteq V \quad \Rightarrow \quad i^*A_i : W \rightarrow W \text{ invertible}
 \end{array}$$

Claim that the induced quadratic form on V/W is $(j^*A^{-1}j)^{-1}$ $W^\perp \xrightarrow{j} V \xrightarrow{A^{-1}} V \xrightarrow{j^*} W^\perp$

~~the real~~ The real assertion I need is that given V with (\cdot, \cdot) , A s.a. > 0 , W then the s.a. of induced on V/W is $(j^*A^{-1}j)^{-1}$ where $j: W^\perp \hookrightarrow V$. Maybe better is i^*A_i for restriction and $(p^*A^{-1}p^*)^{-1}$ for a projection.

$$\begin{array}{ccc}
 W & \xrightarrow{i} & V & \xrightarrow{p} & W^\perp \\
 i^*A_i & & & & (p^*A^{-1}p^*)^{-1}
 \end{array}$$

Can we define

f

$i^*(A)$	$p_*(A)$
----------	----------

523

~~So we have a coherent sheaf.~~

Consider $H = H^+ \oplus H^-$ and a subquotient V/W of H .

$$\begin{array}{ccc} V & \xrightarrow{i} & H \\ P \downarrow & & \downarrow P' \\ V \cap W^\perp & \xrightarrow{i'} & W^\perp \end{array}$$

$$A_s = s\pi_+ + s^{-1}\pi_-$$

$$L^* A_s i = s \underbrace{(L^* \pi_+ L)}_{\text{add to } \perp} + s^{-1} \underbrace{(L^* \pi_- L)}_{\text{add to } \perp}$$

hence they commute
can be simultaneously diagonalize.

Claim
$$P_* (L^*(A_s)) = L'^*(P'_*(A_s))$$

$\therefore V = \bigoplus_{0 \leq \omega \leq \infty} V_\omega$ orth ~~or~~ direct sum

where

$$i^* \pi_+ L = \bigoplus_{\omega} \frac{1}{1+\omega^2} \pi_\omega$$

$$L^* \pi_+ i = \bigoplus_{\omega} \frac{\omega^2}{1+\omega^2} \pi_\omega$$

$$L^*(A_s) = L^* A_s L = \bigoplus_{\omega} \frac{s + s^{-1}\omega^2}{1+\omega^2} \pi_\omega$$

$$P_* (L^*(A_s)) = (P L^*(A_s)^{-1} P^*)^{-1}$$

$$P_* (L^*(A_s)) = \left(\sum_{\omega} \frac{s(1+\omega^2)}{s^2+\omega^2} P \pi_\omega P^* \right)^{-1}$$

$$P'_*(A_s) = (P' (s\pi_+ \oplus s^{-1}\pi_-)^{-1} P'^*)^{-1}$$

$$= (P' (s^{-1}\pi_+ + s\pi_-) P'^*)^{-1}$$

$$= \left(\underbrace{s^{-1}(P' \pi_+ P'^*)}_{\text{sum}} + \underbrace{s(P' \pi_- P'^*)}_{\text{add to } \perp} \right)^{-1}$$

sum \rightarrow add to \perp .

524 recovery. Given $\sum_{\omega} \frac{s(1+\omega^2)}{s^2+\omega^2} a_{\omega}$

with $a_{\omega} \geq 0$ and $\sum a_{\omega} = 1$. Each a_{ω} is a s.a. operator $0 \leq a_{\omega} \leq 1$ so can be dilated ~~to~~ to a projection in an essentially unique way.

$$a_{\omega} \quad -1 \leq 2a_{\omega} - 1 \leq 1$$

$$\begin{matrix} & & 1-a_{\omega} \\ \sqrt{1-(2a_{\omega}-1)^2} & & \\ = \sqrt{-4a_{\omega}^2+4a_{\omega}} & = & \begin{pmatrix} \frac{1+a}{2} & \sqrt{1-a^2} \\ \sqrt{1-a^2} & \frac{1-a}{2} \end{pmatrix} = \begin{pmatrix} a & \sqrt{a(1-a)} \\ \sqrt{a(1-a)} & 1-a \end{pmatrix} \end{matrix}$$

What are you doing. You have ~~$a = a^*$~~ $a = a^*$
 $0 \leq a \leq 1$ i.e. $a^2 \leq a$ and you propose to ~~write~~
~~find~~ $\sum a_{\omega} = 1$ Wait. You have

Start again with $a_{\omega} \geq 0$ $\sum a_{\omega} = 1$ on Z
 You want then to find maps $l_{\omega}: Z \rightarrow V_{\omega}$
 such that $l_{\omega}^* l_{\omega} = a_{\omega}$. Can assume $Z \rightarrow \bigoplus_{\omega} V_{\omega}$
 so you let $V_{\omega} =$ completion of Z ~~with~~ wrt
 $(z, a_{\omega} z)$
 and l_{ω} the canon map. Then

$$\begin{matrix} \text{Then } \\ \text{ } \end{matrix} \quad (l_{\omega} z_1, l_{\omega} z_2) = (z_1, a_{\omega} z_2)$$

$$\parallel$$

$$\text{ } \quad (z_1, l_{\omega}^* l_{\omega} z_2)$$

$\therefore a_{\omega} = l_{\omega}^* l_{\omega}$. Put together $Z \xrightarrow{l = (l_{\omega})} \bigoplus_{\omega} V_{\omega}$

$$\begin{pmatrix} l_{\omega}^* & \dots \end{pmatrix} \begin{pmatrix} l_{\omega} \\ \vdots \end{pmatrix} = \sum \underbrace{l_{\omega}^* l_{\omega}}_{l^* \pi_{\omega} l} = \sum a_{\omega} = 1.$$

$$l = \sum \pi_{\omega} l = \sum l_{\omega}$$

$$l_{\omega}^* l_{\omega'} = l^* \pi_{\omega} \pi_{\omega'} l = \delta_{\omega \omega'} l^* \pi_{\omega} l$$

525 This seems to work well.

I would like a connection with ~~matrix~~ S^2 if possible. ~~You have a rational m~~

You would like to construct H directly from $\sum_{0 \leq \omega \leq \infty} \frac{s(1+\omega^2)}{s^2+\omega^2} a_\omega$ $a_\omega \geq 0, \sum a_\omega = 1$

how can the sum of the residues be 1

$$\frac{s(1+\omega^2)}{s^2+\omega^2} = \frac{1+\omega^2}{2} \left(\frac{1}{s+i\omega} + \frac{1}{s-i\omega} \right)$$

$$(1+\omega^2) a_\omega = \text{Res} \frac{s(1+\omega^2)}{s^2+\omega^2} \text{ in finite plane}$$

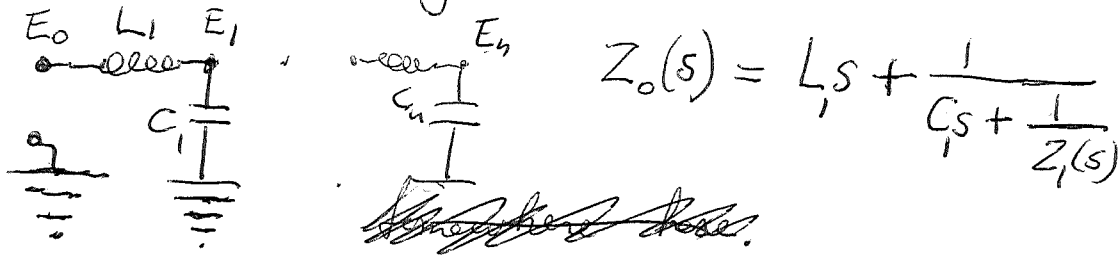
~~MM~~ Question: Whether you can directly recover H from ~~the~~ V/W and the quadratic form. Looks miraculous, but the GNS philosophy says it ought to work. So what do we hope for? Is there any link to K -modules.

Take a line $L = \mathbb{R}$ with $f(s) = \sum \frac{s(1+\omega^2)}{s^2+\omega^2} a_\omega$
 $a_\omega > 0, \sum a_\omega = 1$

~~then~~ suppose there are n - ω 's $\omega_1 < \dots < \omega_n$, say $0, \infty$ not included to simplify. So f has $2n$ poles, its denominator should be $\prod_{j=1}^n (s^2 + \omega_j^2)$ and ~~residues~~ numerator ~~should have~~ ^{has} degree $2n-1$. Thus $s = \infty$ is a zero and there ~~are~~ should be $2n-1$ roots of the numerator. Now each $a_\omega > 0$ gives rise to a line L_ω , so $\bigoplus L_\omega$ has dim n and then H will have dim $2n$ since each L_ω sits inside $H_\omega^+ \oplus H_\omega^-$ as a graph. The idea now

526 is maybe the fact that?

Wait. There should be a ^(partially) self adjoint operator with deficiency indices (1,1) in this situation somewhere. Now the ω 's we have $\pm\omega_1, \dots, \pm\omega_n$ are associated to an operator A belong to the modes where ~~are~~ the boundary condition ~~$f(s) = \infty$~~ is choose $V \neq 0$ voltage but $I = 0$ current. Try model



~~I have a picture here of an n port~~

This graph has $n+2$ vertices and $2n$ edges

The voltage space has $\dim n+1$

$$e = 2n \quad v = n+2 \quad l = n-1 \quad n+2 - 2n = 1 - (n-1) \checkmark$$

~~What's the picture to do~~ The basic picture here should ~~involve~~ involve circuits connected together. So how to proceed?? Actually we know that for each s , there is a line in (E_0, I_1) space given by.

$$\begin{pmatrix} E_0 \\ I_1 \end{pmatrix} = \begin{pmatrix} 1 & L_1 s \\ & 1 \end{pmatrix} \begin{pmatrix} 1 \\ C_1 s \end{pmatrix} \dots \begin{pmatrix} 1 & L_n s \\ & 1 \end{pmatrix} \begin{pmatrix} 1 \\ C_n s \end{pmatrix} \begin{pmatrix} E_n \\ I_{n+1} \\ 0 \end{pmatrix}$$

You have to find a mechanism to explain things. Let's start again. So ~~what~~ where do I begin?

527 Given a rational function $f(s) = \sum \frac{s(1+\omega^2)}{s^2 + \omega^2} a_\omega$
 with $a_\omega \geq 0$, $\sum a_\omega = 1$, you want to find
 a partial self-adjoint operator. What is the
 dim of the space of $f(s)$. Suppose $0 < \omega_1 < \dots < \omega_n < \infty$
 then $a_1, \dots, a_n > 0$ and $\sum a_i = 1$. Looks like
 $2n-1$ constants. $\begin{pmatrix} E_0(s) & \text{has degree } 2n \\ I_1(s) & \text{---} & 2n-1. \end{pmatrix}$

What are you trying to do? I remember
 vaguely that the real theory may use singularities
 at both $0, \infty$. What is the basic picture?

You might try ~~the~~ simplex case

The chain of $SL_2(\mathbb{R})$ matrices should
 be equivalent to ~~approximate~~ a coupling of
 \mathcal{Q} forms - graphs has vertices E_0, \dots, E_n in
 addition to ground and currents I_1, \dots, I_n
 and $2n$ edges. Wait $v = n+2$, $e = 2n$
 $1-l = v-e = n+2 - 2n = 2-n$ $l = 1+n-2 = n-1$.

$$\begin{array}{ccc} n+1 & & 2n \\ \bar{C}^0 & \subset & C^1 \\ E_0 \downarrow & & \\ \mathbb{R} & & \end{array}$$

Look the fundamental problem is to reconstruct
 the polarized Hilbert space C^1 in some form.
 This requires-

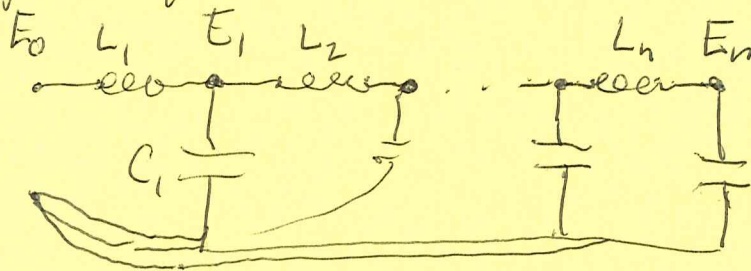
$$\begin{pmatrix} 1 & L_1 s \\ & 1 \end{pmatrix} = \begin{pmatrix} s & \\ & 1 \end{pmatrix} \begin{pmatrix} 1 & L_1 \\ & 1 \end{pmatrix} \begin{pmatrix} s & \\ & 1 \end{pmatrix}^{-1}$$

$$\begin{pmatrix} 1 & 0 \\ C_1 s & 1 \end{pmatrix} = \begin{pmatrix} s & \\ & 1 \end{pmatrix}^{-1} \begin{pmatrix} 1 & \\ C_1 & 1 \end{pmatrix} \begin{pmatrix} s & \\ & 1 \end{pmatrix} \otimes$$

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~~Wade~~ Jan 24

Problem: ~~Wade~~ Partial s.a. operators to explain response from an LC circuit. Take circuit.



$n+2$ vertices include ground

$2n$ edges

$$\begin{pmatrix} I_1 \\ E_0 \end{pmatrix} = \begin{pmatrix} 1 & 0 \\ L_1 s & 1 \end{pmatrix} \begin{pmatrix} 1 & C_1 s \\ 0 & 1 \end{pmatrix} \cdots \cdots \begin{pmatrix} 1 & C_n s \\ 0 & 1 \end{pmatrix} \begin{pmatrix} I_{n+1} = 0 \\ E_n = 1 \end{pmatrix}$$

e.g. $n=1$

$$\begin{pmatrix} I_1 \\ E_0 \end{pmatrix} = \begin{pmatrix} 1 \\ L_1 s \end{pmatrix} \begin{pmatrix} C_1 s \\ 1 \end{pmatrix} = \begin{pmatrix} C_1 s \\ 1 + L_1 C_1 s^2 \end{pmatrix}$$

response function has form $f(s) = \sum \frac{s(1+\omega^2)}{s^2+\omega^2} a_\omega$ with n ω 's (say $0, \infty$ excluded). $a_\omega \geq 0$
 $\sum a_\omega > 0$

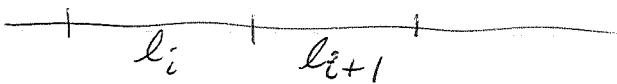
$$\frac{C_1 s}{1 + L_1 C_1 s^2} = \frac{s L_1^{-1}}{s^2 + (L_1 C_1)^{-1}} \quad \omega^2 = \frac{1}{L_1 C_1} \quad a_\omega = \frac{L_1^{-1}}{1 + (L_1 C_1)^{-1}}$$

~~I think there's a Jacobi matrix around here. ~~on~~~~
~~on an $n+1$ dim space.~~

My idea is that the rational function $f(s)$ has degree $2n$ so corresponds to a section of $O(2n)$. Hence we should be looking for a K -module of dims $2n, 2n+1$. There is a ~~matrix~~ Jacobi matrix around on ~~in~~ \mathbb{R}^n roughly; - this gives \therefore $2n+1$ parameters.

It seems that $O(2n)$ is the wrong to consider. Maybe exploit the fact that $f(-s) = -f(s)$. Take quotient of s plane by $s \mapsto -s$. Facts: $-s=s, 2s=0, s=0, \infty$. quotient is S^2 -plane, then ~~quotient~~ rational function f is equivalent to a section of $O(n)$, means looking for K -modules of dims $n, n+1$. Significance of the fact that the ~~K -module~~ roots $s^2 = -\omega^2$ are on $\mathbb{R}_{\leq 0}$?

string formulas



$$\mu_i = \frac{g_{i-1} - g_i}{l_i}$$

$$m_i s^2 g_i = -\frac{g_i - g_{i-1}}{l_i} + \frac{g_{i+1} - g_i}{l_{i+1}}$$

$$m_i s^2 g_i = \mu_i - \mu_{i+1}$$

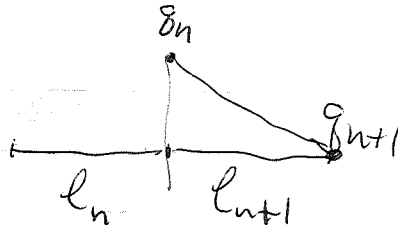
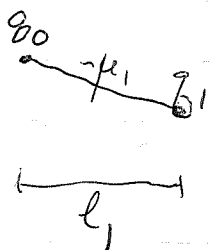
recursion relation

$$g_{i-1} = g_i + l_i \mu_i$$

$$\mu_i = \mu_{i+1} + m_i s^2 g_i$$

$$\begin{pmatrix} g_{i-1} \\ \mu_i \end{pmatrix} = \begin{pmatrix} 1 & l_i \\ 0 & 1 \end{pmatrix} \begin{pmatrix} g_i \\ \mu_i \end{pmatrix} = \begin{pmatrix} 1 & l_i \\ 0 & 1 \end{pmatrix} \begin{pmatrix} 1 & 0 \\ m_i s^2 & 1 \end{pmatrix} \begin{pmatrix} g_i \\ \mu_{i+1} \end{pmatrix}$$

$$\begin{pmatrix} g_0 \\ \mu_1 \end{pmatrix} = \begin{pmatrix} 1 & l_1 \\ & 1 \end{pmatrix} \begin{pmatrix} 1 & \\ & m_1 s^2 & 1 \end{pmatrix} \dots \begin{pmatrix} 1 & l_n \\ & 1 \end{pmatrix} \begin{pmatrix} 1 & \\ & m_n s^2 & 1 \end{pmatrix} \begin{pmatrix} g_n \\ \mu_{n+1} \end{pmatrix}$$



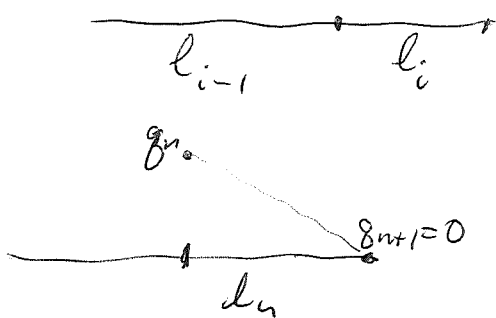
$$\begin{pmatrix} 1 & l_{n+1} \\ & 1 \end{pmatrix} \begin{pmatrix} 1 & \\ & m_{n+1} s^2 & 1 \end{pmatrix} \begin{pmatrix} g_{n+1} \\ \mu_{n+2} \end{pmatrix}$$

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g_{i-1} g_i g_{i+1}

$$m_i \ddot{g}_i = -\frac{g_i - g_{i+1}}{l_i} + \frac{g_{i-1} - g_i}{l_{i-1}}$$

$$= -\mu_i + \mu_{i-1}$$



$$\mu_{i-1} = \mu_i + m_i s^2 g_i$$

$$g_i = g_{i+1} + l_i \mu_i$$

~~$g_n = l_n \mu_n$~~

~~$g_n = l_n \mu_n$~~

~~$\mu_{n-1} = \mu_n + m_n s^2 g_n$~~

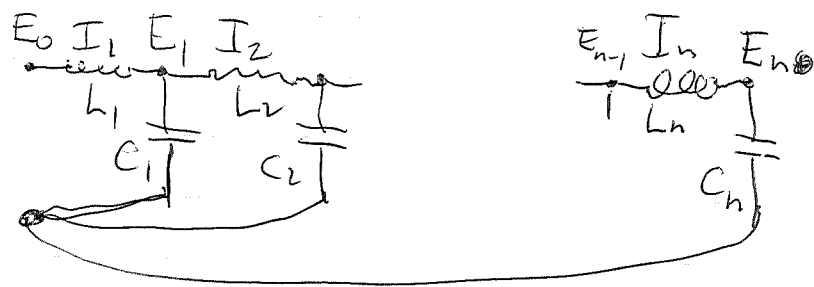
~~$g_{n-1} = g_n + l_{n-1} \mu_{n-1}$~~

$$\begin{pmatrix} g_0 \\ \mu_0 \end{pmatrix} = \begin{pmatrix} 1 & l_0 \\ & 1 \end{pmatrix} \begin{pmatrix} g_1 \\ \mu_0 \end{pmatrix} = \begin{pmatrix} 1 & l_0 \\ 0 & 1 \end{pmatrix} \begin{pmatrix} 1 & 0 \\ m_1 s^2 & 1 \end{pmatrix} \begin{pmatrix} g_1 \\ \mu_1 \end{pmatrix}$$

$$= \begin{pmatrix} 1 & l_0 \\ & 1 \end{pmatrix} \begin{pmatrix} 1 & 0 \\ m_1 s^2 & 1 \end{pmatrix} \dots \begin{pmatrix} 1 & l_n \\ 0 & 1 \end{pmatrix} \begin{pmatrix} 1 & 0 \\ m_n s^2 & 1 \end{pmatrix} \begin{pmatrix} 1 & l_{n+1} \\ \mu_n & 1 \end{pmatrix} \begin{pmatrix} 0 \\ g_{n+1} \\ \mu_n \end{pmatrix}$$

- This leads basically nowhere.
 What is your aim?? Trying to connect
 LC response to K-modules.

There is an obvious space of dim $2n+1$ namely
 $E_0, I_1, E_1, I_2, E_2, \dots, E_{n-1}, I_n, E_n$.



We have the equations relating them

$$\begin{cases} i=1, \dots, n \text{ w } I_{n+1} = 0 \\ I_i = I_{i+1} + C_i s E_i \\ E_{i-1} = E_i + L_i s I_i \\ i=1, \dots, n \end{cases}$$

unique solution depending on E_n , thus get
 a line in this $2n+1$ dim subspace dep on E_n

531 So the only question now is ~~what to do~~ ~~about~~ whether the map has ~~the~~ ~~sequ~~ ~~K~~ form, but this ^(should be) obvious

$$0 \rightarrow \mathcal{L}_s \rightarrow \mathbb{R}^{2n+1} \xrightarrow{as+b} \mathbb{R}^{2n}$$

$$as: (E_0, I_1, \dots, I_n, E_n) \mapsto (L_1 s I_1, C_1 s E_1, \dots, C_n s E_n)$$

$$b: (\dots) \mapsto (E_0 - E_1, I_1 - I_2, \dots, E_{n-1} - E_n + I_n)$$

~~Maybe you need something~~ ~~Maybe the general~~ ~~LC situation~~ ~~can be put in~~ ~~an as-b form~~ Maybe the general LC situation can be put in an as-b form

Repeat what you've learned. Consider $E_0, I_1, E_1, I_2, \dots, E_{n-1}, I_n, E_n$ $2n+1$ variables

$$\begin{cases} E_{i-1} = E_i + L_i s I_i & i=1, \dots, n \\ I_i = I_{i+1} + C_i s E_i & i=1, \dots, n \end{cases} \text{ with } I_{n+1} = 0$$

In equations, get line \mathcal{L}_s

Does this example generalize? Your previous analysis amounts to

$$\begin{array}{ccc} V^{n+1} & \hookrightarrow & H^{2n} = H_+^n \oplus H_-^n \\ E_0 \downarrow & & \text{somehow you have replaced} \\ \mathbb{R} & & \end{array}$$

Look at H with the g of $s \|h_+\|^2 + s^{-1} \|h_-\|^2$ get ~~sg~~ $s g_+ + s^{-1} g_-$ where g_+, g_- are orth forms. response function?

532 Discuss the problem. You want to find a Hilbert space interpretation of an n -port that leads to a partial unitary of codim n , hence to an n -dim vector bundle over S^2 . This may be too optimistic, but is worth pursuing.

Let's examine carefully the case we can compute. ~~Suppose~~ Suppose we start with an n -port i.e. real vector space of dim n equipped with a quadratic form depending rationally on s of the form $q(s) = \sum_{0 \leq \omega \leq \infty} \frac{s(1+\omega^2)}{s^2+\omega^2} a_\omega$ $a_\omega \geq 0$ $\sum a_\omega > 0$.

Take $n=1$. Suppose ~~the~~ $\{\omega | a_\omega \neq 0\}$ has card n label them ~~the~~ $\omega_1 \leq \omega_2 < \dots < \omega_n$ and to simplify suppose $0, \infty$ not included. Identify $V = \mathbb{R}$ so that $\sum a_\omega = 1$. Actually what is f ?

$f_s: V \rightarrow V^*$, f is rational in s of degree $2n$

To write $g = \frac{f}{g}$ quotient of two polys you convert to denom. $\prod_{i=1}^n (s^2 + \omega_i^2)$ of deg $2n$

and the numerator has degree $2n-1$. (If $\omega_n > 0$, $\omega_n = \infty$)
 You ~~make~~ convert denom. to $\prod_{i=1}^{n-1} (s^2 + \omega_i^2)$ of degree $2n-2$

and numerator has degree $2n-1$ & is divisible by s .
 so degree is $2n-1$. ~~So now take denom.~~

If $\omega_1 = 0, \omega_n < \infty$ you convert to denom. $s \prod_{i=2}^{n-1} (s^2 + \omega_i^2)$ of degree $2n-1$, numerator has

degree ~~2n-2~~ ~~2n-1~~, so degree is $2n-1$.

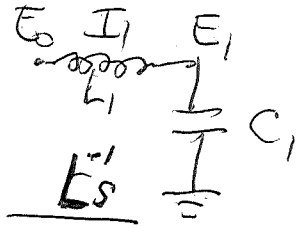
$$\frac{1}{s} + \frac{sb}{s^2 + \omega^2} = \frac{s^2(b) + s^2 + \omega^2}{s(s^2 + \omega^2)}$$

OKAY

533 Where's the action? You have a rational function of degree $2n$. A better way to put it is to say you have

$$g: \mathcal{O} \otimes V \rightarrow \mathcal{O}(2n) \otimes V^*$$

Why this Basis example.

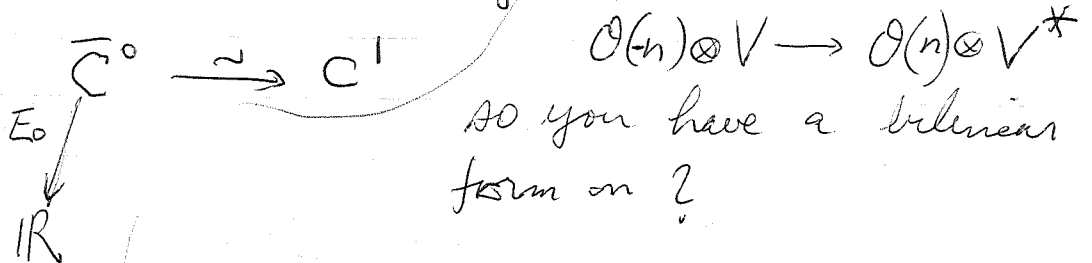


$$\frac{E_0}{I_1} = Ls + \frac{1}{Cs}$$

$$\frac{I_1}{E_0} = \frac{1}{Ls + \frac{1}{Cs}} = \frac{Ls}{s^2 + \omega^2}$$

$$\omega = \frac{1}{\sqrt{LC}}$$

This rational function has degree 2. graph has 3 vertices 2 edges



Problem: How can you go from the response function to the Hilbert space.

Approach: Look for a partial unitary

Possible approaches - maybe you should work on:

Question: Link between moment problem and LC chain

Think of $g_s: V \rightarrow V^*$ as being a map

$$\mathcal{O} \otimes V \rightarrow \mathcal{O}(+2n) \otimes V^*$$

$$\text{or } \mathcal{O}(-n) \otimes V \rightarrow \mathcal{O}(n) \otimes V^*$$

$$\mathcal{O}(2n) \otimes V \xleftarrow{\delta^{-1}} \mathcal{O} \otimes V^*$$

$$\text{" } (\mathcal{O}(-n) \otimes V)^*$$

$$\text{so get } \mathcal{O}(-n) \otimes V \xrightarrow{\delta} \mathcal{O}(n) \otimes V^* \quad ?$$

$$\mathcal{O}(n) \otimes V \xleftarrow{\delta^{-1}} \mathcal{O}(-n) \otimes V^* \quad .$$

~~What is the~~ Moment problem H Euclidean A s.a.

Eg cyclic vector, get moments $(\xi, A^n \xi)$, appropriate positive matrix on polynomials of degree $< \dim H$.

534 can orthonormalize $e \in \{, A\}$, \dots , A^{d-1}
 find A given by symm Jac. matrix | ~~orth basis~~
 orth polys wrt μ

Jan 25, 28 ~~W~~

~~What's~~ What's obvious - what's taking so long
 What's missing.

Consider a 1-port LC network

Here's the problem. ~~What's~~ You have a description for the Green's functions arising from ~~the~~ n -port LC networks, ~~but you want~~ but you want the underlying Hilbert spaces.

The LC network is ~~just a~~ a configuration space picture of something ~~that~~ you are seeking. You have the Green's function of a 2nd order ~~operator~~ DE, and you seek something like a phase space with a first order DE.

What can you do?

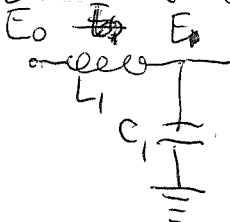
Develop your hunch that partial unitaries are the fundamental object.

Return to harmonic oscillators
 Inverse spectral problem.

Symplectic versus orthogonal

~~What's~~

Consider a 1-port LC network



$$\frac{E_0}{I_1} = L_1 s + \frac{1}{C_1 s + \frac{1}{L_2 s + \frac{1}{C_2 s + \dots}}}$$

What do I know?

~~What's~~

What do you see?

$$\frac{1}{C_n s}$$

You see

$$\begin{matrix} n+1 & & 2n & & n & & n \\ \bar{C}^0 & \xrightarrow{d} & C^1 & = & H^+ \oplus & H^- \\ E_0 \downarrow & & & & & & \\ \mathbb{R} & & & & & & \end{matrix}$$

535 But what you want is the
 vector space of dim $2n+1$ Cons. of
 $(E_0, I_1, E_1, \dots, I_n, E_n)$ Call this V^{2n+1}
 What else? $2n$ equations ~~only~~ for each edge

$$E_{i-1} = E_i + L_i I_i$$

$$I_i = I_{i+1} + C_i E_i$$

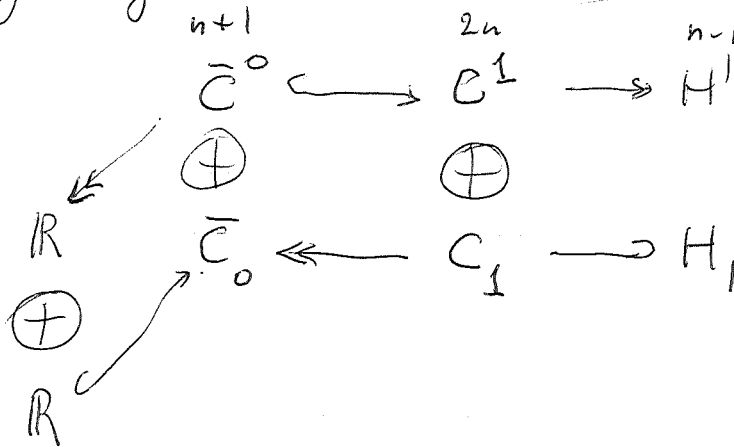
this gives $L_s \hookrightarrow V^{2n+1} \xrightarrow{as-b} W^{2n}$

~~whereas~~ so you get a line bundle over \mathbb{P}^1
 isomorphic to $\mathcal{O}(-2n)$

In terms of the data $\bar{C}^0 \hookrightarrow C^1 = H^+ \oplus H^-$
 $E_0 \downarrow$

it is not easy to see a vector space of
 dim $2n+1$. ~~mapping into \mathbb{R}^n~~ However

Maybe you should ~~use~~ form symplectic
 spaces

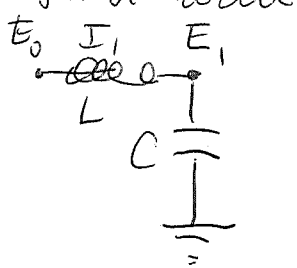


Also now to get away from configuration space.
 Try to organize things - maybe kinematics + dynamics

Problem to be solved. Given a suitable response
 function find the appropriate ~~operator~~ partial
 operator. ~~What~~ Suppose you assume ~~you are seek~~ the
 answer is a partial unitary. Then ~~what~~ you can ask if
 the response function can be transformed to \mathbb{Q} a partial
 unitary. This seems clear.

536 Additional point, Is it possible to break things into a chain of simple things

Start with $\frac{E_0}{I_1} = \frac{Ls + \frac{1}{Cs}}{Cs} = \frac{s^2 + (LC)^{-2}}{Cs}$



3 diml configuration space
vertex space + loop space) NO

in general $\begin{matrix} v-1 & e & l \\ \bar{C}^0 & \hookrightarrow & C^1 & \twoheadrightarrow & H^1 \end{matrix}$ $v-e = 1-l$

you $\bar{C}^0 \leftarrow C^1 \leftrightarrow H^1$

The ~~configuration~~ configuration space of the circuit is the space $\bar{C}^0 \oplus H^1$ which has dimension $v-1+l = e$

If I want I can think of phase space as being either $C^1 \oplus C_1$ or $(\bar{C}^0 \oplus H^1) \oplus (\bar{C}^0 \oplus H^1)$

The latter is a bit non canonical ~~since~~ since the sequence $\bar{C}^0 \hookrightarrow C^1 \twoheadrightarrow H^1$ doesn't split canonically. So you have a situation where the ~~"diluted"~~ "diluted" phase space is $C^1 \oplus \bar{C}^0$ carries response of very simple type s, s^{-1} . Then we have a

Situation: Take a ^{real} symplectic vector - The objects to consider are going to be a real symplectic vector together with a ~~rational~~ rational map from the projective line over \mathbb{R} into the symplectic Grassmannian satisfying some positivity condition. Make this precise.

537 Let M be a real symplectic vector space (f.d)

Then we have the symplectic Grassmannian

consisting of maximal isotropic subspaces of M .

If $\dim(M) = 2d$, then $\text{Sgr}(M) \subset \text{Gr}_d(M)$. If

$$M = \bigoplus_{V^*} V \quad \text{with} \quad \begin{pmatrix} \sigma_1^t \\ \lambda_1^t \end{pmatrix} \begin{pmatrix} 0 & -1 \\ 1 & 0 \end{pmatrix} \begin{pmatrix} \sigma_2 \\ \lambda_2^t \end{pmatrix} = +\lambda_1^t \sigma_2 - \sigma_1^t \lambda_2^t ?$$

then any point of $\text{Gr}_d(M)$ complementary to V^*

is the graph of a map $T: V \rightarrow$

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know $\begin{pmatrix} 1 \\ T \end{pmatrix} V$ is max. isot. \Leftrightarrow

$$\begin{pmatrix} 1 \\ T \end{pmatrix} \sigma_1^t \begin{pmatrix} 0 & -1 \\ 1 & 0 \end{pmatrix} \begin{pmatrix} 1 \\ T \end{pmatrix} \sigma_2 = \sigma_1^t T \sigma_2 + (T \sigma_1)^t \sigma_2$$

~~Better~~ $M = \bigoplus_{V^*} V \Rightarrow \begin{pmatrix} \sigma \\ \omega^t \end{pmatrix} \begin{pmatrix} \sigma_1^t & \\ \omega_1^t & \end{pmatrix} \begin{pmatrix} 0 & 1 \\ -1 & 0 \end{pmatrix} \begin{pmatrix} \sigma_2 \\ \omega_2^t \end{pmatrix}$

$$= (\sigma_1^t \ \omega_1^t) \begin{pmatrix} \omega_2^t \\ -\sigma_2 \end{pmatrix}$$

$$M = \bigoplus_{V^*} V \Rightarrow \begin{pmatrix} \sigma \\ \lambda^t \end{pmatrix} \begin{pmatrix} \sigma_1^t & \\ \lambda_1^t & \end{pmatrix} \begin{pmatrix} 0 & 1 \\ -1 & 0 \end{pmatrix} \begin{pmatrix} \sigma_2 \\ \lambda_2^t \end{pmatrix}$$

$$= (\sigma_1^t \ \lambda_1^t) \begin{pmatrix} \lambda_2^t \\ -\sigma_2 \end{pmatrix} = \sigma_1^t \lambda_2^t - \lambda_1^t \sigma_2 = \lambda_2^t \sigma_1 - \lambda_1^t \sigma_2$$

~~$\begin{pmatrix} \sigma_2 \\ \lambda_2^t \end{pmatrix} \in \mathbb{R}^2$~~
 ~~$\begin{pmatrix} \sigma_1 \\ \lambda_1^t \end{pmatrix} \in \mathbb{R}^2$~~
 ~~$\begin{pmatrix} \sigma \\ \lambda^t \end{pmatrix} \in \mathbb{R}^2$~~

$$\dim \text{Sgr}(M) = \frac{d(d+1)}{2}$$

$$\dim \text{Sp}(M) = \frac{2d(2d+1)}{2} = 2d^2 + d$$

$$\dim \text{stabilizer of } AX+B = d^2 + \frac{d(d+1)}{2}$$

GL_d symm