

465 Jan 15, 97.

~~Scattering and partial isom.~~

Link scattering and partial isom. Scattering situation. $H^2(S^1) \supset \varphi H^2(S^1)$. Is there a partial isometry assoc to this layer.

~~example~~ Do some examples of scattering.

Try perturbation viewpoint. Start with $U_0 = \text{mult by } z \text{ on } L^2(S^1)$, and consider a perturbation U_1 of U_0 . This means that $U_1 = U_0(U_0^{-1}U_1)$ where $U_0^{-1}U_1$ is a unitary ~~with~~ such that $U_0^{-1}U_1 - I$ has finite support as a matrix. This means simplest case is to let $U_0^{-1}U_1 \begin{pmatrix} \xi^n \\ \xi^m \end{pmatrix} = \begin{cases} \xi^n & n \neq 0 \\ \xi^m & n = 0. \end{cases}$

More complicated is $U_0^{-1}U_1 \begin{pmatrix} \xi^n \\ \xi^m \end{pmatrix} = \xi^n \quad n \neq 0, 1$

~~given by~~ given by ~~some~~ some 2×2 matrix.

e.g. I've put the perturbation on the right of U_0 . Suppose U_0 ~~what the shift happens.~~

$U_0^{-1}U_1$ sends ξ_0 to ξ_1
 ξ_1 to ξ_0

Then U_1 sends ξ_0 to ξ_2 bound state
 ξ_1 to ξ_1

How do you analyze? Eigenvectors u_ξ^\pm incoming and outgoing. Is it obvious these exist, i.e. that you can solve the equations: $U_1 u_\xi = \xi u_\xi$

$u_\xi = \sum c_n \xi^n$. To solve $U_1 u_\xi = \xi u_\xi$

$$U_1 U_0^{-1} U_0 u_\xi = \xi u_\xi \quad \text{let } U_0^{-1} U_0 u_\xi = \xi u_\xi$$

?

466 To solve $u_1 v^s = \int v^s$
 where $u_1 = u_0 (u_0^{-1} u_1) = u_0 + u_0 K$

Then $(u_0 + u_0 K) v^s = \int v^s$

$$(I - u_0) v^s = u_0 K v^s$$

should be a Volterra eq.

$$(I - u_0 - u_0 K) v^s = 0$$

so how do we handle this?

_____ Perturbation might be harder than necessary.

You want to solve the equation find

~~$\psi(n)$~~ $\psi(n)$ such that $\psi(n+1) = u_1 \psi(n) \quad \forall n$

$\psi(n) = u_1^n \psi(0)$ transform

$$\hat{\psi}_s = \sum_{n \geq 0} s^{-n} \psi(n)$$

~~$$u_1 \hat{\psi}_s = s \sum_{n \geq 0} s^{-n-1} \psi(n+1) = s \hat{\psi}_s$$~~

$$u_1 \hat{\psi}_s = \sum_{n \geq 0} s^{-n} \underbrace{u_1 \psi(n)}_{\psi(n+1)} = s \sum_{n \geq 0} s^{-n-1} \psi(n+1)$$

$$= s \sum_{n \geq 1} s^{-n} \psi(n) = s (\hat{\psi}_s - \psi_0)$$

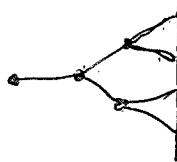
~~$$s^{-1} u_1 \hat{\psi}_s = \hat{\psi}_s - \psi_0 \quad \psi_0 = (1 - s^{-1} u_1) \hat{\psi}_s$$~~

$$\hat{\psi}_s = \frac{1}{1 - s^{-1} u_1} \psi_0$$

Now put in that $u_1 u_0^{-1} = 1 + K$

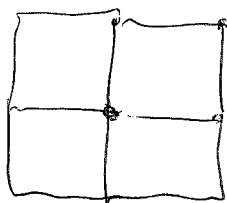
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Ojers. What is the resistance of a tree?



$$\frac{1}{2}, \left(1 + \frac{1}{2}\right) \frac{1}{2}, \left(1 + \frac{3}{4}\right) \frac{1}{2}$$

$$\frac{3}{4} \quad \frac{7}{8}$$



Use $L^2(\mathbb{Z} \times \mathbb{Z}) = L^2(S' \times S')$
 Green's function $\frac{1}{\sin^2(\frac{\theta_1}{2}) + \sin^2(\frac{\theta_2}{2})}$
 logarithmic singularity.

Go back to discrete "wave eqn". $\psi(n+1) = U_1 \psi(n)$

Transform $\hat{\psi}_\xi = \sum_{n \geq 0} \xi^{-n} \psi(n)$

$$\xi^{-1} U_1 \hat{\psi}_\xi = \sum_{n \geq 0} \xi^{-n-1} \psi(n+1) = \hat{\psi}_\xi - \psi(0)$$

$$\hat{\psi}_\xi = \frac{1}{1 - \xi^{-1} U_1} \psi(0)$$

Now use that $U_1 U_0^{-1} = 1 + K$.

$$u_1 = u_0 + (K u_0)$$

$$1 - \xi^{-1} U_1 = 1 - \xi^{-1} U_0 - \xi^{-1} (K u_0)$$

Look at inhomog. eqn.

$$(1 - \xi^{-1} U_1) \hat{\psi}_\xi = 0$$

$$(1 - \xi^{-1} U_0) \hat{\psi}_\xi = \xi^{-1} K u_0 \hat{\psi}_\xi$$

$$(1 - \xi^{-1} U_0 - \xi^{-1} U_0 K) \hat{\psi}_\xi = 0$$

$$\left(1 - \frac{1}{1 - \xi^{-1} U_0} \xi^{-1} U_0 K\right) \hat{\psi}_\xi = 0$$

$$\hat{\psi}_\xi = (G \xi^{-1} U_0 K) \psi_0$$

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$$\hat{\psi}_y = \frac{1}{1 - \int^{-1} u_1} \psi(0)$$

$$= \frac{1}{1 - \int^{-1} u_0 - \int^{-1} u_0 K} \psi(0)$$

$$U_1^n - U_0^n = \sum_{i=1}^n U_1^{i-1} (u_1 - u_0) U_0^{n-i}$$

$U_0^{-n} U_1^n$ ~~does this converge~~ as $n \rightarrow +\infty$
and $n \rightarrow -\infty$

~~Look~~ Look at $U_1^n U_0^{-n}$ as $n \rightarrow +\infty$.

If you apply this ~~matrix~~ to a vector of finite support, then you know that for n large $K U_0^{-n} \xi = 0$, hence ~~$U_0^{-n} \xi = 0$~~

$$U_1 U_0^{-1} U_0^{-n} \xi = U_0^{-n} \xi \quad \text{or} \quad U_1^{n+1} U_0^{-n-1} \xi = U_1^n U_0^{-n} \xi$$

level of Laurent polys. the wave operators exist.

Make some examples: First how about the eigenvectors.

$$\text{~~Let~~ } (1 - \int^{-1} u_0) \xi^\xi = 0$$

$$\xi^\xi = \sum c_n \xi_n$$

$$(1 - \int^{-1} u_0) \xi^\xi = \sum c_n \xi_n - c_n \int^{-1} \xi_{n+1}$$

$$= \sum (c_n - \int^{-1} c_{n-1}) \xi_n$$

$$\boxed{c_n = \int c_{n-1}}$$

$$c_n = \int^n c_0.$$

Is it clear that $\forall \xi \exists \xi^\xi$ for U_1 ?

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discrete scattering. put the perturbation

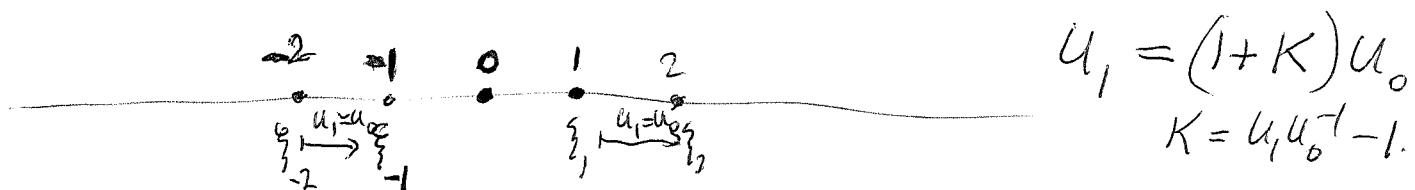
into a ~~box~~ box and see if you can ~~find~~ find in the box ~~a~~ a partial isometry. Yes!

Simple example. Suppose the perturbation ~~is~~ ~~limited to two regions~~ has support of length 1.

See if you can set up the wave operator.

~~Idea~~ Idea given ξ consider $U_1^n U_0^{-n} \xi$ as n increases. $U_1^n (U_1 U_0^{-1}) U_0^{-n} \xi$

Assume K supported at $n=0, 1$.



$$U_1 = (1+K)U_0$$

$$K = U_1 U_0^{-1} - 1.$$

for what ξ_n do we have $W^{-1} \xi_n = \xi_n$.

Thus $(U_1 U_0^{-1}) \xi_n = \xi_n$ for $n \leq -1$

$U_1 - U_0 = K U_0$ means that $U_1 \xi_n \neq U_0 \xi_n \Rightarrow n = -1$ or 0 .

Conventions so far: $U_1 - U_0 = K U_0$ $|K| = \{0, 1\}$
 $|K U_0| = \{-1, 0\}$

Try $V = \oplus \xi_{-1} \oplus \oplus \xi_0$. Then ξ_1 is a dist. line in V .

$U_1^{-1} \xi_1$ another dist. line.

Situation. Suppose you have H^- incoming, H^+ outgoing, H^-, H^+ perpendicular, let $V = (H^-)^\perp \cap (H^+)^\perp$, so that

$H = H^- \oplus V \oplus H^+$. Orth basis ξ_n $n \leq -1$ for H^-

$U \xi_n = \xi_{n+1}$ $n \leq -2$. Orth basis ξ_n for $n \geq 1$ H^+ such that

$U \xi_{n+1} = \xi_n$. Look at $U(\xi_{-1})$. Is this in V

470 Go back to a partial ~~unitary~~ ^{isom.} $W \xrightarrow{a, b} V$
 and enlarge V as follows. We have two lines
 in W , $\text{Ker}(a^*)$, $\text{Ker}(b^*)$. Call one L^+ and the
 other L^- . Combine $z^{-1}H_-^2(S', L^-) \oplus V \oplus zH_+^2(S', L^+)$

$\oplus z^{-2}L^- \oplus z^{-1}L^- \oplus V \oplus zL^+ \oplus z^2L^+ \oplus \dots$ Easy enough.
 and it seems to work.

Dominic Joyce version. Need a real version. I think
 you want to look at $L^2(S', H)$ $H = \mathbb{C}^2$.

You have singled out $\{0, \infty\}$ on S^1 . ~~It seems you need~~

Try to find the complexification of what
 you are seeking. You will have $L^2(S') \otimes \mathbb{C}^2$
 with a funny σ . Is there a unitary operator
 like $\begin{pmatrix} z^{1/2} & 0 \\ 0 & z^{-1/2} \end{pmatrix}$? Then run some perturbation analysis.

Try for $H \otimes \mathcal{O}(1)$ V 4 dim ~~sub~~
 Inside V you have two dim 2 subspaces
 and an isom between them. You need σ acting
 V^σ is 4 dim W^τ 1-dim. Get dims straight.

For $L = \mathcal{O}(1)$ you have $W = \Gamma(\mathcal{O}(L(-1))) = \mathbb{C}$ $V = \Gamma(L) = \mathbb{C}^2$.
 and $W \oplus W \xrightarrow{(a, b)} V$. In the σ -case the ~~module~~
 σK -module is simply $\mathbb{R} \rightarrow H$

Next go back to $L^2(S')^+$ perturbation leading
 to $W \rightleftharpoons \tilde{V}$. What was the point. something entering
 from left exiting from right



\leftarrow
 2 dim probably reversed by σ
 here

471 Anyway what happens to z ? The basic unitary. What do we have geometrically. You ~~take the~~ ultimately are dealing with functions on S^1 with values in \mathbb{C}^2 which are σ invariant, where σ is $z \mapsto -z$ on S^1 and something like σ on \mathbb{C}^2 . Do we have ~~any~~ anything corresponding to U ? Guess that you want $\Gamma(S^1, H \otimes O(1))$. It's hard to tell, but suppose you think of some space made out of pairs of antipodal points on S^2 . You are going to ~~get~~ get a

Discuss carefully. You have over S^2 a line bundle $O(1)$ everything with σ action except that $\sigma^2 = -1$ on $O(1)$. You have σ -invariant covering consisting of the ~~two~~ disks $|z| \leq 1$ and $|z| \geq 1$ intersecting along the circle. It's ~~more~~ meaningful to discuss ^{the smooth} sections of $O(1)$ over

Ex. Let $W \xrightarrow{a} V \xrightarrow{b}$ be a partial ism. of codim 1. Then can form $H = H^- \oplus V \oplus H^+$ with unitary U as follows

$$\begin{array}{c}
 \bigoplus_{n < 0} U^n \text{Ker}(a^*) \oplus \text{Im } b \oplus \text{Ker}(b^*) \oplus \bigoplus_{n > 0} U^n \text{Ker}(b^*) \\
 \downarrow U \qquad \searrow ab^{-1} \qquad \qquad \qquad \downarrow U \\
 \bigoplus_{n < 0} U^n \text{Ker}(a^*) \oplus \text{Ker}(a^*) \oplus \text{Im}(a) \oplus \bigoplus_{n > 0} U^n \text{Ker}(b^*)
 \end{array}$$

We have extended the partial unitary to a unitary. Actually another possibility ~~would~~ might be to extend ab^{-1} to ~~form~~ a contraction and dilate somehow.

472 However what I really want to do is to try to understand if there's a σ version. Take simple case where $\dim(V) = 2$. We are given

$$V = \text{Im}(b) + \text{Ker}(b^*)$$

$$V = \text{Ker}(a^*) + \text{Im}(a)$$

Now how can σ act? How can σ act? ~~say~~ say $\sigma^2 = -1$ on V .

Jan 16. ~~Try~~ Try to find a σ -version of a partial unitary with $\dim(V) = 2$. ~~Here~~

$$V = \text{Im}(b) \oplus \text{Ker}(b^*)$$

$$V = \text{Ker}(a^*) \oplus \text{Im}(a)$$

We want σ to act on V say $\sigma^2 = -1$. σ must act on W respect the correspondence

~~as~~ K -module $W \xrightleftharpoons[b]{a} V$ in some sense, a, b are related to $0, \infty$ pair of anti-podal points.

~~It~~ It seems logical that σ should interchange a, b . Take $V = \mathbb{H}$ with $\sigma = j$. Pick α

a line ℓ $\text{Im}(b) = \mathbb{C}(\alpha + \beta j)$ $\alpha \in \mathbb{C}$. Take

$$\text{Im}(a) = \sigma \text{Im}(b) = \mathbb{C}(-1 + \bar{\alpha} j) = \mathbb{C}((-1/\bar{\alpha}) + j)$$

It seems that ~~this is~~ $\text{Im}(a)$ is $\text{Im}(b)^\perp$. \therefore Get nothing at all.

Anyway let's go over coupling. ~~Maybe~~ Maybe a first step might be to ~~look at the~~ write up various aspects of the partial unitary ~~situation~~ situation.

Review basic construction start with f.d Hilbert spaces W, V and two isometries $W \xrightleftharpoons[b]{a} V$, then form

$$H = \underbrace{\bigoplus_{n \geq 0} U^n \text{Ker}(b^*)}_{\text{Im}(a)} \oplus \underbrace{\text{Ker}(a^*)}_{\text{Im}(b)} \oplus \underbrace{\bigoplus_{n \geq 0} U^n \text{Ker}(a^*)}_{\text{Im}(a)}$$

$$\downarrow U \quad \downarrow U$$

$$\bigoplus_{n \geq 0} U^n \text{Ker}(b^*) \oplus \text{Ker}(b^*) \oplus \text{Im}(b) \oplus \bigoplus_{n \geq 0} U^n \text{Ker}(a^*)$$

$$\downarrow U$$

$$V$$

473 What actually appears is that you have $H = H^- \oplus W \oplus H^+$ orthogonal decomp.

H^- incoming $U^{-1}H^- \subset H^-$
 H^+ outgoing $UH^+ \subset H^+$

$$\bigcap_{n \rightarrow +\infty} U^n H^- = H, \quad \bigcap_{n \rightarrow -\infty} U^n H^- = 0$$

$$\bigcup_{n \rightarrow -\infty} U^n H^+ = H, \quad \bigcup_{n \rightarrow +\infty} U^n H^+ = 0$$

Put $V = W \oplus (H^+ \ominus UH^+)$
 = orthogonal complement of $H^- \oplus UH^+$.

What do you want to say. That the basic object is a ~~group~~ H, U, H^+, H^- .

Then $W = \perp$ comp. of $H^- \oplus H^+$
 $V = W \oplus (K^+)$ $K^+ = H^+ - UH^+$
 $H^- \oplus W \oplus K^+ \oplus U(H^+)$

$$\begin{aligned} U^{-1}H^- &\subset H^- \\ UH^+ &\subset H^+ \\ H^+ &\perp H^- \\ \bigcap_n U^{-n}H^- &= 0 \\ \bigcup_n U^n H^+ &= 0 \end{aligned}$$

$$H^- \oplus U(K^-) \oplus U(W) \oplus U(H^+)$$

In this situation you have two complements for $H^- \oplus U(H^+)$ namely $W \oplus \del{K^+}$ and $U(K^-) \oplus U(W)$

So get a partial unitary.

How partial unitaries arise. Given $H \xleftrightarrow{U}$

and a subspace W of H let $V = \del{W} \oplus W + U(W)$

Is it possible to classify? Maybe what's relevant here is the scattering operator.

Problem: Classify partial unitaries. ~~What happens~~

of codim 1 to begin ~~with~~ with. You have ~~the~~ decomposition into indecomposable K -modules. It is possible that the answer is ~~the~~ a certain type of scattering operator.

474 Review what you've done. To a partial unitary $W \begin{matrix} \xrightarrow{a} \\ \xrightarrow{b} \end{matrix} V$ you've associated a scattering situation

Need discussion - Problem is to completely understand partial unitaries, then go back to partial hermitian operators.

Consider the basic construction starting from $W \begin{matrix} \xrightarrow{a} \\ \xrightarrow{b} \end{matrix} V$

$$\bigoplus_{n < 0} U^n(\text{Ker } a^*) \oplus \text{Ker}(b) \oplus \text{Ker}(b^*) \oplus \dots$$

$$\dots \oplus \text{Ker}(a^*) \oplus \text{Im}(a) \oplus \bigoplus_{n > 0} U^n \text{Ker}(b^*)$$

What is the scattering operator? If there are no bound states what you have

$$\bigoplus_{n \in \mathbb{Z}} U^n(\text{Ker } a^*) \cong H \cong \bigoplus_{n \in \mathbb{Z}} U^n \text{Ker}(b^*)$$

~~What is the scattering operator?~~ You know this is given by a bdd measurable map $S^1 \xrightarrow{\varphi} \text{Unit}(\text{Ker } a^*, \text{Ker } b^*)$

~~What is the scattering operator?~~ So how do I calculate this map? This is the basic invariant of a partial unitary.

There's lots to do before it's all clean. Look for eigenvector. $W^* \begin{matrix} \xleftarrow{a^*} \\ \xleftarrow{b^*} \end{matrix} V^*$ $L_\lambda = \text{Ker}(a^* - b^*)$ generically this is a line in V^* depending

on λ . ~~What is the scattering operator?~~ You want a map $\text{Ker}(a^*) \rightarrow \text{Ker}(b^*)$ depending on λ . Basically you have two quotient lines α ? What do you have? ~~What is the scattering operator?~~ $V = \text{Im } b + \text{Ker}(b^*)$

$= \text{Ker}(a^*) + \text{Im}(a)$. You should probably consider $\text{Ker}(b^*)$ and $\text{Ker}(a^*)$ as ~~the~~ quotient lines of $V/\text{Im } b$, $V/\text{Im } a$ of V . But then $V/(a\lambda + b)W$ is a variable quotient line joining them.

475 Continue: ~~You have a, b, W~~ You start with $W \xrightarrow[a]{a} V$ and you want some sort of maps (ultimately unitary) ~~looking~~ $V/bW = \text{Ker}(b^*)$ and $V/aW = \text{Ker}(a^*)$. Thus you want a line inside $V/bW \times V/aW$ depend on λ . We have a quotient line $V/(\lambda - b)W$. Maybe what you need is to give the complements $\text{Ker}(b^*)$ and $\text{Ker}(a^*)$. Then you do have $\text{Ker}(b^*) \times \text{Ker}(a^*) \subset V$, can follow with $V \rightarrow V/(\lambda - b)W$

(Earlier today you had a faint idea - find the bound states by intersecting $W \xrightarrow{(a, b)} V \times V$ with diagonal, or a variant of the diagonal - graph of s).

This looks promising. Check as follows. What you need are complements to $\text{Im}(b), \text{Im}(a)$. ~~then~~ you get a correspondence between these complements depending on λ i.e. $\text{Im}(b)^\perp \rightarrow V/(\lambda - b)W \leftarrow \text{Im}(a)^\perp$

Notice a choice of complements allows us to extend ab^{-1} : $\text{Im}(b) \rightarrow \text{Im}(a)$ to a map $V \rightarrow V$, which has eigenvalues. Compare viewpoints. $\text{Im}(b)^\perp$ OKAY

$$0 \rightarrow \mathcal{O}(-1) \otimes W \rightarrow \mathcal{O} \otimes V \rightarrow F \rightarrow 0$$

$$\uparrow$$

$$\mathcal{O} \otimes \text{Im}(a)^\perp$$

~~Take this~~

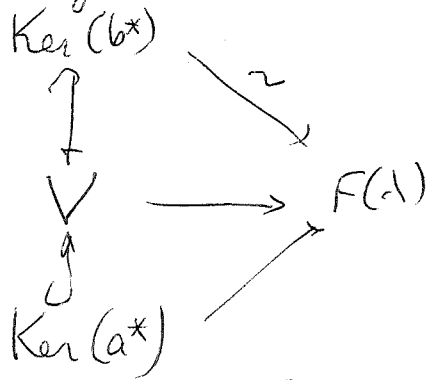
~~$$0 \rightarrow W \xrightarrow{a\lambda - b} V \rightarrow F \rightarrow 0$$~~

$$0 \rightarrow W \xrightarrow{a\lambda - b} V \rightarrow F(\lambda) \rightarrow 0$$

at $\lambda = 0$ $(a\lambda - b)W = bW$
is comp to $\text{Ker}(b^*) = (bW)^\perp$

Probably you can estimate norm $\|b^*a\| \leq 1$
 $b^*(a\lambda - b) = (b^*a)\lambda - 1$ OK for $|\lambda| < 1$.

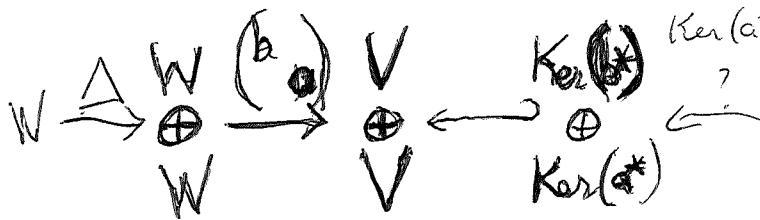
476 ~~Basic~~ Basically OKAY, namely for $|\lambda| < 1$
 you get



So we get a function $\text{Ker}(a^*) \rightarrow \text{Ker}(b^*)$ depending on λ . Needs more work tomorrow. What about now?

Compute. Suppose

$$\text{Im}(b) \oplus \text{Ker}(b^*) \cong V$$



$$\text{Ker}(a^*) \oplus \text{Im}(a) = V$$

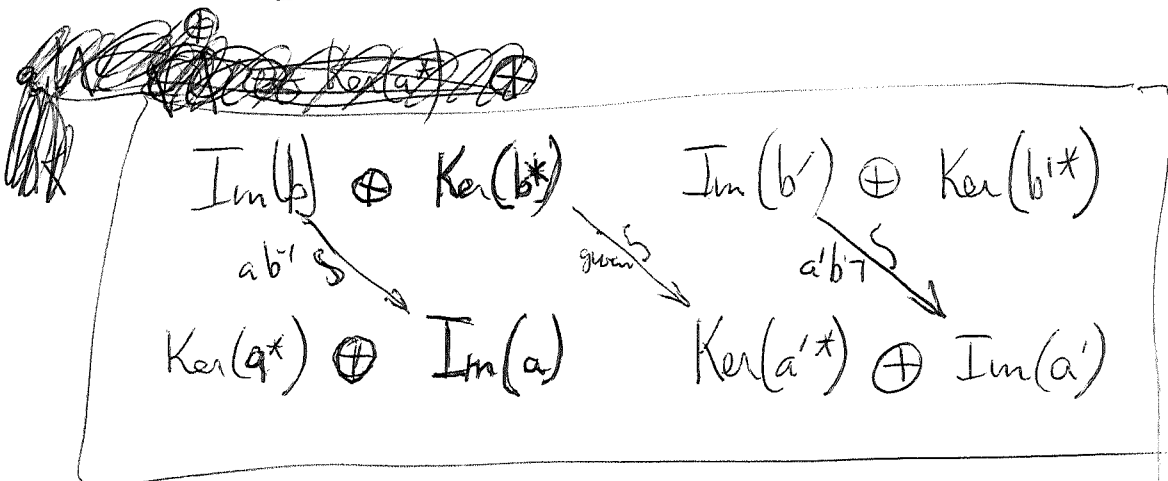
So V 2-dim with 2 splittings

$$V = \begin{array}{c} \text{Im}(b) \\ \oplus \\ \text{Ker}(b^*) \end{array} \longrightarrow \begin{array}{c} \text{Ker}(a^*) \\ \oplus \\ \text{Im}(a) \end{array}$$

$$V = \begin{array}{c} \text{Im}(b) \\ \oplus \\ \text{Im}(a) \end{array} \oplus \begin{array}{c} \text{Ker}(b^*) \\ \oplus \\ \text{Ker}(a^*) \end{array}$$

get 2x2 inv. matrix to describe splittings.

Coupling: $W \xrightleftharpoons[b]{a} V \quad X \xrightleftharpoons[b']{a'} Y$



coupling

477 These coupling pictures have to be translated into matrices.

ideas for tomorrow. Correlate your ~~LC circuit~~ LC circuit response function with ~~unitary~~ partial unitary response function. How can these be connected? Restrict to 1 ports. I think that a ~~LC circuit~~ LC circuit response function is equivalent to a partial hermitian operator - at least you can go from the LC response fn. to a ~~unitary~~ continued fraction (There may be some pitfalls).

Question: Take a Lagrange type harmonic oscillator
 $L = \frac{1}{2} \dot{q}^t m \dot{q} - \frac{1}{2} q^t k q$ | $p = \frac{\partial L}{\partial \dot{q}} = m \dot{q}$ | $\frac{d}{dt} \left(\frac{\partial L}{\partial \dot{q}} \right) = \frac{\partial L}{\partial q}$
 $m \ddot{q} + k q = 0$ Now add a external dependent on q .
 $V = \frac{1}{2} k q^2 - q^t F$ $m \ddot{q} + k q = F(t)$ $s = e^{-i\omega t}$

$$\hat{q}(s) = \int_0^{\infty} e^{-st} q(t) dt$$

$$F(t) = \text{Re}(\hat{F}(s) e^{st})$$

$\hat{q}(s) = (ms^2 + k)^{-1} \hat{F}(s)$

general response to a periodic applied force dep. on q .

~~What is the response to a periodic applied force dep. on q ?~~

Next a quotient space of position space $V \rightarrow V/W$

Say $q \mapsto q_1 = (1000)q$. Assume $\begin{pmatrix} q_1 \\ \vdots \\ q_n \end{pmatrix} \mapsto \begin{pmatrix} q_1 \\ \vdots \\ q_k \end{pmatrix}$

This isn't as straightforward as I thought

(E) $F(t) \in V^*$ V position space in q
 $F(t) \in (V/W)^*$

478 Jan 17 Lag type harm osc. $g(t) \in V \quad \forall t$

K.E. = $\frac{1}{2} \dot{g}^t m g$ ~~2~~ P.E. = $\frac{1}{2} g^t k g$ $m, k: V \rightarrow V^*$
 symm. pos. def.

equations of motion $m \ddot{g} + k g = 0.$

applied force $F(t)$ depending on t , $F(t) \in V^*$, new

~~eqn.~~ DE $m \ddot{g} + k g = F$ altered pot. energy $\frac{1}{2} g^t k g - g^t F$
 solve via L.T. ~~$\hat{g}(s) = \frac{1}{ms^2+k} \hat{F}(s)$~~ $\hat{g}(s) = \frac{1}{ms^2+k} \hat{F}(s)$ so

~~you have a map~~ you have a map $V^* \rightarrow V$ namely $(ms^2+k)^{-1}$
 so the response is a ~~quadratic~~ quadratic form on V^* dep.

rationality on s . Let V/W be a quotient space
 (~~the~~ particles you are looking at), $(V/W)^* = W^0 \subset V^*$

Then get $(V/W)^* \hookrightarrow V^*$
 $\downarrow (ms^2+k)^{-1}$
 $V/W \leftarrow V$

Conclude the response seen from outside is the restriction
 of $(ms^2+k)^{-1}$ to $(V/W)^*$. Now ^{spectral} theory for symm. ops,

say V splits canonically $V = \bigoplus V_\omega$ $V_\omega = \text{Ker}(-m\omega^2+k)$

$j: V \rightarrow V/W$
 $\downarrow \quad \nearrow j_\omega$
 V_ω

~~$(ms^2+k)^{-1} g^t = \sum_\omega \frac{1}{-m\omega^2+k} j_\omega j_\omega^t$~~

~~$\frac{1}{ms^2+k} = \bigoplus_\omega \frac{1}{ms^2}$~~

$V = \bigoplus V_\omega$ where $V_\omega = \{ \xi \mid \omega^2 m \sigma = k \sigma \}$
 where $(m^{-1}k)^{1/2} = \omega.$

$(ms^2+k)^{-1} = (s^2+m^{-1}k)^{-1} m^{-1}$

$\frac{1}{ms^2+k} = \sum_\omega l_\omega \frac{1}{s^2+\omega^2} m^{-1} l_\omega^t$ ≥ 0 on V
 $= \sum_\omega \frac{1}{s^2+\omega^2} \underbrace{(l_\omega m^{-1} l_\omega^t)}_{\text{is def on } V_\omega}$

$V = \bigoplus V_\omega$
 gives $l_\omega: V_\omega \rightarrow V$
 and $l_\omega^t: V^* \rightarrow V_\omega^*$
 $l_{\omega_1}^t l_{\omega_2} = \delta_{\omega_1, \omega_2}$
 $\sum_\omega l_\omega l_\omega^t = I_V$

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$$\frac{1}{ms^2+k} = \sum_{\omega} \frac{1}{s^2+\omega^2} \underbrace{L_{\omega}^{-1} L_{\omega}^t}_{m^{-1}}$$

add up to $m^{-1}: V^* \rightarrow V$

$$f \frac{1}{ms^2+k} f^t = \sum_{\omega} \frac{1}{s^2+\omega^2} f_{\omega} m^{-1} f_{\omega}^t \quad V_{\omega} \xrightarrow{L_{\omega}} V \xrightarrow{f} V/W$$

$$\left(\frac{1}{s-i\omega} \frac{1}{s+i\omega} \right) = \frac{2i\omega}{s^2+\omega^2}$$

Start again $g(t) \in V$, $K.E. = \frac{1}{2} \dot{g}^t m g$, $P.E. = \frac{1}{2} g^t k g - g^t F(t)$ $\left. \begin{matrix} m: V \rightarrow V^* \\ k: \end{matrix} \right\}$

$$m \ddot{g} + k g = F(t) = \text{Re}(\hat{F}(s) e^{st}) \quad s = -i\omega$$

$$\hat{g}(s) = (ms^2+k)^{-1} \hat{F}(s) \quad \omega = \sqrt{\frac{k}{m}}$$

m, k simultaneously diagonalize, ~~orth~~

canonical splitting $V = \bigoplus_{\omega > 0} V_{\omega}$ $V_{\omega} = \{g \in V \mid (m\omega^2 - k)(g) = 0\}$

$$(ms^2+k)^{-1}: V^* \rightarrow V$$

$$V = \bigoplus V_{\omega} \quad L_{\omega}: V_{\omega} \rightarrow V_{\omega} (ms^2+k)^{-1}$$

$$L_{\omega}^t: V_{\omega}^* \leftarrow V^*$$

Point is that $(ms^2+k)^{-1} = \sum \frac{1}{s^2+\omega^2}$

$$(ms^2+k)^{-1} = (m(s^2+m^{-1}k))^{-1} = (s^2+m^{-1}k)^{-1} m^{-1}$$

Again: the aim is to describe response functions assoc. to a Lagrange type harmonic oscillator

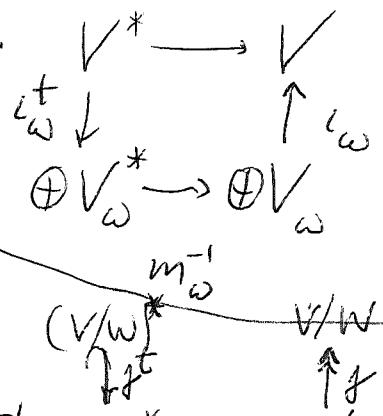
$m, k: V \rightarrow V^*$ symm pos. def.

$$\hat{g} = (ms^2+k)^{-1} \hat{F} \quad (ms^2+k)^t: V^* \rightarrow V$$

$$f (ms^2+k)^t f^t: (V/W)^* \rightarrow (V/W)$$

But $V = \bigoplus_{\omega} V_{\omega}$ $V^* = \bigoplus_{\omega} V_{\omega}^*$ $m = \bigoplus m_{\omega}$ $k = \bigoplus k_{\omega}$

$$\boxed{\omega^2 m_{\omega} = k_{\omega}}$$



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$$ms^2+k = \bigoplus_{\omega} m_{\omega}(s^2+\omega^2) \quad \text{on } \bigoplus V_{\omega}$$

$$(ms^2+k)^{-1} = \bigoplus_{\omega} \frac{1}{s^2+\omega^2} m_{\omega}^{-1} \quad : \bigoplus V_{\omega}^* \rightarrow \bigoplus V_{\omega}$$

$$\sum_{\omega} \frac{1}{s^2+\omega^2} L_{\omega} m_{\omega}^{-1} L_{\omega}^t$$

$$V_{\omega} \xrightarrow{L_{\omega}} V_{\mathbb{R}^n/V_{\omega}}$$

$$V_{\omega}^* \xleftarrow{L_{\omega}^t} V^*$$

$$f(ms^2+k)^{-1}f^t = \sum \frac{1}{s^2+\omega^2} f_{\omega}^t m_{\omega}^{-1} f_{\omega}$$

So the response function is ~~apparently~~ of the form

$$\sum_{\omega} \frac{1}{s^2+\omega^2} A_{\omega} \quad \text{where } A_{\omega} \text{ is a non-negative g.f. on } V^*.$$

$$A_{\omega} = f_{\omega}^t m_{\omega}^{-1} f_{\omega}$$

$$\sum A_{\omega} = f m^{-1} f^t \quad \text{values at } s=\infty$$

$$\sum \frac{1}{\omega^2} A_{\omega} = \sum f_{\omega}^t \frac{1}{\omega^2 m_{\omega}} f_{\omega} = f k^{-1} f^t$$

k_{ω} value at $s=0$.

Curious path in the space of pos. definite quad forms from m^{-1} to k^{-1} , not really $\dot{}$ if $f=1$, then it's the inverse of the linear path ms^2+k

Contrast with the response function for an LC-circuit.

$$\left(\frac{s^{-1} + \omega^2 s}{s^2 + \omega^2} \right)^{-1} = \frac{s^{-1} + \omega^2 s}{1 + \omega^2} = \frac{s(1 + \omega^2)}{s^2 + \omega^2} = \frac{2s}{s^2 + \omega^2} \frac{1 + \omega^2}{2}$$

$$\frac{1}{s-i\omega} + \frac{1}{s+i\omega}$$

$$\sum_{\omega} A_{\omega} \frac{s(1+\omega^2)}{s^2+\omega^2}$$

versus.

$$\sum \frac{1}{s^2+\omega^2} A_{\omega}$$

481 Go over response for an electric circuit.

$$\bar{C}^0 \xrightarrow{d} C^1 = C^{1,C} \oplus C^{1,L}$$

$$E = L \dot{I}$$

$$I = C \dot{E}$$

$$\bar{C}_0 \xleftarrow{\mathcal{Z}^t} C_1 = C_1^C \oplus C_1^L \quad \mathcal{Z}_s^{-1} = \begin{pmatrix} C_s & 0 \\ 0 & (Ls)^{-1} \end{pmatrix}$$

Set ${}^t_d \mathcal{Z}_s^{-1} d : \bar{C}^0 \rightarrow \bar{C}_0$. Write ~~matrix~~ $d = \begin{pmatrix} d^C \\ d^L \end{pmatrix}$

$${}^t_d \mathcal{Z}_s^{-1} d = \begin{pmatrix} {}^t_d C & {}^t_d L \end{pmatrix} \begin{pmatrix} C_s \\ (Ls)^{-1} \end{pmatrix} \begin{pmatrix} d^C \\ d^L \end{pmatrix} = s \begin{pmatrix} {}^t_d C d^L \\ {}^t_d L d^L \end{pmatrix} + s^{-1} \begin{pmatrix} {}^t_d L d^L \\ {}^t_d C d^L \end{pmatrix}$$

This is confusing, but becomes better when you adopt Hilbert picture. $V \subset H^+ \oplus H^-$ $\varepsilon = \begin{pmatrix} 1 & 0 \\ 0 & -1 \end{pmatrix}$ on $\begin{matrix} H^+ \\ \oplus \\ H^- \end{matrix}$

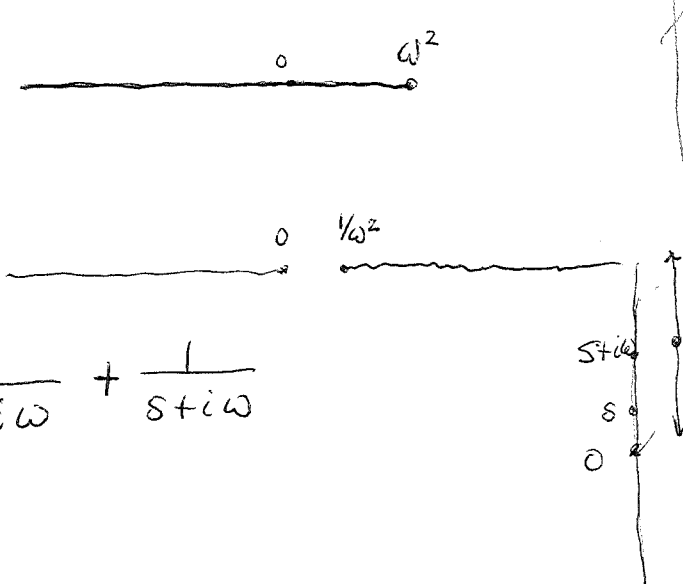
$F = +1$ on V -1 on V^\perp . Know ~~matrix~~

Compare variables. $\frac{s(1+\omega^2)}{s^2+\omega^2}$, $\frac{1}{s^2+\omega^2}$ rational function

of s . If $s \in i\mathbb{R}$, former in $i\mathbb{R}$ latter in \mathbb{R} . What about $\text{Re}(s) > 0$. Apparently $s \mapsto \frac{1}{s^2+\omega^2}$ maps $\text{Re}(s) > 0$?

$$s \mapsto s^2 \mapsto s^2 + \omega^2$$

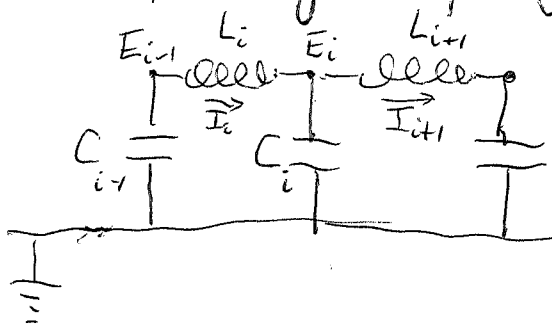
$$\{\text{Re}(s) > 0\} \mapsto \{s^2 \notin \mathbb{R}_{\leq 0}\} \mapsto \{s^2 + \omega^2 \notin \omega^2 + \mathbb{R}_{\leq 0}\}$$



$$\frac{2s}{s^2 + \omega^2} = \frac{1}{s - i\omega} + \frac{1}{s + i\omega}$$

482 The surprise is that the response of the harmonic oscillator leaves much to be desired.

Examples of coupling



$$C_i s E_i = I_i - I_{i+1}$$

$$I_i = I_{i+1} + C_i s E_i$$

$$L_i s I_i = E_{i-1} - E_i$$

$$E_{i-1} = E_i + L_i s I_i$$

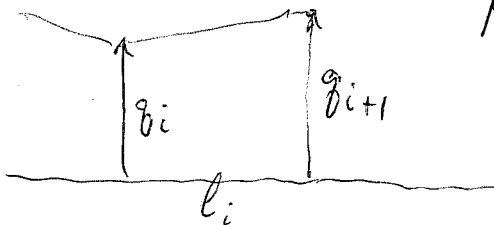
$$\begin{pmatrix} E_{i-1} \\ I_i \end{pmatrix} = \begin{pmatrix} 1 & L_i s \\ & 1 \end{pmatrix} \begin{pmatrix} E_i \\ I_i \end{pmatrix}$$

$$\begin{pmatrix} E_i \\ I_i \end{pmatrix} = \begin{pmatrix} 1 & 0 \\ C_i s & 1 \end{pmatrix} \begin{pmatrix} E_i \\ I_{i+1} \end{pmatrix}$$

$$\begin{pmatrix} E_0 \\ I_1 \end{pmatrix} = \begin{pmatrix} 1 & L_1 s \\ & 1 \end{pmatrix} \begin{pmatrix} 1 & 0 \\ C_1 s & 1 \end{pmatrix} \cdots \begin{pmatrix} 1 & 0 \\ C_n s & 1 \end{pmatrix} \begin{pmatrix} E_n \\ I_{n+1} \\ \vdots \\ 0 \end{pmatrix}$$

$$\frac{E_0}{I_1} = L_1 s + \frac{1}{C_1 s + \frac{1}{L_2 s + \frac{1}{C_2 s + \dots}}}$$

string picture



$$\mu_i = \frac{g_i - g_{i+1}}{l_i} \quad g_i$$

$$m_i s^2 g_i = \mu_{i+1} - \mu_i$$

$$m_i s^2 g_i = \underbrace{\frac{g_{i+1} - g_i}{l_i}}_{-\mu_i} - \underbrace{\frac{g_i - g_{i-1}}{l_{i-1}}}_{-\mu_{i-1}}$$

$$\boxed{\mu_{i-1} = \mu_i + m_i s^2 g_i}$$

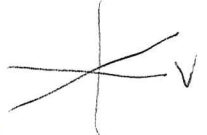
$$g_{i+1} - g_i = l_i (-\mu_i)$$

$$\boxed{g_i = g_{i+1} + l_i \mu_i}$$

$$g_i + l_i \mu_i = g_{i+1} + l_i \mu_i$$

483 Can you relate ~~partial~~ LC circuits to partial unitaries?

Idea - start from quadratic form rational in s with some positivity property. First significant point is that a quadratic form on V is equivalent to a maximal isotropic subspace of $V \oplus V^*$ (symplectic) which is transversal to V^* .

so the point may be that you have  some sort of holom map ~~partial~~ into ~~?~~?

Look carefully at $\dim 1$ case. from an LC circuit we get a rational function of s of the form $f(s) = 0$

$$\sum_{0 \leq \omega \leq \infty} a_\omega \frac{s(1+\omega^2)}{s^2 + \omega^2} \quad \text{with} \quad a_\omega \geq 0 \quad \sum a_\omega = 1$$

~~Properties~~ Properties: $f(s)$ analytic off $i\mathbb{R}$

$$\begin{aligned} \text{Re}(s) \geq 0 &\implies \text{Re}(f(s)) \geq 0. & \overline{f(s)} &= f(\bar{s}). \\ & & f(-s) &= -f(s) \end{aligned}$$

~~What can you say about the degree of f,~~
 i.e. the number of poles = $2 \times$ number of $a_\omega > 0$
 for $0 < \omega < \infty$ + 1 if $a_0 > 0$ + 1 if $a_\infty > 0$.
 number of parameters is the degree.

Next look at partial ~~unitaries~~ unitaries. Here the ~~we~~ we get $g(z)$ rational function of z such that

$$|z| \geq 1 \implies |g(z)| \geq 1. \quad \text{for } |z| \geq 1.$$

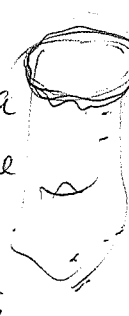
think $g(z)$ has the form $c \prod_{i=1}^n \frac{z - \alpha_i}{1 - \bar{\alpha}_i z}$. Degree of g
 = ~~number of terms~~ degree of map $S^1 \rightarrow S^1 = n$.

number of parameters to describe g is $2n+1$ real parameters, maybe ~~2n+1~~ if you ~~count~~ ^{is} the constant in S^1 .

484 There is this Schwartz lemma to construct by coupling.

Another idea is $U(1,1)$ picture. Partial unitary should give by scattering an analytic map ~~into the disc~~ of the unit disk into a symmetric space for $U(p,p)$ maximal isotropic on the bdry?

~~Go back to~~ Go back to ~~the~~ partial unitaries, describe how to calculate scattering. Next point Go back over your approach to CFT. ^{Oriented} Circle is space. ~~Consider~~ Consider smooth functions modulo constants with skew form $\int dg$ obvious real structure. ~~When~~ When the circle is boundary of the disk you get a polarization. When the circle is the boundary of surface with g holes you find not enough ~~holom.~~ bdry values of holom. functions. ~~Then try~~ ~~0~~ functions. Then the point was to try $\log f$ where f is invertible holom. function.



Anyway continue! ~~Let us decide what to do.~~ Either make notes for electric circuits, or understand partial unitaries better.

Basic construction given $W \begin{matrix} \xrightarrow{a} \\ \xleftarrow{b} \end{matrix} V$, ~~What's~~

~~important~~ here is

$$\begin{array}{ccc} \oplus U^n \text{Ker}(a^*) \oplus U^{-1} \text{Ker}(a^*) \oplus & \left| \begin{array}{c} \text{Im}(b) \oplus \text{Ker}(b^*) \\ \text{Ker}(a^*) \oplus \text{Im}(a) \end{array} \right| & \oplus U \text{Ker}(b^*) \oplus \dots \\ & \xrightarrow{ab} & \\ \oplus & \oplus & \oplus U \text{Ker}(b^*) \end{array}$$

The ~~idea~~ idea here is that we have isos (no bdy states)

$$\bigoplus_{n \in \mathbb{Z}} U^n \text{Ker}(a^*) \xrightarrow{\sim} H \xleftarrow{\sim} \bigoplus U^n \text{Ker}(b^*)$$

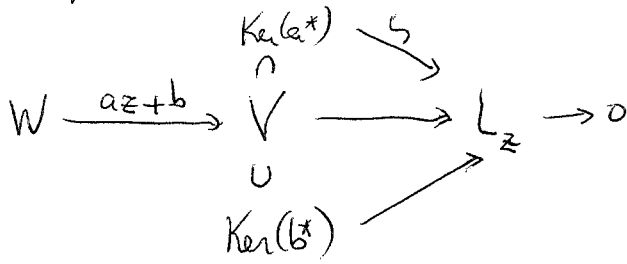
such an isom should be given by $g(z) : S^1 \rightarrow \text{Unit} \left(\begin{array}{c} \text{Ker}(a^*) \\ \text{Ker}(b^*) \end{array} \right)$
 You need control of this

485 You want to compute scattering.

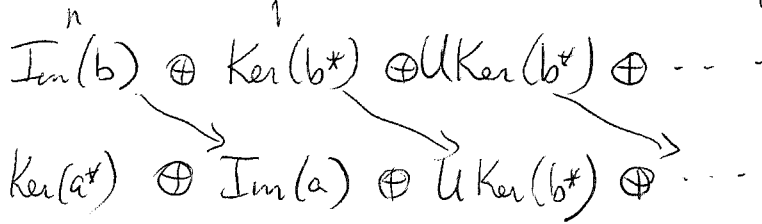
First case $\dim(V)=2$ $\dim(W)=1$. You want a ~~map~~

~~map~~ map $\text{Ker}(a^*) \rightarrow \text{Ker}(b^*)$ depending on z .

Idea:



Alternative approach - look at inclusion of outgoing subspaces

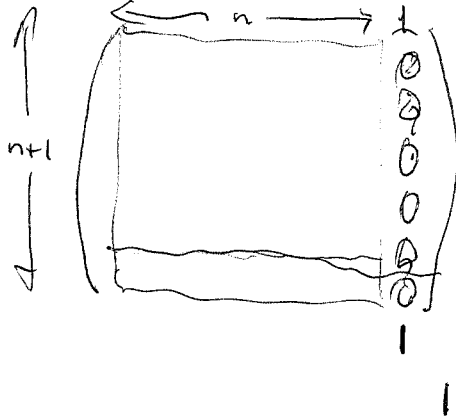


So the idea is to take $\text{Im}(b) \oplus (\text{Ker}(b^*) \oplus U\text{Ker}(b^*) \oplus \dots)$

~~Make a guess~~ Make a ~~shrewd~~ guess. Start with $bW \subset V$

Choose basis so that $bW = \mathbb{C}^n \subset \mathbb{C}^{n+1} = V$.

$$|a_{11}|^2 + |a_{12}|^2 = 1$$



$$n=1$$

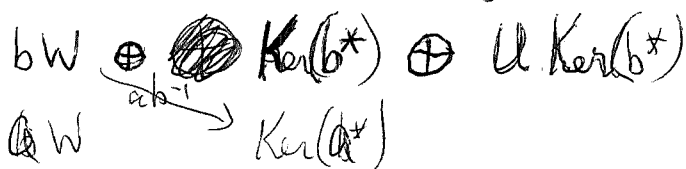
$$\mathbb{C} \subset \mathbb{C}^2$$

$$a \begin{pmatrix} 1 \\ 0 \end{pmatrix} = \begin{pmatrix} a_{11} \\ a_{12} \end{pmatrix}$$

$$S = \begin{bmatrix} a_{11} & 0 & 0 & 0 \\ a_{12} & 0 & 0 & 0 \\ 0 & 1 & 0 & 0 \\ 0 & 0 & 1 & 0 \end{bmatrix}$$

this is an isometry $S^*S = I$.

OKAY - this clarifies things, namely in general given $W \xrightarrow[a]{a} V$ partial unitary you can form a partial isom and so forth.



486 You feel somehow that it should be possible to write down eigenvectors for this operator. ~~It is possible for me to regard ~~it as a task~~~~

$$S = \begin{pmatrix} a_{11} & & & \\ a_{12} & 0 & & \\ & 1 & 0 & \\ & & 1 & 0 \\ & & & \dots \end{pmatrix}$$

Use the structure theory. ~~What is the spectrum of S~~ What is the spectrum of S all of $|z| < 1$. ~~What is the spectrum of S~~ $\text{Ker}(S^* - \lambda) \neq 0$

$$S^* \sum_{n \geq 0} c_n S^n \xi = \sum_{n \geq 1} c_n S^{n-1} \xi + c_0 S^* \xi$$

$$\lambda \sum_{n \geq 0} c_n S^n \xi = \sum_{n \geq 1} \lambda c_{n-1} S^{n-1} \xi$$

need $c_0 S^* \xi = 0$ and $c_n = \lambda c_{n-1} \Rightarrow c_n = \lambda^n c_0$

$$\sum_{n \geq 0} \lambda^n S^n \xi$$

where $S^* \xi = 0$.

$$\xi = \begin{pmatrix} \bar{a}_{12} \\ -\bar{a}_{11} \\ 0 \\ 0 \end{pmatrix}$$

$$S \xi = \begin{pmatrix} a_{11} \bar{a}_{12} \\ a_{12} \bar{a}_{12} \\ -\bar{a}_{11} \end{pmatrix}$$

$$S^* \xi - \lambda \xi = \begin{pmatrix} \bar{a}_{11} - \lambda & \bar{a}_{12} & 0 \\ -\lambda & 1 & 0 \\ -\lambda & 1 & 0 \\ -\lambda & 1 & 0 \end{pmatrix}$$

$$\begin{aligned} (\bar{a}_{11} - \lambda) c_1 + \bar{a}_{12} c_2 &= 0 \\ -\lambda c_2 + c_3 &= 0 \\ -\lambda c_3 + c_4 &= 0 \end{aligned}$$

$$\begin{aligned} \bar{a}_{12} c_2 &= (\lambda - \bar{a}_{11}) c_1 & c_3 &= \lambda c_2 \\ c_2 &= \left(\frac{\lambda - \bar{a}_{11}}{\bar{a}_{12}} \right) c_1 & c_4 &= \lambda c_3 \end{aligned}$$

$$|a_{11}|^2 + |a_{12}|^2 = 1$$

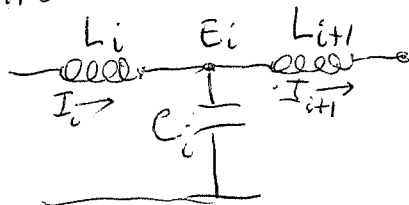
If $|a_{12}|^2 = 0$, then $aW = bW$.

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so it seems we have a eigenvector
~~use~~ z^0, z^1, z^2, \dots as basis for $\text{Ker}(b^*) \cup \text{Ker}(b^*)$

Then get $1 + \frac{\lambda - \bar{a}_{11}}{\bar{a}_{12}} (z + \lambda z^2 + \lambda^2 z^3 + \dots)$

What do we need to connect transmission lines. First derive transmission line equations.



$$C_i \dot{E}_i = I_i - I_{i+1}$$

$$L_i \dot{I}_i = E_{i-1} - E_i$$

$$\gamma \partial_t E = -\partial_x I$$

$$\lambda \partial_t I = -\partial_x E$$

$$\left[\begin{array}{l} \gamma \partial_t E = -\partial_x I \\ \lambda \partial_t I = -\partial_x E \end{array} \right]$$

~~capacitance~~
 γ capax / length
 λ induct / length

$$\lambda \gamma \partial_{tt}^2 E = -\lambda \partial_{tx}^2 I = \partial_{xx}^2 E$$

$$\left(\partial_{xx}^2 - \lambda \gamma \partial_{tt}^2 \right) E = 0$$

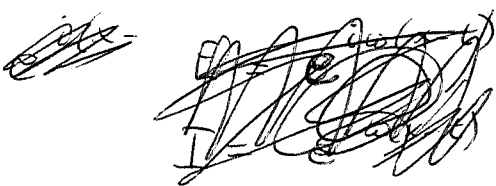
$\frac{1}{\sqrt{LC}}$

$$e^{i(kx - \omega t)} \hat{E}$$

$$-k^2 + \lambda \gamma \omega^2 = 0$$

$$\left(\frac{\omega}{k} \right) = \frac{1}{\sqrt{\lambda \gamma}} = c \text{ freq. indep.}$$

On a ~~transmission~~ transmission line you have waves travelling in two directions. Assume $\gamma = \lambda = 1$



$$\partial_t E + \partial_x I = 0$$

$$\partial_t I + \partial_x E = 0$$

$$E = \hat{E} e^{ik(x-t)}$$

$$I = \hat{I} e^{ik(x-t)}$$

$$\partial_t E = -ik \hat{E} e^{ik(x-t)}$$

$$\partial_x I = ik \hat{I} e^{ik(x-t)}$$

$$\hat{E} = \hat{I}$$

$$\begin{pmatrix} \hat{E} \\ \hat{I} \end{pmatrix} = \begin{pmatrix} \hat{E} \\ \hat{I} \end{pmatrix} e^{-ik(x+t)}$$

$$\hat{E} = -\hat{I}$$

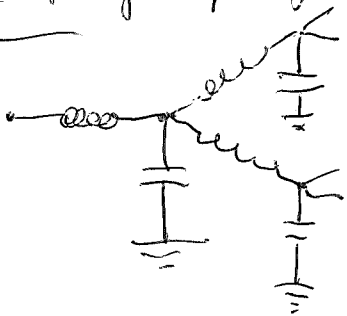
So if I use transmission line need both directions.

488 Jan 18, 98 Return to computing scattering
~~that~~ operator for a ^{partial} ~~incomplete~~ unitary. Your
 missing the ~~the~~ resolvent type calculation.

IDEA: Cramer's rule ~~gives~~ ^{maybe gives} a formula
 for the eigenvector. In the case of ~~a~~ a Jacobi
 matrix the ~~successive~~ orthogonal polys might
 be given by characteristic polys.

Eigenvectors - really the incoming and outgoing ones.
 Maybe there's a T_c matrix appropriate to the ^{partial} unitary
 based on ~~the~~ expressing it as an iterated coupling

Examples: Study in the $SL_2(\mathbb{Z})$ tree the LC circuit with
 capacitances at the vertices and inductances for the edges.
 Mechanically considering a discrete string vibrating transversally
 to the plane with masses at the vertices and weightless
 string along the edges. Other possibilities ~~are~~ would be a
~~uniform density~~ string with mass (uniform to begin
 with) so that one has waves. You ^{maybe} can consider ~~the~~
 vibrations in the plane of the string to get a different 3 way
 coupling at the vertex. Limitations - All these things
 seem to ~~involve~~ bring in the C^* alg generated by $SL_2(\mathbb{Z})$
 which should contain free subgroups of finite ~~at~~ index -
 in fact, there should be a map $\mathbb{Z}/4 * \mathbb{Z}/2 \mathbb{Z}/6 \rightarrow \mathbb{Z}/12$, whence
 a free subgroup of index 12.



$$Z = Ls + \frac{1}{Cs + \frac{2}{Z}} = Ls + \frac{Z}{CsZ + 2}$$

$$Z(CsZ + 2) - Ls(CsZ + 2) = Z$$

$$(Cs)Z^2 + \underbrace{(2 - LCs^2 - 1)}_{1 - LCs^2}Z - 2Ls = 0$$

$$\begin{aligned} b^2 - 4ac &= (1 - LCs^2)^2 - 4(Cs)(-2Ls) \\ &= 1 - 2LCs^2 + L^2C^2s^4 + 8LCs^2 \\ &= 1 + 6LCs^2 + (LCs^2)^2 \end{aligned}$$

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natural osc. for $\overset{L}{\text{---}} \text{---} \overset{C}{\text{---}}$ is $\omega_0 = \frac{1}{\sqrt{LC}}$

$$Z = Ls + \frac{1}{Cs} = \frac{Ls^2 + 1}{Cs}$$

$$LC(\omega_0^2) + 1 = 0$$

assume $\omega_0 = 1$

$$Z = \frac{-(1-s^2) \pm \sqrt{1+6s^2+s^4}}{2Cs}$$

$$s^2 = -3 \pm \sqrt{8}$$

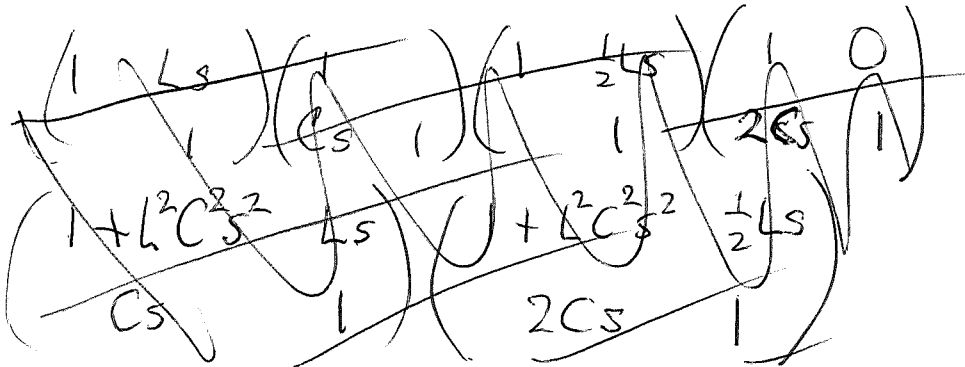
$$= -3 \pm \frac{2\sqrt{2}}{2.808}$$

$$\approx \frac{-(1-s^2) \pm \sqrt{(s^2+5.8)(s^2+0.2)}}{2Cs}$$

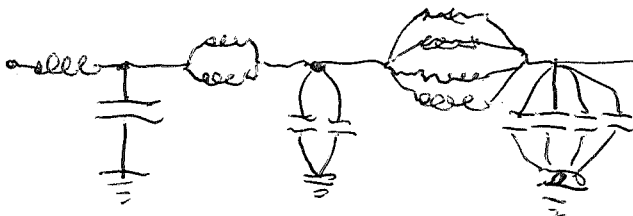
$$2Z = 2Ls + \frac{2}{Cs + \frac{2}{Z}} = (2L)s + \frac{1}{(\frac{C}{2})s + \frac{1}{Z}}$$

$$Z = Ls + \frac{1}{Cs + \frac{1}{\frac{1}{2}Z}}$$

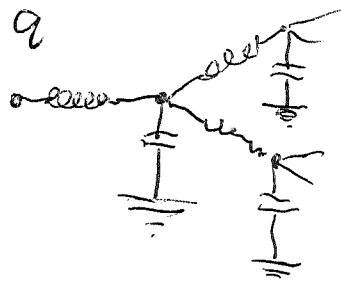
$$= Ls + \frac{1}{Cs + \frac{1}{\frac{1}{2}Ls + \frac{1}{2Cs + \frac{4}{Z}}}}$$



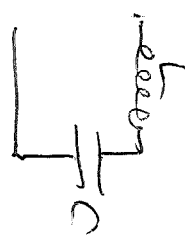
By symmetry



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$$Z(s) = Ls + \frac{1}{Cs + \frac{2}{Z(s)}}$$



$$Z = Ls + \frac{Z}{2Cs + 2}$$

$$Z(2Cs + 2) = Ls(2Cs + 2) + Z$$

$$LCs^2 + 1$$
~~$$\omega^2 = \frac{1}{LC}$$~~

$$Z^2(Cs) + (2 - 1 - LCs^2)Z - 2Ls = 0$$

$$Cs Z^2 + (1 - LCs^2)Z - 2Ls = 0$$

$$Z = \frac{-(1 - LCs^2) \pm \sqrt{(1 - LCs^2)^2 + 8LCS^2}}{2Cs}$$

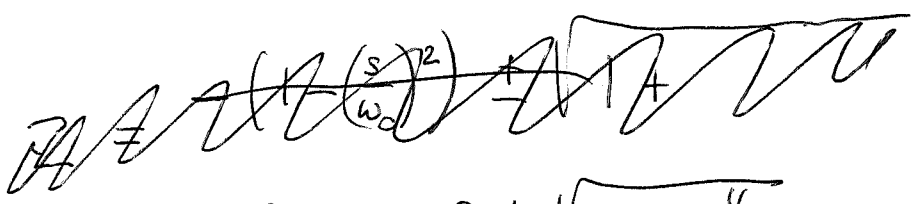
$$b^2 - 4ac = (1 - LCs^2)^2 - 4(Cs)(-2Ls)$$

$$= 1 - 2LCs^2 + L^2C^2s^4 + 8LCS^2$$

$$= 1 + 6LCS^2 + L^2C^2s^4$$

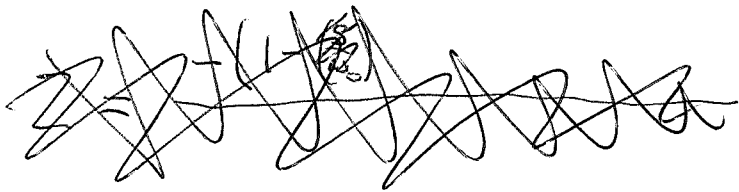
$$\omega_0 = \frac{1}{\sqrt{LC}}$$

$$= 1 + 6\left(\frac{s}{\omega_0}\right)^2 + \left(\frac{s}{\omega_0}\right)^4$$




$$\left(\frac{s}{\omega_0}\right)^2 = \frac{-6 \pm \sqrt{36 - 4}}{2} = \frac{-6 \pm 4\sqrt{2}}{2} = -3 \pm 2\sqrt{2}$$

$$b^2 - 4ac = \left(\left(\frac{s}{\omega_0}\right)^2 + 5.8\right) \left(\left(\frac{s}{\omega_0}\right)^2 + .2\right)$$



~~Let a and b be partial unitaries~~
~~with $\dim \text{Im } a = \dim \text{Im } b = 1$~~

Observation concerning a ribbon graph.

 Put a ^{uniform measure} string at each edge - associate with each side of the edge waves travelling ~~counter~~ is the pos. direction.

Couple at a vertex ^{an} incoming waves to the appropriate outgoing wave in the cyclic order.

Philosophy. A partial unitary of codim 1 can be coupled to a ~~partial unitary~~ ^{transmission line} to yield a ~~partial unitary~~ scattering function (What is the meaning of a real partial unitary?)

What is a 2-port? A partial unitary of codim 2 where $\text{Ker}(a^*)$ and $\text{Ker}(b^*)$ are ^{each} split into orthogonal lines. Similarly for a 3 port.

Simplest kind of 1-port has $\dim(V) = 1$ ~~dim(V) = 1~~ $W=0$.

$$U^{-1} \text{Ker}(a^*) \oplus \text{Im}(b) \oplus \text{Ker}(b^*)$$

$$\oplus \text{Ker}(a^*) \oplus \text{Ker } a$$

have $\text{Ker}(a^*) = \text{Ker}(b^*) = V$ and the scattering matrix is the identity. Next comes $\dim(V) = 2$. $\dim(W) = 1$.

You might get something ^{cleaner} ~~rather~~ provided you ^{specified} ~~specify~~ $\text{Ker}(a^*)$ and $\text{Ker}(b^*)$ first.

~~The~~ Data an isomorphism $\text{Ker}(a^*) \oplus W \simeq W \oplus \text{Ker}(b^*)$

Can you compose? Given $\text{Ker}(b^*) \oplus X \simeq X \oplus \text{Ker}(c^*)$, then get $\text{Ker}(a^*) \oplus W \oplus X \simeq W \oplus \text{Ker}(b^*) \oplus X \simeq W \oplus X \oplus \text{Ker}(c^*)$.

491 So what? Try then to identify
 an unitary isomorphism $K \oplus W \simeq W \oplus L$ with
 a scattering operator $g(z): K \rightarrow L$ analytic
 $|z| < 1$ unitary boundary values. You form

$$\begin{array}{c} \cdots \oplus U^{-2}K \oplus U^{-1}K \oplus W \oplus L \oplus UL \oplus \cdots \\ \searrow \quad \downarrow \quad \swarrow \\ \cdots \oplus U^{-1}K \oplus K \oplus W \oplus UL \oplus \cdots \end{array}$$

This is somehow the Eilenberg trick.

Notice that an isom. $K \oplus W \simeq W \oplus L$ is ~~the~~
~~same as~~ a stable isomorphism ^{between} K and L . Does
 a stable isomorphism between K and L lead to an
 isomorphism $K[t, t^{-1}] \simeq L[t, t^{-1}]$? Does a
 stable isom. have a characteristic poly? Certainly you
 get a K -module

$$\begin{array}{c} W \xrightarrow{\quad} K \oplus W \\ \searrow \quad \quad \quad \parallel \\ \quad \quad \quad \quad \quad V \\ \quad \quad \quad \quad \quad \parallel \\ \quad \quad \quad \quad \quad W \oplus L \end{array}$$

How much can you do algebraically? Or is
 this all linked to Hilb. space. Actually what
 do you learn $K \oplus W \simeq W \oplus L$ yield

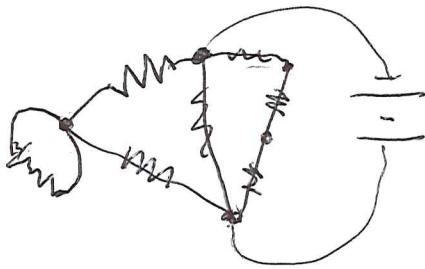
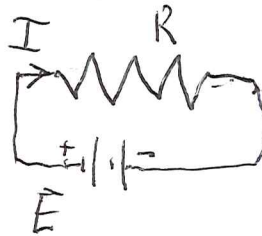
$$K \oplus K \oplus W \simeq K \oplus W \oplus L \simeq W \oplus L \oplus L$$

$$K[u^{-1}] \oplus W \simeq W \oplus L$$

491 for tomorrow's lecture.

Resistance networks.

Ohm's Law $E = RI$



Conn graph

$e = \dim C_i$
 $v-2$ conditions

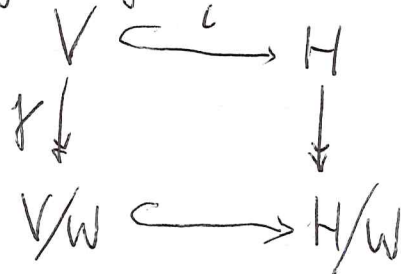
Bott's approach. Introduce ~~metrics~~ inner products on C^0, C^1 . V ?

Look, you have to straighten out the difficulty ~~with~~ arising from Hilbert space viewpoint and quadratic form viewpoint. ~~Quadratic form is~~

~~Hilbert space~~ LC case



Key argument



On H we have a quadratic form $s \|h_+\|^2 + s^{-1} \|h_-\|^2$ restrict to V get $\langle \sigma, (sL_+^* L_+ + s^{-1} L_-^* L_-) \sigma \rangle$.

~~Given~~ Given pos. def. form Q on H , how do you get the induced form on V/W ? ~~Probably~~ Probably minimize Q on the coset $v+W$.

$\bar{Q}(v+W) = \inf_w Q(v+w)$ Coset

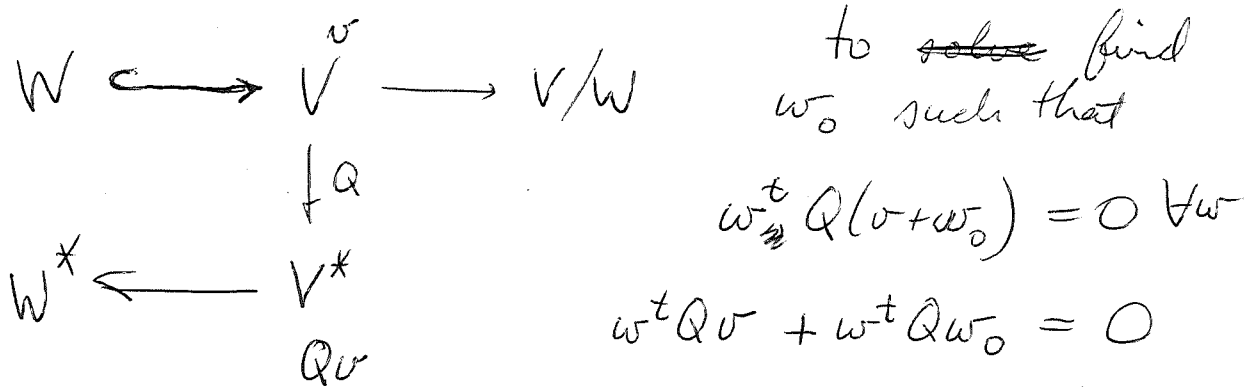
~~say~~ say $Q(v) = \langle v, Qv \rangle = v^t Qv$

492 Assume $Q(v+w)$ min. at w_0 . Then

$$\begin{aligned} \delta(\sigma^t Q \sigma) &= (\delta \sigma)^t Q \sigma + \sigma^t Q (\delta \sigma) \\ &= (\delta w)^t Q \sigma + \sigma^t Q (\delta w) \\ &= 2(\delta w)^t Q \sigma \end{aligned}$$

$\therefore Q\sigma \perp W$.

Start again. Given $v \in V$. Let w_0 be $\sigma(v+w_0)$ is min $\Rightarrow Q(v+w_0) \perp W$.



Question: Given $Q(v, v)$

symm. pos. def bilinear form

Choose $w_0 \in W$ s.t.

$Q(v+w_0, v+w_0)$ min.

Then $Q(v+w_0, w) = 0$

$Q(W, v+w_0) = 0$

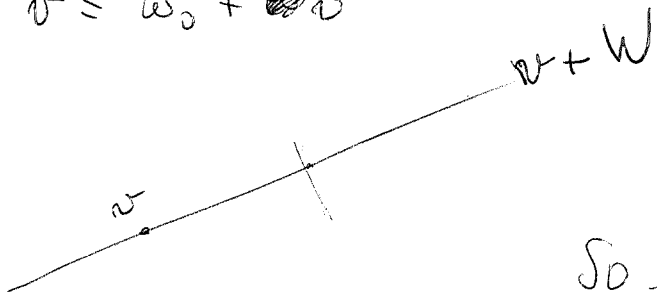
i.e. $v+w_0 \in W^\perp$ wrt Q . Thus you write

$v = w_0 + v'$

Point is $Q(v, -)$

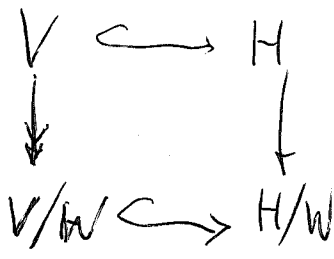
on W is represented $Q(w_0, -)$

$\therefore Q(v-w_0, w) = 0$.



So from concrete viewpoint. you take v to $Q(v, -)$, then

invert $Q|_W$



quadratic form pos. def on H
 restrict to V then push down
 to V/W or first push down
 to H/W then restrict to V/W

in both cases you get the restrict of the q form to the orthogonal complement of $W \subset V$. How do you calculate it? Take $v+W \in V/W$, ~~if~~ if $v \perp W$ nothing further needed, otherwise look at $Q(v, -)$ in W^* , use Q rest. to W gives iso $W \rightarrow W^*$ to write $Q(v, -) = Q(w, -)$, and then $v-w \perp W$. Notice ^{to do} explicitly you need to invert $W \rightarrow W^*$. So start with $Q = s\|h_+\|^2 + s^{-1}\|h_-\|^2$ on H. Look at ~~the~~ the arb. subspace W and calculate the restriction?

Given $K \oplus W \xrightarrow{\sim} W \oplus L$ unitary iso f.d. Hilb. spaces.

Use to define a unitary U on $\bigoplus_{n \leq 0} U^n K \oplus W \oplus \bigoplus_{n > 0} U^n L = H$

Look at this as a ~~mod~~ module over $\mathbb{C}[u, u^{-1}]$ containing K and L, so we get maps.

$$\bigoplus_{n \in \mathbb{Z}} U^n K \longrightarrow H \longleftarrow \bigoplus_{n \in \mathbb{Z}} U^n L$$

Commuting with U. You also have projections in the other direction. Algebraically it might be interesting to ask when the bound states split off leaving the scattering isom.

Actually something ~~is~~ subtle seems to be happening here. You seem to be getting a lot of structure out of a correspondence. Suppose you were to consider $K, L \subset V$ and an isom $V/K \cong V/L$. in ~~contrast~~ contrast to what you looked at already $\text{Im}(b), \text{Im}(a) \subset V$ and ~~the~~ the isomorphism $ab^{-1}: \text{Im}(b) \cong \text{Im}(a)$. You seem now to have some kind of spectral theory

494 Jan 19. Today's lecture in 4 hours.

Let's spend the next few hours getting formulas straight. Subspace $V \subset H^+ \oplus H^-$. Suppose $V = \begin{pmatrix} 1 \\ 1 \end{pmatrix} H^+$

$$\varepsilon = \begin{pmatrix} 1 & 0 \\ 0 & -1 \end{pmatrix}$$

$$\varepsilon = \varepsilon^* = \varepsilon^{-1}$$

$$F = +1 \text{ on } V \\ -1 \text{ on } V^\perp$$

$$F^2 = F^* = F^{-1}$$

group generated by 2 inv. is a dihedral group. $g = F\varepsilon$

$$\varepsilon g \varepsilon^{-1} = \varepsilon F \varepsilon \varepsilon = \varepsilon F = g^{-1}$$

Get splitting into indecomposable orthogonal reps. of dihedral group.

$$g = +1$$

$$\varepsilon = F = 1 \\ \varepsilon = F = -1$$

~~one dim~~

$$g = -1$$

~~$$\varepsilon = F = 1 \\ \varepsilon = F = -1$$~~

$$\varepsilon = 1 = -F \\ \varepsilon = -1 = F$$

V 1-diml.

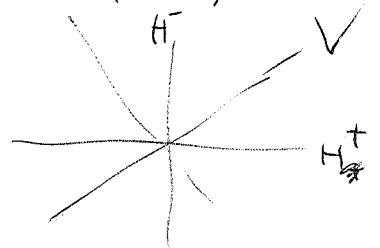
$$g = \begin{pmatrix} \cos \theta & -\sin \theta \\ \sin \theta & \cos \theta \end{pmatrix} \quad \varepsilon = \begin{pmatrix} 1 & 0 \\ 0 & -1 \end{pmatrix} \quad \text{on } \mathbb{R}^2$$

$$0 < \theta < \pi/2 < \theta < \pi$$

$$F = \begin{pmatrix} \cos \theta & \sin \theta \\ \sin \theta & -\cos \theta \end{pmatrix}$$

$$V = \begin{pmatrix} \cos \theta \\ \sin \theta \end{pmatrix} \mathbb{R} \\ = \begin{pmatrix} 1 \\ \tan \theta \end{pmatrix} \mathbb{R}$$

$$V^\perp = \begin{pmatrix} \sin \theta \\ -\cos \theta \end{pmatrix}$$



$$V = \begin{pmatrix} 1 \\ 1 \end{pmatrix} H^+ \quad \text{put } X = \begin{pmatrix} 0 & -T^* \\ T & 0 \end{pmatrix} \quad \text{skew adj.}$$

$$g = \frac{1+X}{1-X}$$

~~$$g = \frac{1+X}{1-X} = F \varepsilon (1+X) = F \varepsilon$$~~

~~Handwritten scribbles~~

$$V = \begin{pmatrix} 1 \\ 1 \end{pmatrix} H^+$$

$$V^\perp = \begin{pmatrix} -T^* \\ 1 \end{pmatrix} H^-$$

$$F \begin{pmatrix} 1 & -T^* \\ T & 1 \end{pmatrix} = \begin{pmatrix} 1 & +T^* \\ T & -1 \end{pmatrix} = \begin{pmatrix} 1 & -T^* \\ T & 1 \end{pmatrix} \begin{pmatrix} 1 & 0 \\ 0 & -1 \end{pmatrix}$$

$$F(1+X) = (1+X)\varepsilon$$

$$1+X = F(1+X)\varepsilon = F\varepsilon(1-X) = g(1-X)$$

495

$$F = \frac{(1+x)^2}{1-x^2} \varepsilon = \begin{pmatrix} 1-T^* & -2T^* \\ 2T & 1-T^* \end{pmatrix} \begin{pmatrix} 1+T^* \\ 1+T^* \end{pmatrix}^{-1} \begin{pmatrix} 1 & 0 \\ 0 & -1 \end{pmatrix}$$

$$= \begin{pmatrix} \frac{1-T^*}{1+T^*} & \frac{+2T^*}{1+T^*} \\ \frac{2T}{1+T^*} & -\frac{1-T^*}{1+T^*} \end{pmatrix}$$

not what's important

back to stuff on quadratic forms.

$$V \hookrightarrow H^+ \oplus H^- = H$$

$$\begin{matrix} \downarrow & & \downarrow \\ V/W & \hookrightarrow & H/W \end{matrix}$$

Have $Q_s(h) = s\|h_+\|^2 + s^{-1}\|h_-\|^2$ on H . Get induced quadratic form on V/W .

$$V \xrightarrow{T} V^* \quad \text{same as } \begin{matrix} v \mapsto (v' \mapsto \langle T_0, v' \rangle) \\ \langle T_0, v_2 \rangle \end{matrix} \quad \text{bilinear form.}$$

$$V = V^{**} \xrightarrow{T^t} V^* \quad \begin{matrix} v \mapsto (\lambda \mapsto \langle \lambda, v \rangle) \mapsto (\lambda \mapsto \langle \lambda, T_0 \rangle) \\ (v' \mapsto T_0 v' \mapsto \langle T_0, v' \rangle) \end{matrix}$$

$$T^t(v) \text{ is } v' \mapsto \langle T_0, v' \rangle \quad T = T^t \Leftrightarrow \langle T_0, v \rangle = \langle T_0, v' \rangle.$$

T nondeg. when T an isom. $T^{-1}: V^* \rightarrow V = (V^*)^*$

$$T^{-1}(\lambda_1, \lambda_2) = \langle \lambda_1, T^{-1}\lambda_2 \rangle = \langle T^*(T^{-1}\lambda_1), T^{-1}\lambda_2 \rangle$$

Focus upon your aim.

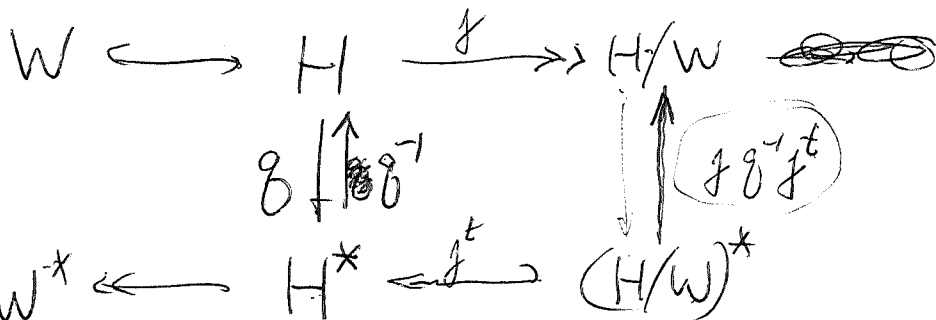
$$V \quad H \quad \begin{matrix} s, s^{-1} \\ g_s: H \rightarrow H^* \end{matrix}$$

$$\begin{matrix} V & \hookrightarrow & H \\ \downarrow & & \downarrow g_s \\ V/W & \hookrightarrow & H/W \\ \downarrow & & \downarrow g_s \\ V^* & \xleftarrow{i^*} & H^* \\ \downarrow & & \downarrow \\ V^* & \xleftarrow{i^*} & H^* \end{matrix}$$

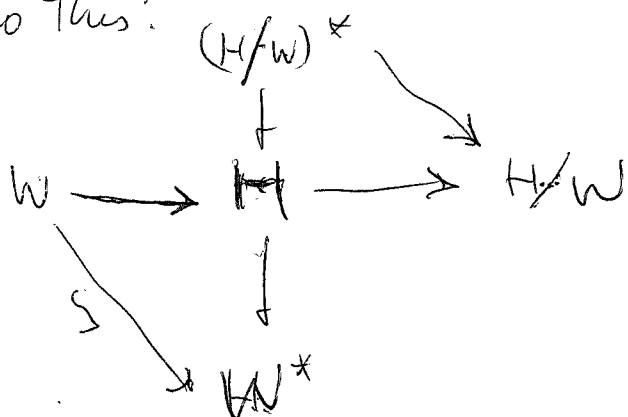
s, s^{-1}

~~Handwritten scribbles~~

$$g: H \rightarrow H^*$$



How to do this?



$$\lambda = \cos \theta$$

$$V \subset H^+ \oplus H^-$$

$$\frac{1}{2}(F\varepsilon + \varepsilon F) = \lambda \in (-1, 1)$$

Remove where $g = F\varepsilon$ is -1 .

$$\omega = \frac{\sqrt{1-\lambda^2}}{\lambda}$$

$$F\varepsilon = -1$$

splits into

$$\varepsilon = +1$$

$$F = -1$$

$$V \perp \cap H^+$$

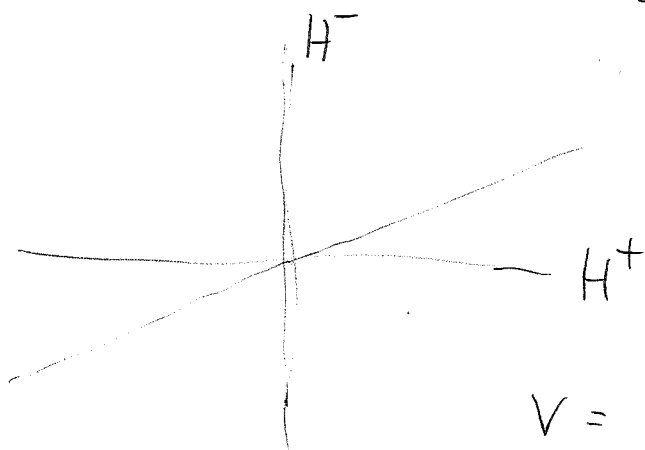
$$\varepsilon = -1$$

$$F = +1$$

$$V \cap H^-$$

$$\begin{cases} V^+ = H^+ \\ H^- = 0 \end{cases}$$

$$V = H^-, H^+ = 0$$



~~Handwritten scribbles~~

$$V = \begin{pmatrix} 1 \\ \tau \end{pmatrix} H^+$$

$$V^\perp = \begin{pmatrix} -\tau^* \\ 1 \end{pmatrix} H^-$$

$$X = \begin{pmatrix} 0 & -\tau^* \\ \tau & 0 \end{pmatrix}$$

skew adj.

$$F \begin{pmatrix} 1 & -\tau^* \\ \tau & 1 \end{pmatrix} = \begin{pmatrix} 1 & -\tau^* \\ \tau & 1 \end{pmatrix} \begin{pmatrix} \varepsilon & 0 \\ 0 & -\varepsilon \end{pmatrix}$$

$$F(1+X) = (1+X)\varepsilon$$

$$F = (1+X)\varepsilon(1+X)^{-1} = \overbrace{(1+X)(1-X)^{-1}}^g \varepsilon$$

$$\tau\tau^* = \omega^2$$

Idea: $W \subset V$ Euclidean space, Q quadratic form (pos. def) on V , ~~the~~ A the corresp. ^{pos. s.a.} op: $Q(v) = (v, Av)$.

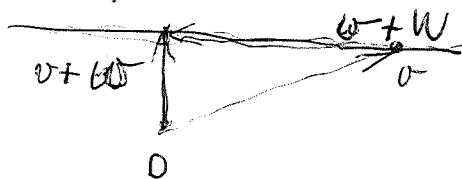
(1) Restriction of Q to W has the corresp op = the composition $W \hookrightarrow V \xrightarrow{A} V \xrightarrow{i^*} W$

(2) Induced by Q form on $V/W = W^\perp$ has the corresp op ~~is the~~ inverse of

$$W^\perp \xrightarrow{j} V \xrightarrow{A^{-1}} V \xrightarrow{j^*} W^\perp$$

Proof of (1) $Q(w) = (w, Aw) = (i^*w, Ai^*w) = (w, \underline{i^*Ae}w)$

(2) What is the induced form on V/W take coset $v+W$



so you select $w \in W$ such that $Q(v+w)$ is minimum and then take this minimum value.

$$g^{V/W}(v) = \inf_{w \in W} Q(v+w) \quad \text{at a min point.}$$

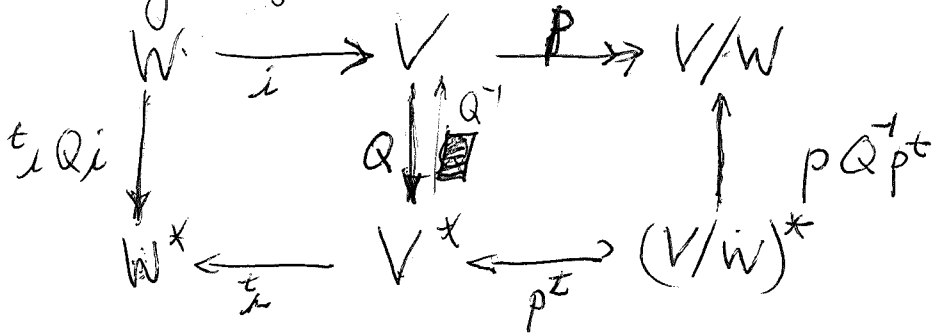
$$Q(v+w) = (v+w, A(v+w))$$

At the minimum point $(v+w, A(v+w)) = 0$ or $A(v+w) \perp W$ or $A(v+w) \in W^\perp$. or $v+w \in A^{-1}W^\perp$ Thus

$$\begin{array}{ccccccc} A(v+w) & \hookrightarrow & A(v+w) & \hookrightarrow & v+w & \hookrightarrow & v \\ \cap & & \cap & & \cap & & \cap \\ W^\perp & \hookrightarrow & V & \xrightarrow{A^{-1}} & V & \xrightarrow{j^*} & W^\perp \end{array}$$

498 Make ~~several~~ several attempts. ~~Start~~

~~with~~ Try again -



$$({}^t i Q_i w_1, {}^t i Q_i w_2) = ({}^t i Q_i w_1, Q_i w_2)$$

Q1 First method: Assume ${}^t i Q_i$ is an isom. Then

$$\begin{array}{ccccccc}
 \text{Consider } 0 & \longrightarrow & W & \xrightarrow{i} & V & \xrightarrow{P} & V/W \longrightarrow 0 \\
 & & & & \downarrow Q & & \\
 0 & \longleftarrow & W^* & \xleftarrow{{}^t i} & V^* & \xleftarrow{P^t} & (V/W)^* \longleftarrow 0
 \end{array}$$

Assume ${}^t i Q_i$ is an isom, Then $r = ({}^t i Q_i)^{-1} {}^t i Q$ is a retraction for i : $r i = 1$, so the ^{top} exact sequence also $Q_i ({}^t i Q_i)^{-1}$ is a section of ${}^t i$, so the bottom exact sequence splits. ~~What happens?~~ In this case we do get an induced map. What happens?

I know that $Q_i W$ is a complement for $W^0 = (V/W)^*$
~~Q_i W~~ $K = \text{Ker}({}^t i Q) = Q^{-1}(W^0)$

$$V = iW \oplus Q^{-1}(W^0)$$

$$V^* = Q_i W \oplus W^0$$

Still ~~is~~ unclear.