

59 What can you do tomorrow??

You are now free to work out deB's formulas. ~~deB's~~

reflection positivity. Suppose given a V ~~vector space~~ f.d. Hilbert space and a family $\phi(t) \in \mathcal{L}(V)$ for $t \geq 0$ such that $\phi(0) = 1$, $\phi(t+t') = \phi(t)\phi(t')$.

Free field theory, Gaussian
Think about real ^{Gaussian} stochastic process

$$Y \xrightarrow{\begin{pmatrix} c\varepsilon^* + A^* \\ \oplus \\ \mathbb{C} \end{pmatrix}} X \xrightarrow{\varepsilon} Y$$

Review yesterday's formulas.

$$\varepsilon = \frac{1}{2}(a+b)$$

$$c\varepsilon + A = ca$$

$$A = \frac{1}{2}(a-b)$$

$$c\varepsilon^* + A^* = \frac{1}{2}(a^* + b^*) - \frac{1}{2}(a^* - b^*) = cb^*$$

$$(c\varepsilon^* + A^*)\varepsilon = cb^* \frac{1}{2}(a+b) = \frac{1}{2}(1 + b^*a)$$

$$(c\varepsilon^* + A^*)A = cb^* \frac{1}{2}(a-b) = \frac{1}{2}(1 - b^*a)$$

I think ~~to~~ you ~~of~~ should work in $Y \oplus Y$ as much as possible. Unitary picture $\|y_1\|^2 - \|y_2\|^2$

$$W = \begin{pmatrix} a \\ b \end{pmatrix} X \subset \begin{pmatrix} Y \\ Y \end{pmatrix}, \quad W^\circ = W \oplus \begin{pmatrix} \text{Ker}(a^*) \\ \oplus \\ \text{Ker}(b^*) \end{pmatrix}$$

What is interesting? $\begin{pmatrix} 1 \\ z \end{pmatrix} Y$ is isotropic for the hermitian form when $|z| = 1$. $W^\circ \cap \begin{pmatrix} 1 \\ z \end{pmatrix} Y$ is a

line consisting of $y_1 = ax + v^+$ where $zy_1 = y_2$
 $y_2 = bx + v^-$

Solutions of $(az - b)x = -zv^+ + v^- \Rightarrow S(z)zv^+ = v^-$
 $S(z): V^+ \rightarrow V^-$
 $(1 - bb^*) / (1 - zab^*)^{-1}$

61 You want to derive a spectral representation for ~~the~~ element of y . Review simplest version.

$$\begin{array}{ccc}
 & & \downarrow e_{n+1} \\
 X & \xrightarrow{(\lambda \varepsilon - A)} & Y \\
 & & \downarrow \varepsilon^* \\
 & & X
 \end{array}$$

$$y \xrightarrow{\begin{pmatrix} \varepsilon^* \\ e_{n+1}^* \end{pmatrix}} X \oplus \mathbb{C} \xrightarrow{(\lambda \varepsilon - A \ e_{n+1})} Y$$

$$y = (\lambda \varepsilon - A \ e_{n+1}) \begin{pmatrix} \varepsilon^* \\ e_{n+1}^* \end{pmatrix} (\lambda - \varepsilon^* A)^{-1} y$$

$$\tilde{y}(\lambda) = e_{n+1}^* (\lambda - A_{n+1})^{-1} y$$

$\tilde{y}(\lambda)$ is a ~~real~~ rational function with n simple poles the eigenvalues of $\varepsilon^* A$

$$(\lambda \varepsilon - A) x_n + \tilde{y}(\lambda) e_{n+1} = y$$

Go back to $u^\lambda = \begin{pmatrix} u_1^\lambda \\ \vdots \\ u_{n+1}^\lambda \end{pmatrix}$ $(\lambda - A_{n+1}) u^\lambda = \begin{pmatrix} 0 \\ \vdots \\ 0 \\ a_{n+1} u_{n+2}^\lambda \end{pmatrix}$



$$\begin{aligned}
 (\lambda - \bar{\mu}) (u^\mu, u^\lambda) &= (u^\mu, \lambda u^\lambda) - (\mu u^\mu, u^\lambda) \\
 &= (u^\mu, (\lambda - \tilde{A}) u^\lambda) - ((\mu - \tilde{A}) u^\mu, u^\lambda)
 \end{aligned}$$

$$\begin{aligned}
 &= a_{n+1} \bar{\mu} \begin{vmatrix} u_{n+1}^\mu & u_{n+2}^\lambda \\ u_{n+1}^{\bar{\mu}} & u_{n+1}^\lambda \end{vmatrix} - a_{n+1} \begin{vmatrix} u_{n+2}^\mu & u_{n+1}^\lambda \\ u_{n+1}^\mu & u_{n+2}^\lambda \end{vmatrix} \\
 &= a_{n+1} \begin{vmatrix} u_{n+1}^\mu & u_{n+1}^\lambda \\ u_{n+2}^{\bar{\mu}} & u_{n+2}^\lambda \end{vmatrix} \begin{matrix} (x, \lambda \varepsilon^* u^\lambda) \\ \parallel \\ (x, A^* u^\lambda) \end{matrix}
 \end{aligned}$$

$\begin{pmatrix} u^\lambda \\ \lambda u^\lambda \end{pmatrix} \in W^0 \cap \begin{pmatrix} 1 \\ \lambda \end{pmatrix} \mathbb{C}$ $(\varepsilon x, \lambda u^\lambda) \stackrel{?}{=} (A x, u^\lambda)$

$$\begin{pmatrix} u^\mu \\ \mu u^\mu, (-1 \ 1) \begin{pmatrix} u^\lambda \\ \lambda u^\lambda \end{pmatrix} \end{pmatrix} = (\lambda - \bar{\mu}) (u^\mu, u^\lambda) = a_{n+1} \begin{vmatrix} u_{n+1}^{\bar{\mu}} & u_{n+1}^\lambda \\ u_{n+2}^{\bar{\mu}} & u_{n+2}^\lambda \end{vmatrix}$$

~~splitting the operator~~

basic problem: Given an operator A cyclic vector ξ
~~to embed Y into~~
 $\xi \mapsto v_0^*(\lambda - A)^{-1}\xi$ transforms Y to rational functions. Injective

$$(\lambda - A)^{-1}\xi = (\lambda^{-1} + \lambda^{-2}A + \lambda^{-3}A^2 + \dots)\xi$$

So $\oint \frac{d\lambda}{2\pi i} \frac{f(\lambda)}{(\lambda - A)} \xi = f(A)\xi$

$$\int \frac{d\lambda}{2\pi i} f(\lambda) v_0^*(\lambda - A)^{-1}\xi = v_0^* f(A)\xi$$

Go back to $W = \begin{pmatrix} \mathbb{C} \\ A \end{pmatrix} X \subset \begin{matrix} Y \\ \oplus \\ Y \end{matrix}$ $W^0 \cap \begin{pmatrix} 1 \\ \lambda \end{pmatrix} Y = \mathbb{C} \begin{pmatrix} u^\lambda \\ \lambda u^\lambda \end{pmatrix}$

$$0 = \left(\begin{pmatrix} \varepsilon y \\ Ax \end{pmatrix}, \begin{pmatrix} 0 & 1 \\ -1 & 0 \end{pmatrix} \begin{pmatrix} y \\ \lambda y \end{pmatrix} \right) = (\varepsilon x, \lambda y) - (Ax, y) = (x, (\lambda \varepsilon^* - A^*)y) = 0$$

$y \in \text{Ker}(\lambda \varepsilon^* - A^*)$

suppose you pick a line L in $W^0/W \ni p_1 L \notin \varepsilon X$.
 This gives an extension of the partial operator.

Example $L = \begin{pmatrix} u^{\lambda_0} \\ \lambda_0 u^{\lambda_0} \end{pmatrix}$. What is the spectrum of the resulting operator?

~~you seem to have a line~~
 should be related to the ~~resulting~~ λ such that $\begin{pmatrix} u^\lambda \\ \lambda u^\lambda \end{pmatrix}$ maps to L in W^0/W .

Have $W = \begin{pmatrix} \mathbb{C} \\ A \end{pmatrix} X \subset \begin{matrix} Y \\ \oplus \\ Y \end{matrix}$ $W^0 \cap \begin{pmatrix} 1 \\ \lambda \end{pmatrix} Y = \mathbb{C} \begin{pmatrix} u^\lambda \\ \lambda u^\lambda \end{pmatrix}$

$\mathbb{C} u^\lambda = \text{Ker}(\lambda \varepsilon^* - A^*)$. A line L in W^0/W , ~~is~~
~~independent~~ "independent of ε " corresponds to an extension of $A\varepsilon^{-1}$ to an operator on Y which has eigenvalues,

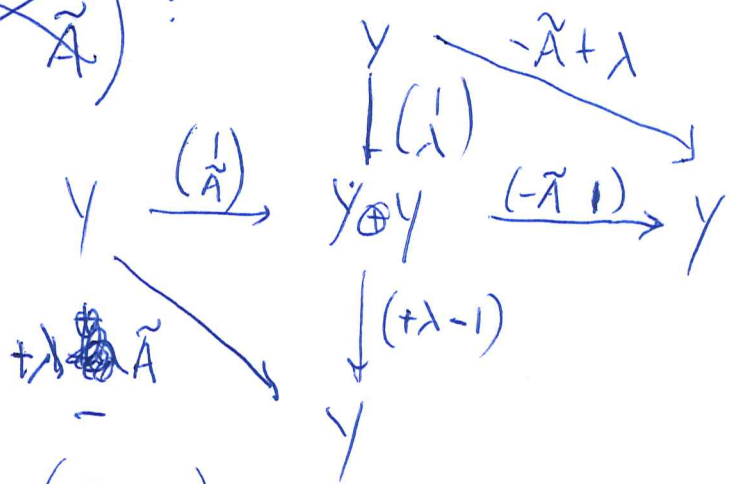
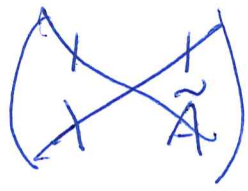
so you should get a pencil of ^{positive} λ divisors of degree $n+1$. ~~Another description~~ You have to distinguish between the λ ~~line~~ P' and $P'(W^0/W)$.

There is a rational map $\lambda \mapsto \ker(\lambda \varepsilon^* - A^*) \simeq W^0 \cap \binom{1}{\lambda} Y$ which sends real axis to isotropic lines in W^0/W degree?

Philosophy: A line in W^0/W is a kind of boundary condition to be added in order to obtain a well defined operator \tilde{A} having a spectrum, ~~etc~~ resolvent, etc. When W is enlarged to the graph of this operator then the resolvent is ~~etc~~ linked to the way $\sqrt{\tilde{A}}$ and $\binom{1}{\lambda} Y$ intersect

$$\binom{1}{\tilde{A}} Y \cap \binom{1}{\lambda} Y = \left\{ \binom{y}{\lambda y} \mid \lambda y = \tilde{A} y \right\}$$

To invert



$$\begin{pmatrix} 1 & 1 \\ \tilde{A} & \lambda \end{pmatrix}^{-1} = \begin{pmatrix} \lambda - 1 & 1 \\ -\tilde{A} & 1 \end{pmatrix} \frac{1}{\lambda - \tilde{A}}$$

What are the natural questions?? You really are missing ~~the~~ the appropriate viewpoint. First question is how ~~to~~ to describe the spectrum. For each λ you have this line $W^0 \cap \binom{1}{\lambda} Y$ mapping to W^0/W and the line $L \hookrightarrow W^0/W$. So the spectrum is described

64 as those λ s.t. this is not transverse i.e.

if ~~the~~ $W'/W = L$, then $W^0 \cap (\frac{1}{\lambda}) \rightarrow W/W$ vanishes. So we have a dual section of the sub line bundle

$$0 \rightarrow \text{Ker}(\lambda \varepsilon^* - A^*) \xrightarrow{\text{degree } -n} Y \xrightarrow{\lambda \varepsilon^* - A^*} X \rightarrow 0$$

So the spectrum is a divisor of degree n .

approach $W = \begin{pmatrix} \varepsilon \\ A \end{pmatrix} X \subset \begin{matrix} Y \\ \oplus \\ Y \end{matrix}$ lines in W^0/W

except for the line $\begin{pmatrix} 0 \\ e_{n+1} \end{pmatrix} \in \mathbb{C} e_{n+1} = (\varepsilon X)^\perp = \text{Ker}(\varepsilon^*)$

the same as are the ~~graphs~~ extensions of $\begin{pmatrix} \varepsilon \\ A \end{pmatrix} X$ to $\begin{pmatrix} 1 \\ \tilde{A} \end{pmatrix} X$ where $\tilde{A}: Y \rightarrow Y$ equiv. $\varepsilon^* \tilde{A} = A^*$, $\tilde{A} \varepsilon = A$. in terms of $\begin{matrix} \forall x' \in X \\ y \in Y \end{matrix}$

J-matrix

$$\tilde{A} = \begin{bmatrix} b_1 & a_1 & & 0 \\ & \backslash & & \\ a_1 & & a_{n+1} & \\ & & \backslash & \\ 0 & a_{n+1} & b_n & a_n \\ & & & \backslash \\ & & & a_n & * \end{bmatrix}$$

* arbitrary

so extensions are described by $b_{n+1} \in \mathbb{C}$, hermitian $\Leftrightarrow b_{n+1}$ real.

the interesting point is to represent lines in W^0/W in the form $W^0 \cap (\frac{1}{\mu}) \simeq \text{Ker}(\mu \varepsilon^* - A^*)$.

What this means is you will choose b_{n+1} to be the coefficient arising from $\begin{pmatrix} u^\mu \\ \mu u^\mu \end{pmatrix}$. This

means $\tilde{A} u^\mu = \mu u^\mu$

$e_{n+1}^* (\tilde{A} u^\mu) = a_n u_n^\mu + b_{n+1} u_{n+1}^\mu$ and this is to be μu_n^μ
 $: a_n u_n^\mu + (b_{n+1} - \mu) u_{n+1}^\mu = 0$

65 so we fix the boundary condition so that u^i is an eigenfunction vector $\tilde{A}u^i = i u^i$.

~~At the same time we have this~~

Now that we have this \tilde{A} which is nearly hermitian you will get a spectral representation once you know $\tilde{A} - \tilde{A}^*$ which is essentially the imaginary part of $b_{n+1} = i - \frac{a_n u_n^i}{u_{n+1}^i}$.

Back to refl positivity. Try to understand the simple harmonic oscillator.

$\langle 0 | x(t_1) x(t_2) \dots x(t_n) | 0 \rangle$ somehow time ordered.

$x(t) = e^{-\frac{i}{\hbar} H t} x e^{+\frac{i}{\hbar} H t}$

Basic example: Forced simple harmonic oscillator $m\ddot{x} + kx = F(t) \in C_0^\infty(\mathbb{R})$

Review: $W = \begin{pmatrix} \mathbb{C} \\ A \end{pmatrix} X \subset \begin{matrix} Y \\ \oplus \\ Y \end{matrix}$ $W^\circ \cap \begin{pmatrix} 1 \\ \lambda \end{pmatrix} Y = \left\{ \begin{pmatrix} y \\ Hy \end{pmatrix} \mid \begin{pmatrix} \lambda \varepsilon^* - A^* \end{pmatrix} y = 0 \right\}$

W°/W is 2 diml. $W^\circ = \left\{ \begin{pmatrix} y_1 \\ y_2 \end{pmatrix} \mid \begin{pmatrix} y_1 \\ Ax \end{pmatrix} = \begin{pmatrix} y_2 \\ \varepsilon x \end{pmatrix} \forall x \right\}$

e.g. $y_1 = \varepsilon x, y_2 = Ax$

i.e. $A^* y_1 = \varepsilon^* y_2$

Suppose e_{n+1} is a unit vector gen. $\text{Ker}(\varepsilon^*) = (\varepsilon X)^\perp$.

~~then~~ $W^\circ \supset W + \begin{pmatrix} \text{Ker } A^* \\ 0 \end{pmatrix} + \begin{pmatrix} 0 \\ \text{Ker } \varepsilon^* \end{pmatrix}$

Can you have $\begin{pmatrix} \varepsilon x \\ Ax \end{pmatrix} \in \begin{matrix} \text{Ker } A^* \\ \oplus \\ \text{Ker } \varepsilon^* \end{matrix}$ Yes.
 i.e. $\varepsilon^* Ax = 0$ & $A^* \varepsilon x = 0$

Any line L in W°/W corresp to a $(n+1)$ -subsp. V of Y ~~containing~~ W , \mathcal{L}

Look at $p_1: V \rightarrow Y$ either $p_1 V \subset \varepsilon X$ whence $\text{Ker}(p_1|_V)$ is a line in $\begin{matrix} \oplus \\ Y \end{matrix}$ $\text{cont } 0$ in W° i.e. $(\varepsilon X)^\perp$.

or $p_1: V \rightarrow Y$ and then V is graph of $\tilde{A}: Y \rightarrow Y$
 must have $\Gamma_{\tilde{A}} \subset W^0$ i.e. $(y, Ax) = (\tilde{A}y, \varepsilon x) \quad \forall x, y$
 i.e. $A = \tilde{A}^* \varepsilon$ and $W \subset \Gamma_{\tilde{A}}$ i.e. $\tilde{A} \varepsilon = A$.
 $A^* = \varepsilon^* \tilde{A}$

about \tilde{A} is determined by $e_{n+1}^* \tilde{A} e_{n+1} = b_{n+1} \in \mathbb{C}$

But you want the de Branges picture, which is based on a specific choice for \tilde{A} , namely she uses the line $\text{Im}g \left\{ W^0 \cap \binom{1}{\mu} Y \rightarrow W^0/W \right\}$ where $\mu = i$.

so $\Gamma_{\tilde{A}} = W \oplus \mathbb{C} \begin{pmatrix} u^i \\ iu^i \end{pmatrix} \quad \tilde{A} u^i = i u^i$

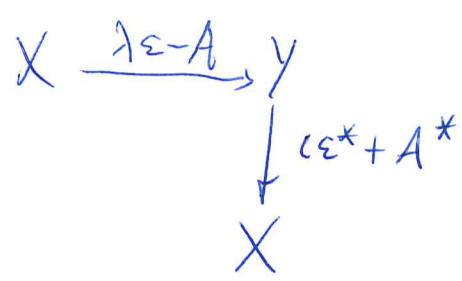
which means $\det(\lambda - \tilde{A})$ has the factor $\lambda - i$.

Now u^μ killed $\mu \varepsilon^* - A^*$ in general,

~~u^i is general~~ You want isometric embedding
 I ~~think~~ think I want to use $e_{n+1}^* (\lambda - \tilde{A})^{-1} y$ for the isometric embedding. Need to know about $\tilde{A} - \tilde{A}^*$.

$\varepsilon^* (\tilde{A} - \tilde{A}^*) = A^* - (\tilde{A} \varepsilon)^* = A^* - A^* = 0$, also
 $(\tilde{A} - \tilde{A}^*) \varepsilon = A - A = 0$.

Can you relate \tilde{A} to the stuff before



$$\begin{array}{ll}
 \varepsilon = \frac{1}{2}(a+b) & c\varepsilon - A = ib \\
 A = \frac{i}{2}(a-b) & -i\varepsilon^* - A^* = -ib^* \\
 & c\varepsilon^* + A^* = ib^*
 \end{array}$$

$$\begin{aligned}
 (c\varepsilon^* + A^*) \varepsilon &= cb^* \frac{1}{2}(a+b) = \frac{i}{2}(1 + b^*a) \\
 (c\varepsilon^* + A^*) A &= cb^* \frac{i}{2}(a-b) = \frac{1}{2}(1 - b^*a) \\
 -i(c\varepsilon^* + A^*) \varepsilon &= \frac{1}{2}(1 + b^*a) \\
 \varepsilon^* \varepsilon + A^* A &= 1
 \end{aligned}$$

67 Start with $d\mu$ on \mathbb{R} prob. measure, when scalar product on $\mathcal{O}[\lambda]$. Restrict to $Y = F_{n+1}$ polys of degree $\leq n$. Get $(\begin{smallmatrix} \varepsilon \\ A \end{smallmatrix}) F_n$. ~~Reproducing kernel?~~ What can you say about point evaluation.

$$y(\alpha) = (e_\alpha, y) \quad \text{in } Y = F_{n+1}.$$

$$(e_\alpha, (A - \alpha)x) = 0 \quad \forall x$$

$$\text{so } ((A^* - \bar{\alpha})e_\alpha, x) = 0 \quad \forall x \in F_n \quad \text{so}$$

Confused. $Y = F_{n+1} = \mathbb{C}1 + \mathbb{C}\lambda + \dots + \mathbb{C}\lambda^n$
 $X = F_n = \mathbb{C}1 + \dots + \mathbb{C}\lambda^{n-1}$

$$y(\alpha) = (e_\alpha, y) \quad \text{defines } e_\alpha \in Y \text{ for } \alpha \in \mathbb{C}.$$

$$\text{Then } (e_\alpha, (\lambda - \alpha)x) = 0 \quad \text{so } e_\alpha \perp (\lambda\varepsilon - A)x$$

$$\text{so } (\bar{\alpha}\varepsilon^* - A^*)e_\alpha = 0$$

~~Together measure~~ ~~Take~~ Let p_1, \dots, p_{n+1} be the orthogonal polys. ~~Then~~

$$e_\alpha = \sum_i \overline{p_i(\alpha)} p_i \quad e(\alpha, \lambda)$$

$$\text{so that } (e_\alpha, y) = \sum_i p_i(\alpha) (p_i, y)$$

$$(e_\alpha, p_j) = \sum_i p_i(\alpha) \delta_{ij} = p_j(\alpha).$$

$$\int e(\alpha, \lambda) y \, d\mu(\lambda) = y(\alpha)$$

$$\int e(x'', x') \, d\mu(x') \int e(x', x) y(x) \, d\mu(x) = \int y(x')$$

$$\int d\mu(x') e(x'', x') e(x', x) = e(x'', x)$$

58 Review. $W = \begin{pmatrix} \Sigma \\ A \end{pmatrix} X \subset W^0 = \left\{ \begin{pmatrix} y_1 \\ y_2 \end{pmatrix} \mid (y_1, Ax) = (y_2, \varepsilon x) \quad \forall x \right\}$

~~Consider a line~~ Consider a line V/W in W^0/W where $W \subset V \subset W^0$. Assembling $V \cap \underbrace{\text{Ker}(p_1: W^0 \rightarrow Y)}_{\text{Ker } \varepsilon^*} = 0$

then $p_1: V \xrightarrow{\sim} Y$ so $V = \begin{pmatrix} 1 \\ \tilde{A} \end{pmatrix} X$ where $\tilde{A}\varepsilon = A$ and $(y, Ax) = (\tilde{A}y, \varepsilon x) \quad \forall x, y$

equiv. $\varepsilon^* \tilde{A} = A^*$. ~~And this~~ $\therefore \tilde{A}^* \varepsilon = A^*$, so $(\tilde{A}^* - A^*) \varepsilon = 0 \Rightarrow \varepsilon^* (\tilde{A}^* - A^*) = 0$. So $\tilde{A}^* - A^* = e_{n+1} b_{n+1} e_{n+1}^*$ e_{n+1} unit v. gen. $\text{Ker}(\varepsilon^*)$.

$$W^0 \cap \begin{pmatrix} 1 \\ \lambda \end{pmatrix} Y = \left\{ \begin{pmatrix} y \\ \lambda y \end{pmatrix} \mid (y, Ax) = (\lambda y, \varepsilon x) \quad \forall x \right\}$$

i.e. $(\lambda \varepsilon^* - A^*) y = 0$

$$W^0 \cap \begin{pmatrix} 1 \\ \mu \end{pmatrix} Y = \begin{pmatrix} 1 \\ \mu \end{pmatrix} \text{Ker}(\mu \varepsilon^* - A^*) = \begin{pmatrix} u^\mu \\ \mu u^\mu \end{pmatrix}$$

What would you like to do? Couple to a transmission line. Look at Hardy space

$H = L^2(\mathbb{R}, \frac{d\omega}{2\pi}) = H^+ \oplus H^-$ You would like to make $H^- \oplus Y \oplus H^+$ a self adjoint op. combining $A\varepsilon^{-1}$ with mult by ω on H^+ . Work with subspaces of $H^- \oplus Y \oplus H^+$. What happens with H^+

$$W^+ = \left\{ \begin{pmatrix} f \\ \omega f \end{pmatrix} \mid \int_{f \in H^+} (1+\omega^2) |f|^2 \frac{d\omega}{2\pi} < \infty \right\}$$

Keep on trying.

$$(W^+)^0 = \left\{ \begin{pmatrix} f_1 \\ f_2 \end{pmatrix} \mid (f_1, \omega f) = (f_2, f) \quad \forall f \in D_\omega^+ \right\}$$

$$\varepsilon = \frac{a+b}{2}$$

$$A = \frac{i(a-b)}{2}$$

$$W^+ = \begin{pmatrix} i \\ \omega \end{pmatrix} D_\omega^+ \subset \begin{pmatrix} H^+ \\ H^+ \end{pmatrix} \xrightarrow{\frac{1}{\sqrt{2}} \begin{pmatrix} 1 & -i \\ 1 & i \end{pmatrix}} \begin{pmatrix} H^+ \\ \oplus H^+ \end{pmatrix}$$

$$\begin{pmatrix} \varepsilon \\ A \end{pmatrix} = \frac{1}{2} \begin{pmatrix} 1 & 1 \\ i & -i \end{pmatrix} \begin{pmatrix} a \\ b \end{pmatrix}$$

$$\begin{pmatrix} 1 & -i \\ 1 & i \end{pmatrix} \begin{pmatrix} 1 \\ \omega \end{pmatrix} = \frac{1}{1+\omega} \begin{pmatrix} 1-\omega \\ 1+\omega \end{pmatrix} \begin{pmatrix} a \\ b \end{pmatrix} = i \begin{pmatrix} -i-1 \\ -i & 1 \end{pmatrix} \begin{pmatrix} \varepsilon \\ A \end{pmatrix}$$

69 $\begin{cases} a = \varepsilon + iA \\ b = \varepsilon + iA \end{cases} \quad \begin{matrix} 1-i\omega \\ 1+i\omega \end{matrix}$

$$\begin{pmatrix} 1 & -i \\ 1 & +i \end{pmatrix} \begin{pmatrix} 1 \\ \omega \end{pmatrix} D_\omega = \begin{pmatrix} (1-i)D_\omega \\ (1+i)D_\omega \end{pmatrix} = \begin{pmatrix} (1-i\omega)D_\omega \\ (1+i\omega)D_\omega \end{pmatrix}$$

column 1 $(1-i\omega)D_\omega = H^+$ since $1-i\omega=0$ when $\omega=-i$
 $(1+i\omega)D_\omega = (\omega-i)D_\omega$

kernel of valuation at $\omega=i$. so $(W^+)^{\oplus}$ has

a
 Review. Take $\begin{pmatrix} z \\ A \end{pmatrix} X \subset \begin{matrix} Y \\ \oplus \\ Y \end{matrix}$ and $\begin{pmatrix} 1 \\ \omega \end{pmatrix} D_\omega \subset \begin{matrix} H^+ \\ \oplus \\ H^+ \end{matrix}$

$$\begin{pmatrix} 1 \\ \omega \end{pmatrix} D_\omega = \begin{pmatrix} (1+\omega^2)^{-1/2} \\ \omega(1+\omega^2)^{-1/2} \end{pmatrix} H^+ \subset \begin{matrix} H^+ \\ \oplus \\ H^+ \end{matrix}$$

$$W = \begin{pmatrix} a \\ b \end{pmatrix} X \subset \begin{matrix} Y \\ \oplus \\ Y \end{matrix} \quad W^0 = W \oplus \begin{matrix} \text{Ker}(a^*) \\ \oplus \\ \text{Ker}(b^*) \end{matrix}$$

~~do~~ do harmonic oscillator

$$x'' + x = F(t) \quad \text{has solution}$$

$$x = \int_{-\infty}^t G(t-t') F(t') dt'$$

$$(\partial_t^2 + 1)G(t) = \delta(t)$$

vanishing as $t \rightarrow -\infty$

$$G(t) = \begin{cases} \sin t & t \geq 0 \\ 0 & t \leq 0 \end{cases}$$

and general solution

$$x = \text{Re}(Ae^{-it}) + \int_{-\infty}^t \sin(t-t') F(t') dt'$$

~~which is $\int_{-\infty}^t \sin(t-t') F(t') dt'$~~

What happens as $t \rightarrow \infty$

$$\begin{aligned} \int_{-\infty}^t \sin(t-t') F(t') dt' &= \int \text{Re}(-ie^{i(t-t')}) F(t') dt' \\ &= \int_0^\infty \text{Re}(-ie^{iu}) F(t-u) du \end{aligned} \quad ?$$

70 for $t \gg 0$ dryer stop 11:00, then 80 min

$$x(t) = \int_{-\infty}^{\infty} \text{Re}(-ie^{i(t-t')}) F(t') dt'$$

$$= \text{Re} \int_{-\infty}^{\infty} -ie^{it} e^{-it'} F(t') dt'$$

$$H = \omega a^* a$$

$$H = \frac{1}{2m} p^2 + \frac{1}{2} k g^2$$

$$[p, g] = \left[\frac{\hbar}{i} \partial_x, x \right]$$

$$(\omega g - ip)(\omega g + ip)$$

$$\cancel{[g, p]} = \frac{\hbar}{i}$$

$$[g, p] = i\hbar$$

$$[\omega g - ip, \omega g + ip] = \omega \hbar i^2 - i\omega \frac{\hbar}{i} = -2\omega \hbar$$

$$[a, a^*] = \left[\frac{\omega g + ip}{\sqrt{2\hbar\omega}}, \frac{\omega g - ip}{\sqrt{2\hbar\omega}} \right] = 1$$

$$\omega \frac{(\omega g - ip)(\omega g + ip)}{2\hbar\omega}$$

$$H = \frac{p^2}{2m} + \frac{1}{2} k g^2$$

$$\dot{g} = \frac{\partial H}{\partial p} = \frac{p}{m} \quad \dot{p} = -\frac{\partial H}{\partial g} = -k g$$

$$m\ddot{g} + k g = 0 \quad \left(\frac{k}{m} = \omega^2 \right)$$

$$H = \frac{p^2}{2m} + \frac{m}{2} \omega^2 g^2$$

$$= \hbar\omega \left(\frac{-ip}{\sqrt{2m\omega\hbar}} + \sqrt{\frac{m\omega}{2\hbar}} g \right) \left(\frac{ip}{\sqrt{2m\omega\hbar}} + \sqrt{\frac{2m\omega}{2\hbar}} g \right) = \hbar\omega \left(H - \frac{1}{2} \right)$$

not important

$$\left[\frac{ip}{\sqrt{2m\omega\hbar}}, \sqrt{\frac{2m\omega}{2\hbar}} g \right] = \frac{1}{2\hbar} \hbar = \frac{1}{2}$$

suppose $H = \omega a^* a$ $g = a + a^*$

$$\langle 0 | g e^{-itH} g | 0 \rangle = \langle 0 | a e^{-it\omega a^* a} a^* | 0 \rangle$$

$$t = -i\tau$$

$$e^{-i(-i\tau)H} = e^{-\tau H}$$

$$= \langle 0 | a e^{-it\omega} a^* | 0 \rangle = e^{-it\omega}$$

Review: $W = \begin{pmatrix} a \\ b \end{pmatrix} X \subset \bigoplus_y$ $L^2(\mathbb{R}, \frac{d\omega}{2\pi}) = H^- \oplus H^+$ Hardy spaces

To understand

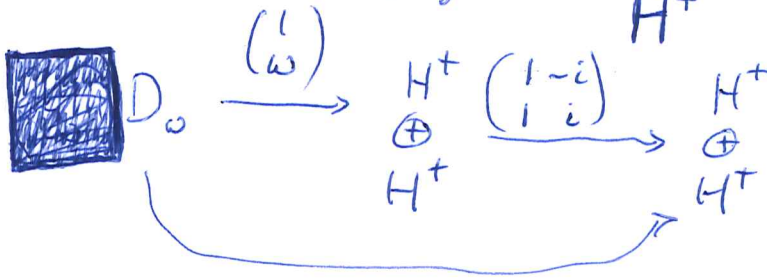
$$\begin{pmatrix} 1 \\ \omega \end{pmatrix} D_\omega \subset \begin{matrix} H^+ \\ \oplus \\ H^+ \end{matrix}$$

$$\varepsilon = \frac{1}{2}(a+b)$$

$$A = \frac{1}{2}(a-b)$$

$$\varepsilon - iA = a$$

$$\varepsilon + iA = b$$



$$a = 1 - i\omega$$

$$b = 1 + i\omega$$

$$ba^{-1} = \frac{1+i\omega}{1-i\omega}$$

Idea is that $1+i\omega$ vanishes at $\omega = i$ so that $(1+i\omega)D_\omega$ should ~~be~~ have codim 1.

You need $L^2(S^1)$ $L^2(\mathbb{R})$

$$z \quad \frac{1+i\omega}{1-i\omega} = \frac{-\omega+i}{\omega+i}$$

$$1 \quad \int_{-\infty}^{\infty} \left| \frac{\sqrt{2}}{\omega+i} \right|^2 \frac{d\omega}{2\pi} = \int \frac{2}{1+\omega^2} \frac{d\omega}{2\pi}$$

$$= \frac{2}{2\pi} \arctan \omega \Big|_{-\infty}^{\infty} = \frac{\pi}{\pi} = 1$$

Go back to ~~Chapman Algebra~~

$$H = H^- \oplus Y \oplus H^+$$

Let's return to $H = \dots \oplus u^1 V^- \oplus aX \oplus V^+ \oplus uV^+ \oplus \dots$

$$\begin{matrix} \searrow & & \searrow & & \searrow \\ \dots \oplus u^1 V^- \oplus & V^- \oplus bX \oplus & uV^+ \oplus \dots \end{matrix}$$

and try for the hermitian analogues. So what to do next? Is there a simple way to describe H ?

Suppose you try to generalize $f^* u^n f = (f^* u f)^n$ for $n \geq 0$ to the continuous case.

72 ~~Look~~ Look for H with $u^t = e^{it} h$ h herm.

$$f^* u^t f = (f^* u f)^t \quad \text{for } t \geq 0. \quad \text{Meaning?}$$

~~Maybe~~ Maybe it works:

?

discrete case: $f^* u^n f = \begin{cases} (f^* u f)^n & n \geq 0 \\ (f^*)^{-n} & n \leq 0. \end{cases}$

form $\sum_{n \in \mathbb{Z}} \bar{z}^n f^* u^n f = \sum_{n \geq 0} (\bar{z}^{-1} \gamma)^n + \sum_{n < 0} (\bar{z} \gamma^*)^n$

$$= \frac{1}{1 - \bar{z}^{-1} \gamma} + \frac{\bar{z} \gamma^*}{1 - \bar{z} \gamma^*}$$

$$= \frac{1}{1 - \bar{z}^{-1} \gamma} \underbrace{(1 - \bar{z} \gamma^* + (1 - \bar{z} \gamma^*) \bar{z} \gamma^*)}_{1 - \gamma \gamma^*} \frac{1}{1 - \bar{z} \gamma^*}$$

analogue is

$$\int_{-\infty}^{\infty} e^{-i\omega t} \underbrace{f^* u^t f}_{\begin{cases} \beta^t & t \geq 0 \\ (\beta^*)^{-t} & t \leq 0 \end{cases}} dt$$

$$\int_0^{\infty} e^{-i\omega t} e^{t\beta} dt + \int_{-\infty}^0 e^{-i\omega t} e^{-t\beta^*} dt$$

$$\left[\frac{e^{(-i\omega + \beta)t}}{-i\omega + \beta} \right]_0^{\infty} + \left[\frac{e^{-(i\omega + \beta^*)t}}{-(i\omega + \beta^*)} \right]_{-\infty}^0$$

$$\frac{1}{i\omega - \beta} - \frac{1}{i\omega + \beta^*} = \frac{1}{i} \left(\frac{1}{\omega + i\beta} - \frac{1}{\omega - i\beta^*} \right)$$

$$= \frac{1}{\omega - i\beta^*} \underbrace{\left(\frac{\omega - i\beta^* - (\omega + i\beta)}{i} \right)}_{-(\beta + \beta^*)} \frac{1}{\omega + i\beta} = \frac{1}{\omega - \alpha^*} (-i(\alpha - \alpha^*)) \frac{1}{\omega - \alpha}$$

Put $\beta = i\alpha$

73 Can you use this somehow. The idea is to produce a nearly hermitian operator directly from $\begin{pmatrix} B \\ A \end{pmatrix}$

We expect to find a self adjoint operator H such that e^{itH} such that $f^* e^{itH} f = e^{it\alpha}$

so $f^* \frac{1}{\omega - H} f = \frac{1}{\omega - \alpha}$ $\text{Im}(\alpha) \geq 0$
 to what?

$W = \begin{pmatrix} \varepsilon \\ A \end{pmatrix} X = \begin{matrix} Y \\ \oplus \\ Y \end{matrix}$ $W^0 = \left\{ \begin{pmatrix} y_1 \\ y_2 \end{pmatrix} \mid (y_1, Ax) = (y_2, x) \quad \forall x \right\}$
 $W^0_n \left(\begin{matrix} 1 \\ \lambda \end{matrix} \right) Y = \left\{ \begin{pmatrix} y \\ \lambda y \end{pmatrix} \mid y \in \text{Ker}(\lambda \varepsilon^* - A^*) \right\}$

partial unitary picture

$W = \begin{pmatrix} a \\ b \end{pmatrix} X = \begin{matrix} Y \\ \oplus \\ Y \end{matrix}$ $W^0 = \left\{ \begin{pmatrix} y_1 \\ y_2 \end{pmatrix} \mid (y_1, ax) = (y_2, bx) \quad \forall x \right\}$
 i.e. $a^* y_1 = b^* y_2$

given $\begin{pmatrix} y_1 \\ y_2 \end{pmatrix} \in W^0$ let $x = ~~a^* y_1~~ a^* y_1$. Then

$\begin{pmatrix} y_1 \\ y_2 \end{pmatrix} - \begin{pmatrix} a a^* y_1 \\ b a^* y_1 \end{pmatrix} = \begin{pmatrix} y_1' \\ y_2' \end{pmatrix}$ where $a^* y_1' = 0$
 $b^* y_2' = 0$

$\therefore W^0 = W \oplus \begin{matrix} \text{Ker } a^* \\ \oplus \\ \text{Ker } b^* \end{matrix}$

$L_z = W^0_n \left(\begin{matrix} 1 \\ z \end{matrix} \right) Y = \left\{ \begin{pmatrix} y \\ zy \end{pmatrix} \mid a^* y = z b^* y \right\}$
 $y \in \text{Ker}(\begin{matrix} a^* - z b^* \end{matrix})$

~~$L_z = \dots$~~

$$\begin{array}{ccccccc} 0 & \rightarrow & L_z & \rightarrow & Y & \xrightarrow{a^* - z b^*} & X & \rightarrow & 0 \\ & & \downarrow & & \downarrow \begin{pmatrix} 1 \\ z \end{pmatrix} & & \parallel & & \\ 0 & \rightarrow & W^0 & \rightarrow & Y \oplus Y & \rightarrow & X & \rightarrow & 0 \\ & & \downarrow & & \downarrow & & & & \\ & & Y & = & Y & & & & \end{array}$$

74 Review again. Consider $W = \begin{pmatrix} a \\ b \end{pmatrix} X \subset Y$ isotropic $\|ax\|^2 = \|bx\|^2$
 for $\|y_1\|^2 - \|y_2\|^2$. Find $W^\circ = W \oplus \begin{matrix} \text{Ker}(a^*) \\ \text{Ker}(b^*) \end{matrix}$. ~~What is~~

~~the strategy~~ Now pick a line in W°/W . The one you take is $\begin{pmatrix} 0 \\ \text{Ker}(b^*) \end{pmatrix}$. Note that any line in W°/W corresponds to a V , $W \subset V \subset W^\circ$. ~~Is V the graph of~~ When is V ~~the~~ graph. Look at $p_i: V \rightarrow Y$

Wait. You know that $W^\circ \ominus W = \begin{matrix} \text{Ker } a^* \\ \oplus \\ \text{Ker } b^* \end{matrix}$, so ~~the~~ $V \ominus W$ is a line $\subset \begin{matrix} \text{Ker } a^* \\ \oplus \\ \text{Ker } b^* \end{matrix}$. ~~So the impact~~

~~You know that~~

Let's review the situation in the ^(partial) unitary $O(n)$ case. You have Y a $n+1$ dim Hilb space, an n dim v.s. X and maps $a, b: X \rightarrow Y \ni 1) a^*b = b^*a$ all $z \in \mathbb{C} \cup \infty$
 2) $\|ax\| = \|bx\|$ all x . \therefore get $\|\cdot\|$ on $X \ni a^*a = b^*b = 1$.

~~From this you get~~ Form

$$H_{\text{int}} = \underbrace{aX \oplus V^+ \oplus uV^+}_{\overline{V^- \oplus bX \oplus}}$$

$$y = aa^*y + \pi^+y$$

$$uy = ba^*y + \pi^+y = aa^*ba^*y + \pi^+ba^*y + u\pi^+y$$

$$u^2y = aa^*(ba^*)^2y + \pi^+(ba^*)^2y + u\pi^+(ba^*)y + u^2\pi^+y$$

$$y = u^{-N} \left\{ aa^*(ba^*)^N y + u^N \pi^+(ba^*)^N y + \dots \right.$$

so move into functions

$$y \rightsquigarrow \begin{matrix} \pi^+y + u^{-1}\pi^+(ba^*)y + u^{-2}\pi^+(ba^*)^2y + \dots \\ \pi^+(1 - z^{-1}ba^*)^{-1}y \end{matrix}$$

75 $Y \xrightarrow{\begin{pmatrix} za^* \\ e^* \end{pmatrix}} X \xrightarrow{\begin{pmatrix} az-b & e \end{pmatrix}} Y$ is solution of $(az-b)x + e e^t = y$
 $\oplus \xrightarrow{V^+}$
 $\begin{pmatrix} za^* \\ e^* \end{pmatrix} (1 - z^{-1} b a^*)^{-1} y$

$(az-b)z^{-1}a^* + ee^* = 1 - z^{-1}ba^*$

How do you use, organize, these ideas?

~~You want to deform~~

You have $Y \xrightarrow{\begin{pmatrix} -b^* \\ e^* \end{pmatrix}} X \xrightarrow{\begin{pmatrix} az-b & e \end{pmatrix}} Y$
 $\oplus \xrightarrow{V^-}$

How to organize? You might work with \oplus
 Y

$W = \begin{pmatrix} a \\ b \end{pmatrix} X$, $\begin{pmatrix} 1 \\ z \end{pmatrix} Y$, $W^0 = W \oplus \begin{pmatrix} V^+ \\ \oplus \\ V^- \end{pmatrix}$

Your previous success is based upon the splitting $Y = \begin{pmatrix} X \\ \oplus \\ V^+ \end{pmatrix}$ or $Y = \begin{pmatrix} X \\ \oplus \\ V^- \end{pmatrix}$. You

somehow use the

Start again. You have $X \xrightarrow{\begin{pmatrix} a \\ b \end{pmatrix}} Y$ ~~$a^*a = b^*b = 1$~~
 $\|ax\|^2 = \|bx\|^2 \quad \forall x \iff W = \begin{pmatrix} a \\ b \end{pmatrix} X$ isot. form $\|y_1\|^2 = \|y_2\|^2$.

$W^0 = W \oplus \begin{pmatrix} \text{Ker}(a^*) \\ \oplus \\ \text{Ker}(b^*) \end{pmatrix}$. The basic spectral representation arises from the splitting $Y \xrightarrow{\begin{pmatrix} +b^* \\ e^* \end{pmatrix}} X \xrightarrow{\begin{pmatrix} az+tb & e_- \end{pmatrix}} Y$
 $\oplus \xrightarrow{V^-}$
 $Y \xrightarrow{\begin{pmatrix} b^* \\ e^* \end{pmatrix}} X \xrightarrow{\begin{pmatrix} b & e \end{pmatrix}} Y$. Leads to solution

of $(az-b)x = -y + \tilde{y}(z)e$ is $\begin{pmatrix} x \\ \tilde{y}(z) \end{pmatrix} = \begin{pmatrix} b^* \\ e^* \end{pmatrix} (1 - zab^*)^{-1} y$

But then you ~~can~~ can prove that $\int |\tilde{y}(z)|^2 \frac{d\theta}{2\pi} = \|y\|^2$
 It's this residue trick you need to understand

76 better. YES. How might you ~~show~~ prove it. First you might take 2 elements $y, y' \in Y$ and somehow ~~prove~~ understand

$$(y', (1 - z'ba^*)^{-1} e e^* (1 - zab^*)^{-1} y) \quad ee^* = 1 - bb^*$$

trick

~~$$(y', (1 - z'ba^*)^{-1} e e^* (1 - zab^*)^{-1} y)$$~~

$$\frac{1}{1 - z'z} + \frac{zz^*}{1 - zz^*} = \frac{1}{1 - z'z} ((1 - z'z) + zz^* + 1 - zz^*) \frac{1}{1 - zz^*}$$

$$\frac{1}{1 - z'z} (1 - zz^*) \frac{1}{1 - zz^*} = \frac{A}{1 - z'z} +$$

Invariant approach - see next 2 pages.

$P' = PT$, where T is 2-dim equipped with pseudoscalar product, Y is a Hilbert space,

$T \otimes Y$ has ^{product} pseudoscalar product, canonical sequence

$$0 \rightarrow \mathcal{O}(-1) \otimes Y \rightarrow \mathcal{O} \otimes T \otimes Y \rightarrow \mathcal{O}(1) \otimes Y \rightarrow 0$$

L^2 sections of $\mathcal{O}(-1)$ over the real P' should form a Hilbert space in an intrinsic way, hence also

~~sections~~ L^2 sections of $\mathcal{O}(-1) \otimes Y$. So what else

happens? You now want to ~~go on~~ proceed to spectral representation. You need to choose a

line in W^0/W and ~~adjoint~~ conjugate (or adjoint) line. Corresponds to choose $W < V < W^0$ and its annihilator V^0 . Interested in ~~not~~ $V \neq V^0$

Invariant approach. T 2 dimensional space with hermitian form of signature $1, -1$. Y Hilbert space. $T \otimes Y$ is Klein space. Have basic exact sequence

$$0 \rightarrow \mathcal{O}(-1) \otimes Y \xrightarrow{\partial \otimes} T \otimes Y \rightarrow \mathcal{O}(1) \otimes Y \rightarrow 0$$

~~Take~~ Fake W isotropic in $T \otimes Y$ for the pseudo scalar product, get W^0/W . Where is the K -mod?

Example: $T = \mathbb{C}^2 = \{ |z_1|^2 - |z_2|^2 \}$
 $T \otimes Y = \underbrace{Y \oplus Y}_{W^0} = \{ \|y_1\|^2 - \|y_2\|^2 \}$

$$0 \rightarrow \begin{pmatrix} 1 \\ z \end{pmatrix} Y \hookrightarrow Y \oplus Y \xrightarrow{(z-1)} Y \rightarrow 0$$

So get $\partial \otimes W^0 \rightarrow \mathcal{O}(1) \otimes Y$ which should be

~~OKAY~~ OKAY provided $\partial \otimes W^0$ and $\mathcal{O}(-1) \otimes Y$ intersect transversally, i.e. $W^0 + \begin{pmatrix} 1 \\ z \end{pmatrix} Y = Y$. How is this related to $W \cap \begin{pmatrix} 1 \\ z \end{pmatrix} Y = 0$? There should be some relation between

$$W^0 = \left\{ \begin{pmatrix} y_1 \\ y_2 \end{pmatrix} \mid a^* y_1 = b^* y_2 \right\}$$

$$W^0 \cap \begin{pmatrix} 1 \\ z \end{pmatrix} Y = \left\{ \begin{pmatrix} y \\ zy \end{pmatrix} \mid a^* y = b^* (zy) \right\} \simeq \text{Ker}(a^* - z b^*)$$

$$= (a - \bar{z} b)^\perp$$

~~is~~ \therefore If no bound states

$$W \cap \begin{pmatrix} 1 \\ z \end{pmatrix} Y = \left\{ \begin{pmatrix} ax \\ bx \end{pmatrix} \mid bx = zax \right\} \simeq \text{Ker}(az - b)$$

annihilator relation ship

$$\left(W^0 + \begin{pmatrix} 1 \\ z \end{pmatrix} Y \right)^\circ = W \cap \begin{pmatrix} 1 \\ \bar{z}^{-1} \end{pmatrix} Y$$

So how do you proceed at this point?

We know that W^0/W has induced pseudoscalar product. Suppose it has dim we have $\mathcal{O}(n)$ case

Picking a line

79 be to try to arrange this by choosing the ~~coordinates~~ coordinates. ~~Choose the~~

Suppose given $W \subset T \otimes Y$ isotropic, and $V, W < V < W^0$. Choose a polar. of $T = \begin{pmatrix} \mathbb{C} \\ \mathbb{C} \\ \mathbb{C} \end{pmatrix}$ ps sc pr $k_1^2 - k_2^2$

$W = \begin{pmatrix} a \\ b \end{pmatrix} X$, $\|ax\|^2 = \|bx\|^2$ $W^0 = W \oplus \begin{matrix} \text{Ker } a^* \\ \oplus \\ \text{Ker } b^* \end{matrix} \oplus V$

intersects to give a line $L \subset \begin{matrix} \text{Ker } a^* \\ \oplus \\ \text{Ker } b^* \end{matrix}$. Actually we can restrict the hermitian forms to V , ~~the~~ it vanishes on W and has a sign > 0 or < 0 on $V-W$. In fact ~~there's a map to~~ V/W is a ^{complex} line with scalar product.

If you have $V \oplus \mathbb{C} \omega \otimes Y \xrightarrow{\sim} T \otimes Y$

for ω not in the spectrum, and then your ~~form~~ ^{quotient line V/W} ~~is~~, what equations are ~~you~~ solving?

$$\begin{pmatrix} 1 \\ \gamma \end{pmatrix} Y + \begin{pmatrix} 1 \\ \gamma \end{pmatrix} Y \xrightarrow{\sim} \begin{matrix} Y \\ \oplus \\ Y \end{matrix}$$

So it seems that I get an element of V/W for any ~~choice~~ triple (z, y_1, y_2) $z \notin \text{spec } \gamma$

$$(z-1) : \begin{pmatrix} 1 \\ \gamma \end{pmatrix} Y \xrightarrow{\sim} Y$$

$$z - \gamma : Y \xrightarrow{\sim} Y$$

so given γ

$$\begin{array}{ccc} \mathcal{O} \otimes V & \xrightarrow{\sim} & \mathcal{O}(1) \otimes Y \\ \downarrow & & \\ \mathcal{O} \otimes (V/W) & & \end{array}$$

80 Review. You have $W \subset V \subset W^\circ \subset T \otimes Y$
 and $0 \rightarrow \mathcal{O}(-1) \otimes Y \rightarrow \mathcal{O} \otimes T \otimes Y \rightarrow \mathcal{O}(1) \otimes 1 \rightarrow 0$

The hermitian form on $T \otimes Y$ ~~restricts to~~ restricts to 0 on W so you get a 1-dim quotient V/W with pos. def. herm. form. Spectral transforms, namely go from Y to functions on the real $\mathbb{P}^1 \subset \mathbb{P}(T)$.

Spectrum = where $\begin{pmatrix} 1 \\ z \end{pmatrix} Y$ and V are not complementary. off spectrum. Assume $V = \begin{pmatrix} 1 \\ \gamma \end{pmatrix} Y$

$$\begin{pmatrix} 1 \\ \gamma \end{pmatrix} Y \oplus \begin{pmatrix} 1 \\ z \end{pmatrix} Y = \begin{pmatrix} 1 \\ \gamma \end{pmatrix} Y \oplus \begin{pmatrix} 1 \\ z \end{pmatrix} Y$$

$$\Leftrightarrow (z \ -1) \begin{pmatrix} 1 \\ \gamma \end{pmatrix} = z - \gamma : Y \rightarrow Y \text{ is an isom.}$$

Anyway

$$\begin{pmatrix} 1 & 1 \\ \gamma & z \end{pmatrix}^{-1} = \begin{pmatrix} z & -1 \\ -\gamma & 1 \end{pmatrix} (z - \gamma)^{-1}$$

At the moment you have a map from $\begin{pmatrix} y_1 \\ y_2 \end{pmatrix} \in \begin{pmatrix} Y \\ Y \end{pmatrix}$ to

$$\begin{pmatrix} 1 \\ \gamma \end{pmatrix} (z \ -1) (z - \gamma)^{-1} \begin{pmatrix} y_1 \\ y_2 \end{pmatrix}$$

$$z(z - \gamma)^{-1} y_1 - (z - \gamma)^{-1} y_2 = (z - \gamma)^{-1} (z y_1 - y_2)$$

~~This~~ This is the element of ~~V~~ V which is to be projected onto V/W .

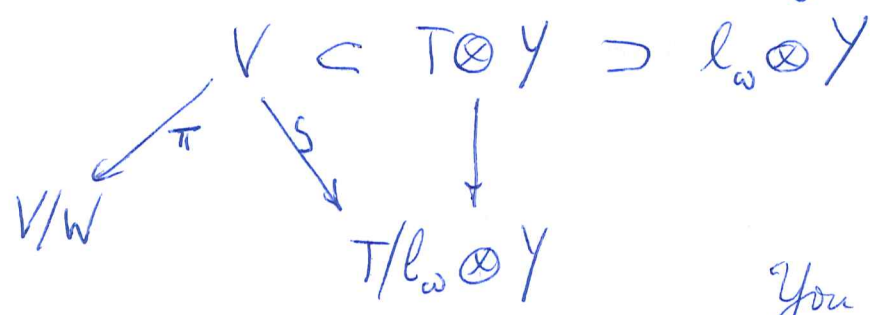
It's worth looking for a proof that $y \mapsto \pi (z - \gamma)^{-1} y = \tilde{y}(z)$ is unitary embedding
 $\tilde{y}(z)^* y(z) = \tilde{y}^*(z^{-1} - \gamma^*)^{-1} \pi^* \pi (z - \gamma)^{-1} y$

$$(1-z\bar{g}^*)^{-1} + \bar{z}'g(1-z'\bar{g})^{-1}$$

$$= (1-z\bar{g}^*)^{-1} \left((1-z\bar{g}^*)z^{-1}g + (1-z'\bar{g}) \right) (1-z'\bar{g})^{-1}$$

$$= (1-z\bar{g}^*)^{-1} (1-g^*g) (1-z'\bar{g})^{-1}$$

Is there an intrinsic way to do this.



so you get an ~~embedding~~ a map
 $Y \xrightarrow{\nu} (T/l_\omega)^* \otimes V \rightarrow (T/l_\omega)^* \otimes V/W$
 You want this for l_ω "real"

Roughly $\tilde{g}(z) = \pi(z-g)^{-1}y = \pi \bar{z}'(1-\bar{z}'g)^{-1}y$
 is analytic ~~on~~ ^{on} and outside $|z|=1$.

I guess what's intriguing is inner product between different $\pi(z-g)^{-1}y$. In the ~~hermitian~~ ^{partial} hermitian case what's interesting is the pairing between u^λ and u^μ . This seems to involve W and W^0 .

In the J-matrix case you have u^λ ~~entire in~~ ^{entire in} λ and a formula $(u^\mu, u^\lambda) = \frac{1}{\mu-\lambda}$ ~~hermitian form~~ ^{hermitian form} applied to bdy values.

~~the lines~~ the lines L_λ $L_{\bar{\lambda}}$ are orth.

so this leads us to ignore V and concentrate on the ~~the~~ family of $L_\lambda \subset W^0/W$

$$L_\omega \equiv W^0 \cap l_\omega \otimes Y$$

what can you say? $L_\omega^0 = W + l_\omega \otimes Y$

$$0 \rightarrow L_\omega \rightarrow W^0 \xrightarrow{n+2} (T/l_\omega) \otimes Y \xrightarrow{n+1} 0$$

Looks like $l_\omega = \mathcal{O}(-n-1)$.

82 ~~So~~ So you have ~~these~~ two of these lines L_0, L_{∞} . C_{ev}

Bring this discussion to an end.

$$W = T \otimes Y \supset L_2 \otimes Y$$

$$W^0 \cap (L_2 \otimes Y) = L_2 \text{ maps inj into } W^0/W$$

main ideas? From the data T, Y, W you seem to get the 2dim W^0/W (Kerim space) and this subline bundle L of $\mathcal{O} \otimes W^0/W$ over \mathbb{P}^1 with certain adjointness properties. Let's describe this as well as we can.

$$W = \begin{pmatrix} a \\ b \end{pmatrix} X \quad W^0 = W \oplus \begin{matrix} \text{Ker } a^* \\ \oplus \\ \text{Ker } b^* \end{matrix} \quad \text{You need}$$

$$W^0 \cap \begin{pmatrix} 1 \\ z \end{pmatrix} Y \quad \begin{pmatrix} y \\ zy \end{pmatrix} = \begin{pmatrix} ax \\ bx \end{pmatrix} + \begin{pmatrix} v^+ \\ v^- \end{pmatrix}$$

$$\text{i.e. } z(ax + v^+) = bx + v^- \quad -z$$

$$(az - b)x = -zv^+ + v^-$$

So the image of L_2 in $W^0/W = \begin{matrix} \text{Ker } (a^*) \\ \oplus \\ \text{Ker } (b^*) \end{matrix}$

Consists of ~~all~~ $\begin{pmatrix} v^+ \\ v^- \end{pmatrix}$ such that $-zv^+ + v^- \in (az - b)X$. Means all $\begin{pmatrix} v^+ \\ zS(z)v^+ \end{pmatrix} v^- \in V^-$.

$$S(z)(zv^+) = v^-$$

Notice that the degree of $zS(z)$ is $n+1$.

Next go back to $W \subset T \otimes Y \supset L_2 \otimes Y$

$L_2 = W^0 \cap (L_2 \otimes Y) \hookrightarrow W^0/W$. Pencil of hyperplane sections of degree $n+1$.

83 Intrinsically you have $L \simeq \mathcal{O}(n-1)$ embedded in $\mathcal{O} \otimes (W^0/W)$. If you take quotient lines of W^0/W (or lines) then you get ~~sections~~ maps $\mathcal{O} \otimes L \rightarrow \mathcal{O}$ i.e. ~~divisors~~ ^{divisors} of degree $n+1$. ~~There are various interesting~~ ~~point~~ metric possibilities. W^0/W has hermitian form so $P(W^0/W)$ has a real projective line. Now the herm. ~~form~~ form on W^0/W pulls back to the one which is restriction of given herm. form on $T \otimes Y$. But $\begin{pmatrix} y \\ zy \end{pmatrix} \mapsto (1-|z|^2) \|y\|^2$, so $z \mapsto L_z$ from $P(T)$ to $P(W^0/W)$ preserves real PT and ^{two} disks.

Let's start now with $W = \begin{pmatrix} z \\ A \end{pmatrix} X \subset \bigoplus_Y$ equipped with $\left(\begin{pmatrix} y_1 \\ y_2 \end{pmatrix}, \begin{pmatrix} 0 & 1 \\ -1 & 0 \end{pmatrix} \begin{pmatrix} y_1 \\ y_2 \end{pmatrix} \right) = (y_1, y_2) - (y_2, y_1)$
 W isotropic means $(\varepsilon x', Ax) = (Ax', \varepsilon x)$
 i.e. $A^* \varepsilon = \varepsilon^* A$ (ind of scalar prod on X).

Intrinsic version T 2diml Krein, Y $n+1$ diml Hilb, $T \otimes Y$ then $2n+2$ diml Krein, W ~~n~~ n diml isot. in $T \otimes Y$, W^0/W then 2 diml Krein, this is the ports or terminals. ~~the~~ $\omega \in PT$, l_ω corresp line in T , assume $W \cap (l_\omega \otimes Y) = 0$ $\forall \omega$ (no bound states), $l_\omega^0 = l_{\bar{\omega}}$, $\bar{\omega}$ = reflection of ω through the real P^1 given by the ~~real~~ null lines for the Krein form, so $W \cap (l_\omega \otimes Y) = 0 \Leftrightarrow W^0 + l_{\bar{\omega}} \otimes Y = T \otimes Y$. ~~is~~ \therefore as ω varies $l_\omega = W^0 \cap (l_\omega \otimes Y)$ is a line subbundle of W^0 , $0 \rightarrow L_\omega \rightarrow \underset{n+2}{W^0} \rightarrow \underset{n+1}{T/l_\omega \otimes Y} \rightarrow 0$, so $\{L_\omega\} \simeq \mathcal{O}(-n-1)$. Also $\omega \mapsto \text{Fung} \{L_\omega \hookrightarrow W^0/W\}$ gives an alg. map $PT \xrightarrow{Z} P(W^0/W)$, ~~this~~ covered by a line bundle $L \rightarrow \mathcal{O}(-1)$ ~~such that Krein form on W compatible with~~

84 Krein forms since the Krein form on W^0 descends to W^0/W . Z is the response function. It preserves the null circles and the $+$, $-$ disks, has degree $n+1$.

To get spectral rep for elements of Y choose V $W \subset V \subset W^0$ so that V/W is a ~~pos~~ ^{neg} line, then V^0/W is a ~~negative~~ ^{posit} line. Get spectrum of ω on $V \cap (L_\omega \otimes Y) \neq 0$ off the spectrum get $V \oplus L_\omega \otimes Y = 0$, this true for ω pos, since Krein form on V is < 0 and on $L_\omega \otimes Y$ is ≥ 0 . So spect in LHP. Off spectrum we have $V \xrightarrow{\sim} T/L_\omega \otimes Y$ and $V \rightarrow V/W$, so we get $O(-1) \otimes Y \xrightarrow{\sim} O \otimes V \rightarrow O \otimes V/W$.

~~the~~

Spectral repn. V/W neg line in W^0/W
 V^0/W corresp. pos. line. Claim $V \cap L_\omega \otimes Y = 0$ for $\text{Im}(\omega) \geq 0$ because the Krein form on $V-W$ is < 0 and ≥ 0 on $L_\omega \otimes Y$. Thus $V \xrightarrow{\sim} T/L_\omega \otimes Y$ in closed UHP.

What's important, what do I want to emphasize?

~~the~~ Krein space $T \oplus H$

What is the response of a trans line? Here H is infinite dim. A transm. line is the ^{direct} sum of a shift and its adjoint. ~~Look at one For you~~

Suppose Y Hilbert space with s such that $s^*s = I$ and $\text{Ker}(s^*)$ one dim. OK. $W = \begin{pmatrix} 1 \\ s \end{pmatrix} Y \subset \begin{matrix} Y \\ Y \end{matrix}$

$W^0 = \begin{pmatrix} s^* \\ 1 \end{pmatrix} Y$? $\left\{ \begin{pmatrix} y_1 \\ y_2 \end{pmatrix} \mid (y_1, y) = (y_2, sy) \quad \forall y \right\} = \left\{ \begin{pmatrix} s^*y_2 \\ y_2 \end{pmatrix} \right\}$

Then $W^0 = W \oplus \begin{pmatrix} 0 \\ \text{Ker}(s^*) \end{pmatrix}$ and $(W^0 \cap \begin{pmatrix} 1 \\ z \end{pmatrix} Y)$

$\begin{pmatrix} y \\ sy \end{pmatrix} = \begin{pmatrix} y \\ zy \end{pmatrix} \quad (z-s)y = 0 \implies (zs^* - 1)y = 0 \implies y = 0 \quad \text{for } |z| < 1.$

on infinite dims need to be careful.
~~Spectrum~~ Spectrum

$$(1 - z^{-1}s)y = 0 \Rightarrow y = 0 \text{ for } |z| < 1$$

You are confused. You probably should review what happens when $W = ?$ For a partial unitary $\begin{pmatrix} a & \\ & b \end{pmatrix} X \subset \bigoplus Y$ there is a complete picture for $|z| \neq 1$, namely two spectral representations associated to the contractors ba^* and ab^*

$$W^0 = W \oplus \begin{matrix} \text{Ker } a^* \\ \oplus \\ \text{Ker } b^* \end{matrix} \cap \begin{pmatrix} 1 \\ z \end{pmatrix} Y \ni \begin{pmatrix} ax \\ bx \end{pmatrix} + \begin{pmatrix} v^+ \\ v^- \end{pmatrix} = \begin{pmatrix} y \\ zy \end{pmatrix}$$

$$z(ax + v^+) = (bx + v^-)$$

$$(az - b)x = -zv^+ + v^-$$

$|z| < 1$: $v^- = (1 - ba^*)(1 - zab^*)^{-1}zv^+$

$|z| > 1$: $zv^+ = (1 - aa^*)(1 - z^{-1}ba^*)^{-1}v^-$

~~Suppose~~ Suppose $\begin{pmatrix} a \\ b \end{pmatrix} = \begin{pmatrix} 1 \\ s \end{pmatrix}$ with $s^*s = 1$.

$$ab^* = s^*$$

$$ba^* = s$$

$$W^0 = \begin{pmatrix} 1 \\ s \end{pmatrix} Y + \bigoplus_{\text{Ker}(s^*)}$$

In general you find the response is a map $V^+ \rightarrow V^-$ for $|z| < 1$. and a map $V^- \rightarrow V^+$ for $|z| > 1$. so in the shift case the response line is $\begin{pmatrix} 0 \\ 0 \end{pmatrix} \mathbb{C}$ for $|z| < 1$. and $\begin{pmatrix} 0 \\ 1 \end{pmatrix} \mathbb{C}$ for $|z| > 1$.

You need to make sense of transsm line theory
 First

86 Trans. line is direct sum of in and out

out: $V^+ \oplus uV^+ \oplus u^2V^+ \oplus \dots$

in $V^- \oplus u^{-1}V^- \oplus u^{-2}V^- \oplus \dots$

Intrinsic picture of ~~the system~~ ^{otherwise} is a $W \subset T \otimes Y$ such that W°/W is one dimensional, sign of herm. form on W°/W gives in or out type. ~~the sign is~~. To find out polarizing T : $W = \begin{pmatrix} a \\ b \end{pmatrix} X \subset \begin{pmatrix} Y \\ Y \end{pmatrix} \supset \begin{pmatrix} 1 \\ z \end{pmatrix} Y = l_z \otimes Y$.

W isotropic means $a^*a = b^*b = 1_X$. Find $W^\circ = W \oplus \begin{matrix} \text{Ker } a^* \\ \oplus \\ \text{Ker } b^* \end{matrix}$, so if ~~the~~ ^{W°/W} dim 1 either $\text{Ker } a^*$ or $\text{Ker } b^*$ is $\mathbb{C} \times$ other is zero.

Assume sign = - on W°/W , i.e. $W^\circ/W = \begin{matrix} \oplus \\ \text{Ker } b^* \end{matrix}$. Then ~~is~~ a isom, so $W = \begin{pmatrix} 1 \\ z \end{pmatrix} Y \subset \begin{pmatrix} Y \\ Y \end{pmatrix}$ where $z^*z = 1$ and $\text{Ker } z^*$ dim 1. $W^\circ = \begin{pmatrix} z^* \\ 1 \end{pmatrix} Y$. ~~ask when~~ Ask

now about response. When is $W^\circ \oplus \begin{pmatrix} 1 \\ z \end{pmatrix} Y = \begin{pmatrix} Y \\ Y \end{pmatrix}$? iff $\begin{matrix} Y \\ Y \end{matrix} \xrightarrow{\begin{pmatrix} z^* \\ 1 \end{pmatrix}} W^\circ \subset \begin{pmatrix} Y \\ Y \end{pmatrix} \xrightarrow{(z-1)}$ is an isom, i.e.

$1-zz^*$ is invertible. True for $|z| < 1$, get spectral embedding $Y \xrightarrow{\begin{pmatrix} z^* \\ 1 \end{pmatrix} (1-zz^*)^{-1}} W^\circ \rightarrow W^\circ/W = \text{Ker}(z^*)$

that is $y \mapsto (1-zz^*)(1-zz^*)^{-1}y$ ~~is~~ $(-z \ 1)$

Problem: When you discussed response you first asked ~~for~~ for $W \cap \begin{pmatrix} 1 \\ z \end{pmatrix} Y = 0$ and then found varying line L_z in W° whose ~~image~~ image in W°/W gives the response function. Spectrum not discussed until V chosen. Here there are only two choices for V , namely $V = W^\circ$ or $V = W$.

Properties of $1-zz^*$ for $|z| > 1$? Does it have kernel $a \neq 0$? $z^*(a_0, a_1, \dots) = (a_1, a_2, \dots) \stackrel{?}{=} z^{-1}(a_0, a_1, \dots)$
 $a_1 = z^{-1}a_0, a_2 = z^{-1}a_1, \dots, a_n = z^{-n}a_0$
 and this sequence is in l^2 . So ~~the~~ spectrum for $V = W^\circ$ is the closed disk $|z| \geq 1$.

87 Now take $V = W = \begin{pmatrix} 1 \\ g \end{pmatrix} Y \subset \begin{matrix} Y \\ Y \end{matrix}$

$$Y \xrightarrow{\begin{pmatrix} 1 \\ g \end{pmatrix}} V \subset \begin{matrix} Y \\ Y \end{matrix} \xrightarrow{(+z - 1)} Y \quad (z - 1) \begin{pmatrix} 1 \\ g \end{pmatrix} = z - g$$

$$(z - g)^{-1} = z^{-1} (1 - z^{-1}g)^{-1} = \sum_{n \geq 0} z^{-n-1} g^n \quad \text{is invertible for } |z| > 1.$$

~~But you don't see~~ but there is no line to project $\begin{pmatrix} 1 \\ g \end{pmatrix} (z - g)^{-1} y$ into.

Summarize. Considering $W = \begin{pmatrix} 1 \\ g \end{pmatrix} Y \subset \begin{matrix} Y \\ Y \end{matrix}$, $W^\circ = \begin{pmatrix} g^* \\ 1 \end{pmatrix} Y$ where $g^*g = 1$, $\text{Ker } g^* \text{ dim } 1$. First study the response, i.e. the intersection $L_z = W^\circ \cap \begin{pmatrix} 1 \\ z \end{pmatrix} Y$.

$$L_z = \begin{pmatrix} g^* \\ 1 \end{pmatrix} \text{Ker} (1 - zg^*: Y \rightarrow Y).$$

Case 1. $|z| > 1$. In this case L_z has dim 1 for all $|z| > 1$ including ∞ , and the line L_z projects onto W°/W .

Case 2. $|z| < 1$ In this case $L_z = 0$

~~Response~~ Response function for a transmission line. $\begin{matrix} Y \\ \oplus \\ Y \end{matrix}$ is the polarized Krein space

$$\begin{matrix} \oplus u^{-1}V^- \oplus V^- \oplus V^+ \oplus uV^+ & \|y\|^2 \\ \searrow \quad \swarrow \quad \searrow \quad \swarrow \\ \oplus u^{-1}V^- \oplus V^- \oplus V^+ \oplus uV^+ & -\|y_2\|^2 \end{matrix}$$

W is the graph of the arrows so that $(ax)^{\perp} = V^-$ up $\begin{pmatrix} V^- \\ 0 \end{pmatrix}$ and $(bx)^{\perp} = V^+$ down $\begin{pmatrix} 0 \\ V^+ \end{pmatrix}$ (observe signs are wrong)

$$\therefore W^\circ/W = \begin{pmatrix} V^- \\ \oplus \\ V^+ \end{pmatrix} \quad L_z = \begin{pmatrix} y \\ zy \end{pmatrix} \in W^\circ. \quad \text{Suppose } |z| < 1$$

start with $\xi \in V^+$, then you get

$$\begin{pmatrix} z^{-1}\xi + z^{-2}u\xi + z^{-3}u^2\xi + \dots \\ \xi + z^{-1}u\xi + z^{-2}u^2\xi + \dots \end{pmatrix} \in L_z$$

provided $|z| > 1$. And a similar element starting from

88 $\eta \in V^-$

$$\left(\begin{array}{c} \dots + zu^{-1}\eta + \eta \\ \dots + z^2u^{-1}\eta + z\eta \end{array} \right) \in L_z \quad \text{provided } |z| < 1.$$

These are the only possibilities for L_z ($|z| \neq 1$). Image of former is $V^+ \oplus V^-$ in W^0/W and image of the latter is $V^- \oplus V^+$ so the response function is

~~constant~~ constant ~~and the disks~~ each of the disks. We have

$$Z_z = \begin{cases} V^+ \oplus V^- & \text{for } |z| > 1 \\ V^- \oplus V^+ & \text{for } |z| < 1. \end{cases}$$

Next ~~to~~ couple a transmission line to a 1-port of type $O(n)$. You do this by means of an isomorphism between the terminals. Actually you take the direct sum of the V -Hilbert spaces and the direct sum of the W 's ~~with~~ with a sign change on the Krein form. Then you ~~need~~ need a maximal isot subspace of $W_1^0/W_1 \oplus W_2^0/W_2$, so there should be degenerate couplings. The dimensions are funny NO Krein isos are $U(2,2)$ $\dim 4$, Lagrangian subspaces descr. by unitaries $U(2)$ $\dim 4$.

~~to the problem~~ Review the situation. The problem is to understand coupling a partial unitary to a transms. line. The result is a unitary operator, only thing we can ask is the spectral measure arising from a convenient cyclic vector

89 Review. When you couple a 1-port to a transmission line you obtain a Hilbert space and unitary operator. ~~Only~~ There are two ^{obvious} cyclic vectors and a less obvious one. ~~from the deB.~~ The obvious ones are V^+, V^- associated ~~cyclic~~ measure is $\frac{d\theta}{2\pi}$. These are related by $S(z)$, factoring $S = P/g$ leads to less obvious ones. zeroes of g are outside S^1 . g is the deBranges function.

Question ~~whether the de Branges theory~~ how this coupling can be understood in terms of the response functions. Given two 1-ports if you couple them the spectrum is given by appropriate difference of the response functions

LC circuit. before considered $C' \oplus C$, as symplectic

$$\left. \begin{array}{l} E = L(-i\omega)I \\ I = C(-i\omega)E \end{array} \right\} \begin{array}{l} I \\ E \end{array}$$

You want to bring in power

$$P = EI \quad \int EI dt = \int LI \dot{I} dt = \frac{1}{2} LI^2 + \text{const.}$$

For a cap

$$\int EI dt = \int E C \dot{E} dt = \frac{1}{2} CE^2$$

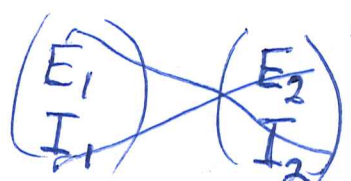
You would like to take 2 diml space of $\begin{pmatrix} E \\ I \end{pmatrix}$, allow E, I to be complex, define hermitian form signature $(1,1)$.

There has to be an intelligent way to handle this, somehow. You want to fit the situation into a $T \otimes Y$ somehow.

Consider $E = L \partial_t I$ $E = L(-i\omega)I$

power is? You have ~~the~~ T equipped with hermitian form. ~~the~~ You would like to associate a 2 diml space to each edge.

~~the~~ Have basic 2 plane of $\begin{pmatrix} E \\ I \end{pmatrix}$ and for ω the line $\begin{pmatrix} L(-i\omega) \\ 1 \end{pmatrix}$, hermitian form ~~the~~

$$\left(\begin{pmatrix} E_1 \\ I_1 \end{pmatrix}, \begin{pmatrix} 0 & i \\ -i & 0 \end{pmatrix} \begin{pmatrix} E_2 \\ I_2 \end{pmatrix} \right) = i(E_1 I_2 - I_1 E_2)$$


Maybe you need Kähler stuff

What you do in the real Lagrangian case for an LC circuit. Basic space is $C^1 \oplus C_1$ with hyperbolic skew form. This is the sum of hyperbolic planes for each edge. skew form is

$$\left\langle \begin{pmatrix} E_1 \\ I_1 \end{pmatrix}, \begin{pmatrix} E_2 \\ I_2 \end{pmatrix} \right\rangle = E_1 I_2 - I_1 E_2$$

and any line is of course isotropic. Now for frequency s you want the line $\begin{pmatrix} Ls \\ 1 \end{pmatrix} \mathbb{R}$

What's the relation between the skew form and the ~~the~~ hermitian form?

~~the~~ skew form $\begin{pmatrix} x'_1 \\ x'_2 \end{pmatrix} \begin{pmatrix} & 1 \\ -1 & \end{pmatrix} \begin{pmatrix} x_1 \\ x_2 \end{pmatrix}$
 $= x'_1 x_2 - x'_2 x_1$ extend to C^2 sdsq.

$$\frac{1}{i} \begin{pmatrix} \bar{z}'_1 \\ \bar{z}'_2 \end{pmatrix} \begin{pmatrix} & 1 \\ -1 & \end{pmatrix} \begin{pmatrix} z_1 \\ z_2 \end{pmatrix} = \frac{\bar{z}'_1 z_2 - \bar{z}'_2 z_1}{i}$$

~~skew~~ herm. form

if $z' = \bar{z}$ $\frac{\bar{z}_1 z_2 - \bar{z}_2 z_1}{i} = 2 \Im(\bar{z}_1 z_2)$ seems to be type (1,1)

91 Try harder. First point is that a symplectic space when complexified is naturally a Krein space. Why. ~~Let~~ Equivalence between hermitian and skew herm. forms. ~~Take~~ Take \mathbb{R}^2 and a skew form

$$\omega\left(\begin{pmatrix} x_1 \\ x_2 \end{pmatrix}, \begin{pmatrix} y_1 \\ y_2 \end{pmatrix}\right) = \left(\begin{pmatrix} x_1 \\ x_2 \end{pmatrix}, \begin{pmatrix} 0 & a \\ -a & 0 \end{pmatrix} \begin{pmatrix} y_1 \\ y_2 \end{pmatrix}\right) \\ = a \begin{vmatrix} x_1 & y_1 \\ x_2 & y_2 \end{vmatrix} \quad a \text{ real.}$$

extend sesquilinear to the complexification

$$\omega\left(\begin{pmatrix} x_1 \\ x_2 \end{pmatrix}, \begin{pmatrix} y_1 \\ y_2 \end{pmatrix}\right) = a \begin{vmatrix} \bar{x}_1 & y_1 \\ \bar{x}_2 & y_2 \end{vmatrix} \\ = a \begin{pmatrix} x_1 \\ x_2 \end{pmatrix}^* \begin{pmatrix} 0 & 1 \\ -1 & 0 \end{pmatrix} \begin{pmatrix} y_1 \\ y_2 \end{pmatrix} \quad \text{skew herm.}$$

if you mult. by i then $\begin{pmatrix} 0 & i \\ -i & 0 \end{pmatrix} = \begin{pmatrix} 1 & \\ & -1 \end{pmatrix}$

$$\begin{pmatrix} 0 & 1 \\ -1 & 0 \end{pmatrix} \begin{pmatrix} 1 & 1 \\ +i & -i \end{pmatrix} = \begin{pmatrix} 1 & 1 \\ +i & -i \end{pmatrix} \begin{pmatrix} i & 0 \\ 0 & -i \end{pmatrix}$$

~~So you want to define~~ The problem is to fit LC circuits into your abstract framework. $T_{\mathbb{R}^2}$ should be specified. \mathbb{R}^2 with volume $\begin{vmatrix} x_1 & y_1 \\ x_2 & y_2 \end{vmatrix} = x^* \begin{pmatrix} 0 & 1 \\ -1 & 0 \end{pmatrix} y$, complexified becomes \mathbb{C}^2 with $\begin{pmatrix} \pm i \end{pmatrix} \begin{vmatrix} \bar{x}_1 & y_1 \\ \bar{x}_2 & y_2 \end{vmatrix}$

you want $\frac{1}{i} \begin{vmatrix} 1 & 1 \\ \bar{\omega} & \omega \end{vmatrix} = \frac{\omega - \bar{\omega}}{i} > 0$ for $\omega \in \text{UHP}$.

In the case of an inductance, you have $\begin{pmatrix} E \\ I \end{pmatrix} \in \mathbb{R}^2$ with

$\begin{vmatrix} E_1 & E_2 \\ I_1 & I_2 \end{vmatrix}$, hence herm. form $\frac{1}{i} \begin{vmatrix} \bar{E}_1 & E_2 \\ \bar{I}_1 & I_2 \end{vmatrix}$ on \mathbb{C}^2 .

$$L_{\omega} = \begin{pmatrix} L(-i\omega) & \\ & 1 \end{pmatrix} \mathbb{C} \quad \left| \begin{array}{cc} L(+i\bar{\omega}) & L(i\omega) \\ & 1 \end{array} \right| \\ = L i(\bar{\omega} + \omega) = \text{~~some expression~~}$$

q2 capacitance

$$L_{\omega} = \begin{pmatrix} 1 \\ C(-i\omega) \end{pmatrix} \mathbb{C}$$

$$\begin{vmatrix} 1 & 1 \\ C(i\bar{\omega}) & C(-i\omega) \end{vmatrix} = C \begin{matrix} (-i\omega - i\bar{\omega}) \\ \text{cancel} \end{matrix} = -Ci(\omega + \bar{\omega})$$

try the line

$$L_s = \begin{pmatrix} L_s \\ 1 \end{pmatrix} \mathbb{C}$$

$$\begin{vmatrix} L_{\bar{s}} & L_s \\ 1 & 1 \end{vmatrix} = L(\bar{s} - s)$$

$$L_s = \begin{pmatrix} 1 \\ C_s \end{pmatrix} \mathbb{C}$$

$$\begin{vmatrix} 1 & 1 \\ C_{\bar{s}} & C_s \end{vmatrix} = C(s - \bar{s})$$

This is a pair. How do I proceed to organize this?

Concentrate on what you have on $\mathbb{C}^1 \oplus \mathbb{C}_1^{\oplus}$

Pairing between factors. Important is for each s

a subspace $N_s \subset \mathbb{C}^1 \oplus \mathbb{C}_1$. ~~There is an L-part~~

This is the direct sum of L, C situations

$$N_s^{\text{ind}} = \begin{pmatrix} L_s \\ \boxed{\text{square}} \end{pmatrix} \mathbb{C}_1^{\text{ind}}$$

$$N_s^{\text{cap}} = \begin{pmatrix} 1 \\ C_s \end{pmatrix} \mathbb{C}^{\text{cap}}$$

$$\begin{pmatrix} x_1 \\ x_2 \end{pmatrix}^* \begin{pmatrix} 0 & i \\ -i & 0 \end{pmatrix} \begin{pmatrix} y_1 \\ y_2 \end{pmatrix} = i \begin{vmatrix} \bar{x}_1 & y_1 \\ \bar{x}_2 & y_2 \end{vmatrix} \quad \begin{vmatrix} 1 & 1 \\ \bar{\omega} & \omega \end{vmatrix} = \omega - \bar{\omega}$$

$$\begin{pmatrix} x_1 \\ x_2 \end{pmatrix}^* \begin{pmatrix} 0 & 1 \\ 1 & 0 \end{pmatrix} \begin{pmatrix} y_1 \\ y_2 \end{pmatrix} = \bar{x}_1 y_2 + \bar{x}_2 y_1 \quad |s + \bar{s}| = s + \bar{s}$$

$${}_{x^0} y = \begin{pmatrix} 0 & 1 \\ 1 & 0 \end{pmatrix} \begin{pmatrix} 0 & i \\ -i & 0 \end{pmatrix} = \begin{pmatrix} -i & \\ & i \end{pmatrix}$$

~~Now you have to divide~~

Now you understand T .

γ is 1 dim

~~These~~ It seems that I need another ingredient

T is fixed, say $T = \mathbb{D}^2$ with $\begin{pmatrix} x_1 \\ x_2 \end{pmatrix}^* \begin{pmatrix} 0 & 1 \\ 1 & 0 \end{pmatrix} \begin{pmatrix} y_1 \\ y_2 \end{pmatrix}$

~~LC circuit~~
 $l_s = \begin{pmatrix} 1 \\ s \end{pmatrix} \in \mathbb{C} \subset \mathbb{T}$ $l_s^\circ = \left\{ \begin{pmatrix} y_1 \\ y_2 \end{pmatrix} \mid \begin{pmatrix} 1 \\ s \end{pmatrix}^* \begin{pmatrix} 0 & 1 \\ 1 & 0 \end{pmatrix} \begin{pmatrix} y_1 \\ y_2 \end{pmatrix} = 0 \right.$

$\therefore l_s^\circ = l_{-\bar{s}}$ So what do you get??
 $y_2 + \bar{s}y_1 = 0$
 $y_2 = -\bar{s}y_1$

So what next? There is a difficulty here. It seems that all we get is ~~is not~~ in the case ~~LC~~ $\dim(Y) = 1$ is ~~is~~?
 What do you need or want?

LC circuit to what extent is it ~~like~~ a harmonic oscillator - it should be except for $\omega = 0, \infty$.
 discuss ~~the~~ modes of the homogeneous system. Have real space $\mathbb{C}^1 \oplus \mathbb{C}_1$ symplectic structure given by natural pairing $\mathbb{C}^1 \times \mathbb{C}_1 \rightarrow \mathbb{R}$. (determined up to sign)

Γ_s have impedance subspace $\Gamma_s \subset \mathbb{C}^1 \oplus \mathbb{C}_1$
 subspaces $\begin{pmatrix} l_s \\ 1 \end{pmatrix} \mathbb{C}_{ind} \oplus \begin{pmatrix} 1 \\ c_s \end{pmatrix} \mathbb{C}_{cap}$ L, C positive
 def. quadratic forms, so Γ_s is Lagrangian. Another Lag. subsp is $W = \mathbb{S}\mathbb{C}^0 \oplus \mathbb{Z}_1$. ~~three modes~~ spectrum is ~~is~~

consists of $s \in W$ not transverse to Γ_s . But you need s complex. So basically you need to complexify phase space

Try again. Basic ~~is~~ ~~is~~ $\begin{pmatrix} \{E_s\} \\ \{I_s\} \end{pmatrix} = N_s$
 runs over the edges. $\Gamma_s =$ ~~is~~ subspace $\ni \begin{matrix} E_s = Ls \text{ ind} \\ I_s = \frac{1}{Cs} \text{ cap.} \end{matrix}$

so $\{\Gamma_s\}$ subvector bundle of $\mathbb{O} \otimes \mathbb{N}$

Analyze hermitian forms. equivalent on a $v.s.$
 hermitian bilinear form $H(x,y)$
 real symm $S(x,y)$ on underlying real v.s. $\ni S(ix,y) = S(x,y)$
 real skew-symm $A(x,y)$ $\ni A(ix,y) = A(x,y)$
 real quadratic form $Q(x)$ $\ni Q(ix) = Q(x)$

$H(x,y) = S(x,y) + iA(x,y)$ real + imag parts.
 $H(x,y) = H(y,x) \iff S$ symm, A skew symm.

$$94 \quad S(x, iy) + iA(x, iy) = i(S(x, y) + iA(x, y))$$

$$\therefore A(x, iy) = S(x, y)$$

$$A(y, ix) = -A(ix, y) = -A(i^2x, iy) = A(x, iy)$$

A skew $\Leftrightarrow S$ symm.

$$Q(x) = S(x, x) \cong A(x, ix)$$

$\cong H(x, x)$

Suppose V is the complexification of $V_{\mathbb{R}}$.
 Choose basis: $V = \mathbb{C}^n$, $V_{\mathbb{R}} = \mathbb{R}^n$. A hermitian $H(x, y)$
 same as herm. matrix which splits into a
 real symm matrix + i times skew symm. matrix.
 Point is that $H(x, y)$ is determined by sesquilinearity
 to $x, y \in V_{\mathbb{R}}$ and then $H(x, y) = \underbrace{S(x, y)}_{\text{real symm}} + i \underbrace{A(x, y)}_{\text{real skew symm.}}$

~~$$H(x_1 + ix_2, y_1 + iy_2) = (H(x_1, y_1) + H(x_2, y_2)) + i(H(x_1, y_2) - H(x_2, y_1))$$~~

$$S = 0 \Leftrightarrow H(x, x) = S(x, x) = 0 \quad \forall x \in V_{\mathbb{R}}$$

Thus equivalence between skew symm. forms on $V_{\mathbb{R}}$
 and herm forms on $V_{\mathbb{C}}$ such that $V_{\mathbb{R}}$ is isotropic.

Back to LC circuit. You have $\mathbb{C}^1 \oplus \mathbb{C}^1$
 = phase space ~~real symm~~ real with a
 symplectic form (up to \pm). Complexification then
 has natural Kerein form.

Notice that any real subspace isotropic wrt A
 is n ^{antim.} isotropic wrt H since $S(x, x) = 0$ for $x \in V_{\mathbb{R}}$.

95 ~~What is the symplectic form?~~ You have to look at $V_n = \left\{ \begin{pmatrix} E \\ I \end{pmatrix} \in \mathbb{R}^{2n} \right\}$ skew form is

$$\begin{pmatrix} E_1 \\ I_1 \end{pmatrix} \begin{pmatrix} 0 & 1 \\ -1 & 0 \end{pmatrix} \begin{pmatrix} E_2 \\ I_2 \end{pmatrix} = E_1 I_2 - I_1 E_2 = \begin{vmatrix} E_1 & E_2 \\ I_1 & I_2 \end{vmatrix}$$

The corresponding hermitian form should be

$$\begin{pmatrix} E_1 \\ I_1 \end{pmatrix}^* \begin{pmatrix} 0 & i \\ -i & 0 \end{pmatrix} \begin{pmatrix} E_2 \\ I_2 \end{pmatrix} = i \begin{vmatrix} \bar{E}_1 & E_2 \\ \bar{I}_1 & I_2 \end{vmatrix}$$

when $\begin{pmatrix} E \\ I \end{pmatrix} = \begin{pmatrix} E_j \\ I_j \end{pmatrix}$ $j=1,2$. $i(\bar{E}I - \bar{I}E)$
 $= 2 \operatorname{Im}(\bar{I}E)$

so what next? Impedance $\begin{pmatrix} E \\ I \end{pmatrix} = \begin{pmatrix} Ls \\ 1 \end{pmatrix} I$

$$i \begin{vmatrix} \bar{Ls} \bar{I} & Ls I \\ \bar{I} & I \end{vmatrix} = iL(\bar{s}-s)|I|^2 = 2\operatorname{Im}(s)L|I|^2$$

$$i \begin{vmatrix} \bar{E} & E \\ C\bar{s}\bar{E} & CsE \end{vmatrix} = iC|E|^2(s-\bar{s}) = C|E|^2(-2\operatorname{Im}s)$$

E, I \otimes

Think real $\begin{matrix} I_L \rightarrow \\ \left[\begin{matrix} I_L \\ E_C \end{matrix} \right] \\ \left[\begin{matrix} E_L \\ I_C \end{matrix} \right] \end{matrix}$ Phase space \mathbb{H} dim 4 before the symplectic reduction.

$$E_L = L \dot{I}_L \quad C \dot{E}_C = \dot{I}_C$$

$$I_L = I_C \quad \text{and} \quad E_L = -E_C$$

96 Try to understand what you can ~~need to consider~~ show the eigenvalues are purely imaginary. The argument involves real spaces and quadratic forms. The argument involves Siegel UHP with positive real part. ~~Consider all~~ Consider all

Have Lagrangian subspace

$$\begin{matrix} \delta\mathbb{C}^0 & \subset & \mathbb{C}^1 \\ \oplus & & \\ \mathbb{Z}_1 & \subset & \mathbb{C}_1 \end{matrix}$$

$$\begin{array}{ccccccc} 0 & \rightarrow & \delta\mathbb{C}^0 & \rightarrow & \mathbb{C}^1 & \rightarrow & \mathbb{R}/\mathbb{Z} \rightarrow 0 \\ & & & & \downarrow N_s & & \\ 0 & \leftarrow & \mathbb{C}/\mathbb{Z}_1 & \leftarrow & \mathbb{C}_1 & \leftarrow & \mathbb{Z}_1 \leftarrow 0 \end{array}$$

when is $W = \delta\mathbb{C}^0 \oplus \mathbb{Z}_1$ transo. to Γ_{N_s}

$$\begin{matrix} \delta\mathbb{C}^0 \\ \oplus \\ \mathbb{Z}_1 \end{matrix} \cap \begin{pmatrix} 1 \\ N_s \end{pmatrix} \mathbb{C}^1 \ni \begin{matrix} \omega \\ N_s \omega \end{matrix}$$

The intersection is $\{\omega \in \delta\mathbb{C}^0 \mid N_s \omega \in \mathbb{Z}_1\}$. What argument to give that this can't happen unless $\text{Re}(s) = 0$. The argument is by self-pairing. You take

say $x = (x_C, x_L)$ that $\begin{pmatrix} x_L \\ x_C \end{pmatrix}^* N_s \begin{pmatrix} x_L \\ x_C \end{pmatrix} = x_L^* \frac{1}{Ls} x_L + x_C^* Cs x_C$ diagonal matrices

$$\begin{pmatrix} x_L \\ x_C \end{pmatrix}^* N_s \begin{pmatrix} x_L \\ x_C \end{pmatrix} = x_L^* \frac{1}{Ls} x_L + x_C^* Cs x_C$$

97 ~~Example~~ Example. Let V be a complex vector space, let $V^t =$ anti dual = dual with opposite complex structure, have pairing $V^t \otimes V \rightarrow \mathbb{C}$ which is ~~sesquilinear~~ sesquilinear.

~~is clear how~~ $\langle t, v \rangle$ $t \in V^t, v \in V$. ~~is clear how~~

~~is clear how~~ ~~make hermitian symmetric~~. So on ~~$V^t \oplus V$~~ $V^t \oplus V$ you have a ~~form~~ form.

sesquilinear form, namely $\begin{pmatrix} t_1 \\ v_1 \end{pmatrix}, \begin{pmatrix} t_2 \\ v_2 \end{pmatrix} \mapsto \langle t_1, v_2 \rangle$

which you can symmetrize in herm.

$$H\left(\begin{pmatrix} t_1 \\ v_1 \end{pmatrix}, \begin{pmatrix} t_2 \\ v_2 \end{pmatrix}\right) = \langle t_1, v_2 \rangle + \underbrace{\langle t_2, v_1 \rangle}_{\substack{\text{linear in } t_2 \\ \text{anti linear in } v_1}}$$

~~Let~~ Given V, W \mathbb{C} -vector spaces, ~~then~~ Let $F(v, w)$ be sesquilinear: ~~be~~ linear in w anti-linear in v , e.g. $V = \mathbb{C}^m, W = \mathbb{C}^n$, α $m \times n$ matrix $F(v, w) = v^* \alpha w$. ~~Let~~ Get

another sesq. form $G(w, v) = \overline{F(v, w)} = w^* \alpha^* v$

and then ~~$H\left(\begin{pmatrix} v_1 \\ w_1 \end{pmatrix}, \begin{pmatrix} v_2 \\ w_2 \end{pmatrix}\right) = F(v_1, w_2) + G(w_1, v_2)$~~

$$= v_1^* \alpha w_2 + w_1^* \alpha^* v_2 = \begin{pmatrix} v_1^* & w_1^* \end{pmatrix} \begin{pmatrix} 0 & \alpha \\ \alpha^* & 0 \end{pmatrix} \begin{pmatrix} v_2 \\ w_2 \end{pmatrix}$$

~~so there is a natural analogue~~ ~~if you pick a Hilbert space structure, then~~

so it seems that there is a ~~an~~ Kreinian analogue of symplectic.

~~Let's go~~ back to LC circuit. Begin with space of $\begin{pmatrix} E \\ I \end{pmatrix} \in \mathbb{C}^2$ and the Hermitian bilinear form

$$\begin{pmatrix} E_1 \\ I_1 \end{pmatrix}^* \begin{pmatrix} 0 & 1 \\ 1 & 0 \end{pmatrix} \begin{pmatrix} E_2 \\ I_2 \end{pmatrix} = \bar{E}_1 I_2 + \bar{I}_1 E_2 = 2 \operatorname{Re}(\bar{E}_1 I_1) \quad \text{if } \begin{matrix} E_2 = E_1 \\ I_2 = I_1 \end{matrix}$$

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$$\begin{pmatrix} Ls \\ 1 \end{pmatrix}^* \begin{pmatrix} 0 & 1 \\ 1 & 0 \end{pmatrix} \begin{pmatrix} Ls \\ 1 \end{pmatrix} = L\bar{s} + Ls = L(s+\bar{s}) = \underline{2} \operatorname{Re}(Ls)$$

$$\begin{pmatrix} 1 \\ Cs \end{pmatrix}^* \begin{pmatrix} 0 & 1 \\ 1 & 0 \end{pmatrix} \begin{pmatrix} 1 \\ Cs \end{pmatrix} = (1 \quad C\bar{s}) \begin{pmatrix} 1 \\ Cs \end{pmatrix} = C(2\operatorname{Re}(Cs)) = \underline{1} = 9.35 \text{ PF}$$

OKAY let's check. You have the above standard hermitian form on $C^1 \oplus C^1$ namely

$$E\bar{I} + \bar{I}E = 2\operatorname{Re}(E\bar{I}), \text{ pairing } \begin{pmatrix} E_1 \\ I_1 \end{pmatrix}^* \begin{pmatrix} 0 & 1 \\ 1 & 0 \end{pmatrix} \begin{pmatrix} E_2 \\ I_2 \end{pmatrix}.$$

Next have subspace $Z_1 \subset C_1$ which is isotropic, and the annihilator is $\delta C^0 \oplus C_1$, so there should be no problem with $\delta C^0 \oplus Z_1$ being maximal isotropic

Summarize. ~~Summarize~~ You made some progress toward linking LC circuits to the invariant version of p-unitaries

Consider LC network. Up to now you have studied the "configuration space" viewpoint, namely, you pick say the voltage space C^1 as config. space. you ~~get~~ have a quadratic (hermitian) form on C^1 depending on s ?

Let's get this straight. ~~Start~~ Start with a real situation and then complexify. Real situation is a vector space D , the dual space D^* , a quadratic form Q_s on D depending on s a subspace Z of D spectrum ~~is the~~ those s such that Q_s is nondeg.

LC network. You are used to a "configuration space version."

99 LC network "configuration space" version

You have a real voltage function space C^1 and dual current function space C_1 , ~~the~~ the impedance of the edges yields a map $N_s: C^1 \rightarrow C_1$ for real s direct sum of types $N_s = \begin{pmatrix} Ls \\ 1 \end{pmatrix} C_1$ or $\begin{pmatrix} 1 \\ Cs \end{pmatrix} C^1$

Your configuration space is a real space V together with a quadratic form $Q_s(v)$ which is the direct sum $V_L \oplus V_C$ $Q_s = (Ls)^{-1} \oplus Cs$.

Check: $(E, I) = (E, Q_s E) = s^{-1} \begin{pmatrix} E_L \\ L^{-1} E_L \end{pmatrix} + s \begin{pmatrix} E_C \\ C E_C \end{pmatrix}$

so you have a real vector space V split into $V_L \oplus V_C$ and ~~the~~ $Q_s = \underbrace{s^{-1} \begin{pmatrix} E_L \\ L^{-1} E_L \end{pmatrix}}_L \oplus \underbrace{s \begin{pmatrix} E_C \\ C E_C \end{pmatrix}}_C$

for pos. def. quad forms L^{-1}, C $s^{-1} L^{-1}(E) + s C(E)$ on V_L and V_C resp. Next we have a subquotient of V - restrict to conservative voltage functions and divide by ~~voltage functions~~ ^{node potentials supported} ~~node potentials supported~~ on the ext. vertices. Look at the induced quadratic form ~~Q~~ on subquotient.

Real Quadratic form version.

Complex hermitian version ^{should be:} Exactly the same, namely V is a complex vector space split into $V_L \oplus V_C$ with hermitian pos. def forms ~~on~~ L^{-1} and C on V_L, V_C resp. Get a modified version

You want to organize, merge two themes:

analysis of partial unitaries - here you encounter isotropic subspaces in a Krein space. In this theory ~~of~~ there is a Hilbert space \mathcal{Y} around LC networks. ~~The~~ somehow this is adapted a symplectic or Krein viewpoint.

You need to double - hermitian forms become isotropic subspaces.

100 ~~But the full story~~ Complexify an LC network.
Originally E, I are real functions of t .

~~Consider~~ You need to double! You have C^1 and C_1 spaces of ^{complex} edge voltage functions and edge current functions. These are anti-dual in a ~~preferred~~ preferred way because of the "power" $\bar{I}E$ for each edge.

~~Given~~ Given $E(t), I(t)$ ^{real} with compact support, ~~then~~ then $\int_{-\infty}^{\infty} E(t) I(t) dt =$ power into the edge

$$\int_{-\infty}^{\infty} E(\omega) I(-\omega) \frac{d\omega}{2\pi} = \int_{-\infty}^{\infty} \overline{I(\omega)} E(\omega) \frac{d\omega}{2\pi}$$

I guess my point is that complex "phase space" is $C^1 \oplus C_1$ and it carries a natural hermitian form, hyperbolic type; also skew-hermitian multiplies by i .

The impedance of each edge yields a subbundle $N_s \subset C^1 \oplus C_1$ direct sum of either $\begin{pmatrix} Ls \\ 1 \end{pmatrix} \mathbb{C}$ or $\begin{pmatrix} 1 \\ Cs \end{pmatrix} \mathbb{C}$ for ~~the~~ $\text{Re}(s) = 0$ this subspace should be isotropic

~~of M_s~~ ~~is~~. The quotient bundle $s \mapsto C^1 \oplus C_1 / N_s$ is holom. + pure of type $(0,1)$, so we can identify $C^1 \oplus C_1$ with the space of holom. sections. If we ~~split $C^1 \oplus C_1$ into~~ should get a canonical isomorphism

$$T \otimes Y \xrightarrow{\sim} C^1 \oplus C_1 \quad \text{OKAY}$$

~~False~~ polarized Hilbert space $U_+ \oplus U_-$

~~Just point~~ Maybe all that's involved is changing to take Y to be a ~~pre~~ Krein space and then $T \otimes Y$ should have the tensor product

101. You need some improvement

Begin with a complex vector space Ω
 form direct sum $D = \Omega \oplus \Omega^\dagger$ where Ω^\dagger is
 the anti-dual, so we have a sesquilinear pairing
 $\Omega^\dagger \otimes_{\mathbb{R}} \Omega \xrightarrow{\langle, \rangle} \mathbb{C}$. Define ^{skew} hermitian form on D by

$$H\left(\begin{pmatrix} x \\ \lambda \end{pmatrix}, \begin{pmatrix} x_1 \\ \lambda_1 \end{pmatrix}\right) = -\overline{\langle \lambda_1, x \rangle} + \langle \lambda, x_1 \rangle$$

Consider a graph $\begin{pmatrix} 1 \\ T \end{pmatrix} \Omega \subset D$. When is this isotropic?

$$H\left(\begin{pmatrix} x \\ Tx \end{pmatrix}, \begin{pmatrix} x_1 \\ Tx_1 \end{pmatrix}\right) = -\overline{\langle Tx_1, x \rangle} + \langle Tx, x_1 \rangle = 0$$

means? $T: \Omega \rightarrow \Omega^\dagger$

$$T^\dagger: \Omega^{\dagger\dagger} \rightarrow \Omega^\dagger$$



defined by $\langle T^\dagger x, x' \rangle$

$$\langle T^\dagger x, x' \rangle = \overline{\langle Tx', x \rangle}$$

means T is hermitian, i.e. $\langle Tx, x' \rangle$ herm. symmetric in x, x' .

Ω complex v.s. Ω^\dagger anti-dual, a map $T: \Omega \rightarrow \Omega^\dagger$
 is equivalent to a sesquilinear form $H(x, x') = \langle Tx, x' \rangle$

$T: \Omega \rightarrow \Omega^\dagger$ same as ~~an anti~~ $T: \Omega \rightarrow \Omega^*$ $T(cx) = \bar{c}T(x)$

$$T^\dagger: \Omega^{\dagger\dagger} \rightarrow \Omega^\dagger \quad \langle T^\dagger x, x' \rangle \quad Tx'$$

Assume Ω is a Hilb space so that one has a canonical
 isom $\Omega \xrightarrow{\sim} \Omega^\dagger$. Then $\Omega \oplus \Omega^\dagger = \Omega \oplus \Omega$ equipped with the
 skew herm. form $\begin{pmatrix} x_1 \\ x_2 \end{pmatrix}^* \begin{pmatrix} 0 & 1 \\ -1 & 0 \end{pmatrix} \begin{pmatrix} y_1 \\ y_2 \end{pmatrix}$. One has $\left(\begin{pmatrix} 1 \\ T \end{pmatrix} \Omega\right)^\circ = \begin{pmatrix} 1 \\ T^* \end{pmatrix} \Omega$

$$\begin{pmatrix} y_1 \\ y_2 \end{pmatrix}^* \underbrace{\begin{pmatrix} 0 & 1 \\ -1 & 0 \end{pmatrix}}_{\begin{matrix} \overline{Tx} \\ -x \end{matrix}} \begin{pmatrix} x \\ Tx \end{pmatrix} = y_1^* Tx - y_2^* x = \left((T^* y_1)^* - y_2^* \right) x = 0 \text{ for } x \Rightarrow y_2 = T^* y_1$$

~~Suppose~~ Suppose now Ω is a polarized Hilbert space $\Omega = \Omega^+ \oplus \Omega^-$ and we equip it with the hermitian operator ~~$\begin{pmatrix} \omega & 0 \\ 0 & -\omega^{-1} \end{pmatrix}$~~ $\begin{pmatrix} \omega & 0 \\ 0 & -\omega^{-1} \end{pmatrix} = \omega \pi_+ - \omega^{-1} \pi_-$

Then ~~for~~ for each $\omega \in \mathbb{P}^1$ you have a subspace of Ω , namely the graph of this operator which is isotropic ~~wrt.~~ wrt. the canonical skew-herm. form where ω is real. The point to make perhaps is that ~~you~~ you get a ^{holom.} subbundle over \mathbb{P}^1 of $\mathcal{O}(\Omega \oplus \Omega)$.

~~Review~~ Know that this holom. subbundle is pure of type $\mathcal{O}(-1)$. Things you ~~know~~ know.

$$0 \longrightarrow \Gamma_{\omega} \longrightarrow \mathcal{O} \otimes \frac{\Omega}{\Omega} \longrightarrow Q_{\omega} \longrightarrow 0$$

$$\mathcal{O}(\Omega^{\oplus 2}) \text{ canon. isom. to } \Gamma_{\text{hol}}(\mathbb{P}^1, \mathcal{O}(2))$$

~~Take Hilbert space~~

Review. Start with a Hilbert space ~~X~~, form double ~~X~~ with ^{skew-}hermitian form $\begin{pmatrix} x_1 \\ x_2 \end{pmatrix}^* \begin{pmatrix} 0 & 1 \\ -1 & 0 \end{pmatrix} \begin{pmatrix} x'_1 \\ x'_2 \end{pmatrix}$

If $T: X \rightarrow X$ is linear then

$$\begin{aligned} \left(\begin{pmatrix} 1 \\ T \end{pmatrix} X \right)^{\circ} &= \left\{ \begin{pmatrix} x'_1 \\ x'_2 \end{pmatrix} \mid \begin{pmatrix} x \\ Tx \end{pmatrix}^* \begin{pmatrix} 0 & 1 \\ -1 & 0 \end{pmatrix} \begin{pmatrix} x'_1 \\ x'_2 \end{pmatrix} = 0 \right. \\ & \quad \left. \begin{aligned} x^* x'_2 &= (Tx)^* x'_1 \\ x'_2 &= T^* x'_1 \end{aligned} \right. \quad \forall x \\ &= \begin{pmatrix} 1 \\ T^* \end{pmatrix} X \end{aligned}$$

So that $\begin{pmatrix} 1 \\ T \end{pmatrix} X$ is isotropic $\Leftrightarrow T = T^*$.

$$\begin{pmatrix} 0 & 1 \\ -1 & 0 \end{pmatrix}^* \begin{pmatrix} 0 & 1 \\ -1 & 0 \end{pmatrix} \begin{pmatrix} 0 & 1 \\ -1 & 0 \end{pmatrix} = \begin{pmatrix} 0 & -1 \\ 1 & 0 \end{pmatrix} \begin{pmatrix} 0 & 1 \\ -1 & 0 \end{pmatrix} \begin{pmatrix} 0 & 1 \\ -1 & 0 \end{pmatrix}$$

103 Consider the \mathbb{C} double $\begin{matrix} X \\ \oplus \\ X \end{matrix}$. Then $\begin{pmatrix} 0 & 1 \\ -1 & 0 \end{pmatrix}$ ~~is an~~
~~autom~~ preserves the skew-hermitian form:

$$\begin{pmatrix} x_1 \\ x_2 \end{pmatrix}^* \begin{pmatrix} 0 & 1 \\ -1 & 0 \end{pmatrix}^* \begin{pmatrix} 0 & 1 \\ -1 & 0 \end{pmatrix} \begin{pmatrix} 0 & 1 \\ -1 & 0 \end{pmatrix} \begin{pmatrix} x'_1 \\ x'_2 \end{pmatrix} \quad \text{and it carries } \begin{pmatrix} 1 \\ \omega \end{pmatrix} X$$

into $\begin{pmatrix} 1 \\ -\omega \end{pmatrix} X$.

So it seems that

$$\begin{pmatrix} 0 & 1 \\ -1 & 0 \end{pmatrix} : \begin{matrix} X \\ \oplus \\ X \end{matrix} \longrightarrow \begin{matrix} X \\ \oplus \\ X \end{matrix}$$

is an autom of the skew-herm. form and it carries

$$\begin{pmatrix} 1 \\ \omega \end{pmatrix} X \quad \text{into} \quad \begin{pmatrix} \omega \\ -1 \end{pmatrix} X = \begin{pmatrix} 1 \\ -\omega^{-1} \end{pmatrix} X.$$

me ~~how~~ how to treat an LC circuit in the framework of $T \otimes Y$ where $T = \begin{matrix} \mathbb{C} \\ \oplus \\ \mathbb{C} \end{matrix}$ standard skew-herm. f.
 $l_\omega = \begin{pmatrix} 1 \\ \omega \end{pmatrix} \mathbb{C}$ and Y is a Hilbert space ~~of analyzing~~

$$\begin{pmatrix} E \\ I \end{pmatrix} \in \mathbb{C}^2 \quad \text{skew-herm. form is} \quad \begin{pmatrix} E_1 \\ I_1 \end{pmatrix}^* \begin{pmatrix} 0 & 1 \\ -1 & 0 \end{pmatrix} \begin{pmatrix} E_2 \\ I_2 \end{pmatrix}$$

$$\text{Impedance line is } \begin{pmatrix} 1 \\ C\omega \end{pmatrix} \mathbb{C} = \begin{vmatrix} \bar{E}_1 & E_2 \\ \bar{I}_1 & I_2 \end{vmatrix}$$

Still not clear, ~~what~~ if I restrict to real frequencies. What's the problem Maybe you should use time evolution

$$\begin{pmatrix} 1 \\ s \end{pmatrix}^* \begin{pmatrix} 0 & 1 \\ 1 & 0 \end{pmatrix} \begin{pmatrix} 1 \\ s \end{pmatrix} = (1 \quad \bar{s}) \begin{pmatrix} s \\ 1 \end{pmatrix} = s + \bar{s}$$

$$\text{In the end you get } H(\begin{pmatrix} 1 \\ s \end{pmatrix} \otimes y) = 2\text{Re}(s) \|y\|^2$$

Consider $\begin{bmatrix} \vdots \\ \vdots \\ \vdots \end{bmatrix}^C$, behavior described by \odot

$E(t), I(t)$ satisfying $I(t) = CE(t)$. Power is $E(t)I(t) = CE\dot{E} = \frac{d}{dt}(\frac{1}{2}CE^2)$ so $\int_a^b EI dt = \frac{1}{2}CE^2 \Big|_a^b$

so the energy going in between times $a+b$. Perhaps you should think of $E(t), I(t) = 0$ for $t \ll 0$ and of exponential growth at $t \rightarrow +\infty$, which is appropriate for LT. Frequency analysis. $E(t) = \text{Re}(E(\omega)e^{-i\omega t})$

~~also~~ also for I where $E(\omega), I(\omega)$ are complex amplitudes satisfying $I(\omega) = C(-i\omega)E(\omega)$. Power

~~generally~~ generally is $\int_{-\infty}^{\infty} E(t)I(t) dt = \int_{-\infty}^{\infty} E(\omega)I(-\omega) \frac{d\omega}{2\pi}$
 $= \int_0^{\infty} \frac{d\omega}{\pi} \left(\frac{E(\omega)I(-\omega) + E(-\omega)I(\omega)}{2} \right) \text{Re}(\overline{E(\omega)}I(\omega))$. For

$I(\omega) = C(-i\omega)E(\omega)$ $\text{Re}(\overline{E(\omega)}I(\omega)) = \text{Re}(-i\omega)C|E(\omega)|^2 = 0$ for ω real corresp to $\int_{-\infty}^{\infty} EI(t) dt = 0$ if E, I have comp. support.

So the picture is the following. An edge yields a 2 diml complex space of $\begin{pmatrix} E \\ I \end{pmatrix}$ equipped with a hermitian ~~form~~ form $\text{Re}(\overline{EI}) = \frac{1}{2}(\overline{E}I + \overline{I}E)$

$= \begin{pmatrix} E \\ I \end{pmatrix}^* \begin{pmatrix} 0 & 1 \\ 1 & 0 \end{pmatrix} \begin{pmatrix} E \\ I \end{pmatrix}$. At frequency ω , ~~the~~ $\begin{pmatrix} E \\ I \end{pmatrix}$ is restricted to lie in the line ~~the~~ $\begin{pmatrix} 1 \\ C(-i\omega) \end{pmatrix} \mathbb{C}$ which is isotropic for this hermitian form. In

general $\begin{pmatrix} 1 \\ T \end{pmatrix} \mathbb{C}^n$ is isotropic for $\begin{pmatrix} E \\ I \end{pmatrix}^* \begin{pmatrix} 0 & 1 \\ 1 & 0 \end{pmatrix} \begin{pmatrix} E \\ I \end{pmatrix}$

$E, I \in \mathbb{C}^n \iff \begin{pmatrix} 1 \\ T \end{pmatrix}^* \begin{pmatrix} 0 & 1 \\ 1 & 0 \end{pmatrix} \begin{pmatrix} 1 \\ T \end{pmatrix} = (1 \ T^*) \begin{pmatrix} 1 \\ T \\ 1 \end{pmatrix} = T + T^*$
vanishes i.e. T skew symmetric. ~~This skew symm.~~

~~feature is part~~ $\begin{pmatrix} L(-i\omega) \\ 1 \end{pmatrix} \mathbb{C}^n = \begin{pmatrix} 1 \\ -1 \\ L(i\omega) \end{pmatrix}$

~~What you want~~ Picture: for a C edge you assoc. a 2d real space \mathbb{R}^2 with herm. form $\begin{pmatrix} E & \\ & I \end{pmatrix}$ and ~~the~~ subspace $\begin{pmatrix} 1 \\ c-i\omega \end{pmatrix} \mathbb{C}$ depending on frequency ω which is isot for $\omega \in \mathbb{R}$. For an L edge the same except the line is $\begin{pmatrix} L-i\omega \\ 1 \end{pmatrix} \mathbb{C}$.

$$g^* \begin{pmatrix} 0 & 1 \\ -1 & 0 \end{pmatrix} g = \begin{pmatrix} 0 & 1 \\ -1 & 0 \end{pmatrix} \quad \text{i.e.} \quad g^{-1} = \begin{pmatrix} 0 & -1 \\ +1 & 0 \end{pmatrix} g^* \begin{pmatrix} 0 & 1 \\ -1 & 0 \end{pmatrix}$$

$$= \begin{pmatrix} 0 & -1 \\ 1 & 0 \end{pmatrix} \begin{pmatrix} \bar{a} & \bar{c} \\ b & d \end{pmatrix} \begin{pmatrix} 0 & 1 \\ -1 & 0 \end{pmatrix} = \begin{pmatrix} 0 & -1 \\ 1 & 0 \end{pmatrix} \begin{pmatrix} -\bar{c} & \bar{a} \\ -d & b \end{pmatrix} = \begin{pmatrix} d & -b \\ -\bar{c} & \bar{a} \end{pmatrix}$$

$$= \frac{1}{ad-bc} \begin{pmatrix} d & -b \\ -c & a \end{pmatrix}. \quad \text{If } ad-bc=1, \text{ this means}$$

$$g \in SL_2(\mathbb{R})$$

$$\text{So the Note } \Rightarrow |\det g|^2 = 1.$$

so the group of such g contains $SL_2(\mathbb{R})$ and

$$c \begin{pmatrix} 1 & 0 \\ 0 & 1 \end{pmatrix} \quad |c|=1. \quad \text{So if you are interested}$$

in 2 dim V with herm. form of ^{sign} type $(+, -)$, then any two are isom. + auto gp is $SL_2(\mathbb{R}) \times \mathbb{T} \subset GL_2(\mathbb{C})$.

~~What~~ Our structure

What I need to do is to go directly

from the family of $\begin{pmatrix} E \\ I \end{pmatrix} \in \mathbb{C}^2$, $\begin{pmatrix} E \\ I \end{pmatrix}^* \begin{pmatrix} 0 & 1 \\ +1 & 0 \end{pmatrix} \begin{pmatrix} E \\ I \end{pmatrix}$, $\begin{pmatrix} 1 \\ c-i\omega \end{pmatrix} \mathbb{C}$

or $\begin{pmatrix} L-i\omega \\ 1 \end{pmatrix} \mathbb{C}$ to a Hilbert space Y , the Krein space $T \otimes Y$ and family $l_\omega \otimes Y$

where $T = \mathbb{C}^2$, $l_\omega = \begin{pmatrix} 1 \\ -i\omega \end{pmatrix}$, $\begin{pmatrix} x_1 \\ x_2 \end{pmatrix}^* \begin{pmatrix} 0 & 1 \\ +1 & 0 \end{pmatrix} \begin{pmatrix} y_1 \\ y_2 \end{pmatrix}$. You

need an isom $\mathbb{C}^2 \ni \begin{pmatrix} x_1 \\ x_2 \end{pmatrix} \longmapsto$

~~What you want~~ You compare $T = \mathbb{C}^2$, with $l_\omega = \begin{pmatrix} 1 \\ -i\omega \end{pmatrix}$ and $H \begin{pmatrix} x_1 \\ x_2 \end{pmatrix} = \begin{pmatrix} x_1 \\ x_2 \end{pmatrix}^* \begin{pmatrix} 0 & 1 \\ 1 & 0 \end{pmatrix} \begin{pmatrix} x_1 \\ x_2 \end{pmatrix} = \bar{x}_1 x_2 + \bar{x}_2 x_1 = 2\text{Re}(\bar{x}_1 x_2)$

106 to $P = \mathbb{C}^2$, $\mathcal{J}_\omega = \begin{pmatrix} 1 & \\ & c(-i\omega) \end{pmatrix} \mathbb{C}$, $H\left(\begin{pmatrix} x_1 \\ x_2 \end{pmatrix}\right) = 2\operatorname{Re}(x_1 x_2)$
 $\sigma = \begin{pmatrix} L(-i\omega) & \\ & 1 \end{pmatrix} \mathbb{C}$

Consider $T \begin{pmatrix} c^{1/2} & 0 \\ 0 & c^{1/2} \end{pmatrix} \rightarrow P$ $\begin{pmatrix} c^{-1/2} & 0 \\ 0 & c^{1/2} \end{pmatrix} \begin{pmatrix} 0 & 1 \\ 1 & 0 \end{pmatrix} \begin{pmatrix} c^{1/2} & 0 \\ 0 & c^{1/2} \end{pmatrix}$
 in the \mathbb{C} -case $\left. \begin{matrix} \begin{pmatrix} c^{-1/2} & 0 \\ 0 & c^{1/2} \end{pmatrix} \begin{pmatrix} 1 \\ 0 \end{pmatrix} \mathbb{C} = \begin{pmatrix} c^{-1/2} \\ c^{1/2} \end{pmatrix} \mathbb{C} = \begin{pmatrix} 1 \\ c_5 \end{pmatrix} \mathbb{C} \\ \begin{pmatrix} 0 & c^{-1/2} \\ c^{1/2} & 0 \end{pmatrix} \begin{pmatrix} c^{-1/2} & 0 \\ 0 & c^{1/2} \end{pmatrix} = \begin{pmatrix} 0 & 1 \\ 1 & 0 \end{pmatrix} \end{matrix} \right\} = \begin{pmatrix} 0 & 1 \\ 1 & 0 \end{pmatrix}$

$T \begin{pmatrix} 0 & i^{1/2} \\ i^{1/2} & 0 \end{pmatrix} \rightarrow P$ 

$\begin{pmatrix} 1 \\ -i\omega \end{pmatrix} \mathbb{C} \mapsto \begin{pmatrix} L(-i\omega) \\ 1 \end{pmatrix} \mathbb{C}$ $\begin{pmatrix} 0 & L \\ 1 & 0 \end{pmatrix} \begin{pmatrix} 1 \\ -i\omega \end{pmatrix} = \begin{pmatrix} L(-i\omega) \\ 1 \end{pmatrix}$

$\begin{pmatrix} 1 \\ -i\omega \end{pmatrix}^* \begin{pmatrix} 0 & 1 \\ 1 & 0 \end{pmatrix} \begin{pmatrix} 1 \\ -i\omega \end{pmatrix} = (1 + i\bar{\omega})(-i\omega + 1)$
 $= i(\bar{\omega} - \omega) = 2\operatorname{Im}(\omega)$

$\begin{pmatrix} -i\omega \\ 1 \end{pmatrix}^* \begin{pmatrix} 0 & 1 \\ 1 & 0 \end{pmatrix} \begin{pmatrix} -i\omega \\ 1 \end{pmatrix} = (i\bar{\omega} + 1)(1 - i\omega)$
 $= i\bar{\omega} - i\omega = 2\operatorname{Im}(\omega)$

$\begin{pmatrix} a & b \\ c & d \end{pmatrix} \begin{pmatrix} 1 \\ -i\omega \end{pmatrix} = \begin{pmatrix} -i\omega \\ 1 \end{pmatrix}$ $\begin{pmatrix} a & b \\ c & d \end{pmatrix} = \text{scalar} \begin{pmatrix} 0 & 1 \\ 1 & 0 \end{pmatrix}$

$\begin{pmatrix} 1 \\ c(-i\omega) \end{pmatrix}^* \begin{pmatrix} 0 & 1 \\ 1 & 0 \end{pmatrix} \begin{pmatrix} 1 \\ c(-i\omega) \end{pmatrix} = c(-i\omega) + \overline{c(-i\omega)}$
 $= Ci(\bar{\omega} - \omega) = (2\operatorname{Im}\omega)C$

$\begin{pmatrix} L(-i\omega) \\ 1 \end{pmatrix}^* \begin{pmatrix} 0 & 1 \\ 1 & 0 \end{pmatrix} \begin{pmatrix} L(-i\omega) \\ 1 \end{pmatrix} = Li\bar{\omega} - Li\omega$
 $= (2\operatorname{Im}\omega)L$

107 Review. An LC circuit has 2 description configuration space: ① polarized Hilbert space $\Omega = \Omega^+ \oplus \Omega^-$ plus a subquotient F_2/F_1 . ~~the~~

~~the~~ Start again. Begin again. Concrete model LC network is a graph with C, L edges. Each edge has "phase space" ~~states~~ $\begin{pmatrix} E \\ I \end{pmatrix} \in \mathbb{C}^2$, hermitian form (power) $\begin{pmatrix} E \\ I \end{pmatrix}^* \begin{pmatrix} 0 & 1 \\ 1 & 0 \end{pmatrix} \begin{pmatrix} E \\ I \end{pmatrix} = 2 \operatorname{Re}(\bar{E}I)$

For an C-edge there is a line $l_\omega = \begin{pmatrix} 1 \\ C(-i\omega) \end{pmatrix} \mathbb{C} \quad C > 0$

For an L- $l_\omega = \begin{pmatrix} L(-i\omega) \\ 1 \end{pmatrix} \mathbb{C} \quad L > 0$

~~isotropic~~ for $\omega \in S^2 = \mathbb{C} \cup \{\infty\}$ which is isotropic for ω real. (for this herm. form graphs ~~are~~ $\begin{pmatrix} 1 \\ T \end{pmatrix} \mathbb{C}$ isotropic iff $T^* = -T$).

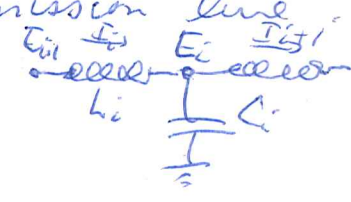
\therefore Each edge gives a 2-dim complex phase space equipped with herm. form type $(1, -1)$ and the family l_ω of lines.

$$\begin{pmatrix} 0 & iL^{1/2} \\ iL^{-1/2} & 0 \end{pmatrix}^* \begin{pmatrix} 0 & 1 \\ 1 & 0 \end{pmatrix} \begin{pmatrix} 0 & iL^{1/2} \\ iL^{-1/2} & 0 \end{pmatrix}$$

$$\begin{pmatrix} 0 & -iL^{-1/2} \\ -iL^{1/2} & 0 \end{pmatrix} \begin{pmatrix} iL^{1/2} & 0 \\ 0 & iL^{1/2} \end{pmatrix} = \begin{pmatrix} 0 & 1 \\ 1 & 0 \end{pmatrix}$$

Organize your thoughts about connecting an LC network to a transmission line.

First transmission line equation



$$\begin{cases} E_{i+1} - E_i = L_i \partial_t I_i \\ I_i - I_{i+1} = C_i \partial_t E_i \end{cases}$$

$$\begin{cases} \partial_x E + \rho \partial_t I = 0 \\ \rho \partial_x I + \partial_t E = 0 \end{cases}$$

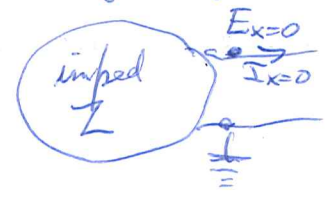
assume speed $\frac{1}{\sqrt{LC}} = 1$
 $l = \rho \quad t = \rho^{-1}$

$$\begin{cases} (\partial_x + \partial_t)(E + \rho I) = 0 \\ (\partial_x - \partial_t)(E - \rho I) = 0 \end{cases}$$

outgoing

~~incoming~~ incoming

$$\begin{aligned} E + \rho I &= A e^{-s(x+t)} \\ E - \rho I &= B e^{+s(x+t)} \end{aligned}$$



$$\frac{E_{x=0}}{I_{x=0}} = -Z$$

$$\begin{pmatrix} 1 & \rho \\ 1 & -\rho \end{pmatrix} \begin{pmatrix} E_{x=0} \\ I_{x=0} \end{pmatrix} = \begin{pmatrix} A \\ B \end{pmatrix} e^{st} \quad \begin{aligned} \frac{-Z + \rho}{-Z - \rho} &= \frac{A}{B} \end{aligned}$$

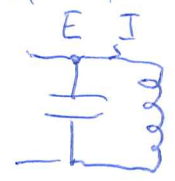
$$\frac{A}{B} = \frac{-\rho + Z}{\rho + Z}$$

typical Z is $Ls \quad \frac{1}{Cs}$

$$S = \frac{A}{B} = \frac{Ls - 1}{Ls + 1}$$

$$\text{or } \frac{\frac{1}{Cs} - 1}{\frac{1}{Cs} + 1} = \frac{1 - Cs}{1 + Cs}$$

I should take the case



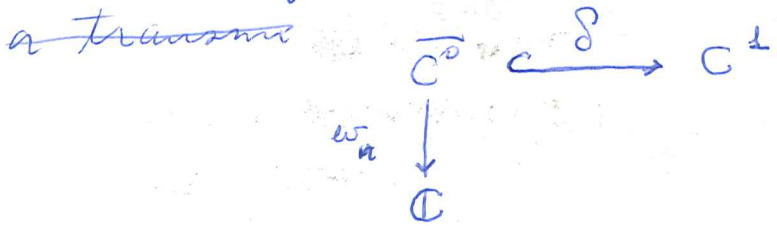
$$\frac{E}{I} = \frac{1}{\frac{1}{Ls} + Cs} = \frac{Ls}{LCs^2 + 1}$$

$$S = \begin{pmatrix} 1 & -1 \\ 1 & +1 \end{pmatrix} \begin{pmatrix} Ls \\ LCs^2 + 1 \end{pmatrix} = \frac{-LCs^2 + Ls - 1}{LCs^2 + Ls + 1}$$

$$S = \frac{-L \pm \sqrt{L^2 - 4LC}}{2LC} \quad \text{neg. real part}$$

109 ~~What you need now:~~ Take a coherent

What you need to do now is to decide how intrinsic coupling to a transmission line is. You have a picture of an LC network - subquotient of a polarized Hilbert space, namely the space of 1-cochains ~~equipped with inner product~~ split into C + L types with the inner product $C|E|^2$ resp $L^{-1}|E|^2$. ~~The~~ modified ~~to~~ form $sC|E|^2$ resp $s^{-1}L^{-1}|E|^2$ induces a hermitian (for s real) form on the subquotient, skew-herm. (for $s \in i\mathbb{R}$). So you have a line with hermitian form. For an actual circuit the line has a basis - voltage at the external node, so the hermitian form is $Z_s|E|^2$. ~~When you couple to~~



What can you do intrinsically. You have a line J and a sesquilinear form hermitian for real s . Can form $J \oplus J^T$. First do real case. You have a real line J and a quadratic form on it. Can form $J \oplus J^*$ symplectic + graph of quad form is ~~isotropic~~. ~~We take here~~ Complex case graph of a sesqui form $J \rightarrow J^T$. For a general subquotient of a polar. Hilb. space you get a sesquilinear form $Z_s(j_1, j_2)$ which is hermitian for s real (herm. means $Z_s(j_1, j) \in \mathbb{R}$) and skew-herm. for $s \in i\mathbb{R}$ (skew-herm. means $a \cdot$ herm). If $J = \mathbb{C}^n$ then $Z_s(j_1, j_2) = (j_1, Z_s j_2)$

110 ~~Output~~ Missing point A transmission line with
~~unit speed~~ has an impedance which identifies
 voltage + current spaces

$$\partial_x E + \rho \partial_t I = 0$$

YES. $(\partial_x + \partial_t)(E + \rho I) = 0$
 $(\partial_x - \partial_t)(E - \rho I) = 0$

$$\rho \partial_x I + \partial_t E = 0$$

Solutions of frequency ω are

$$E + \rho I = A e^{-s(x-t)}$$

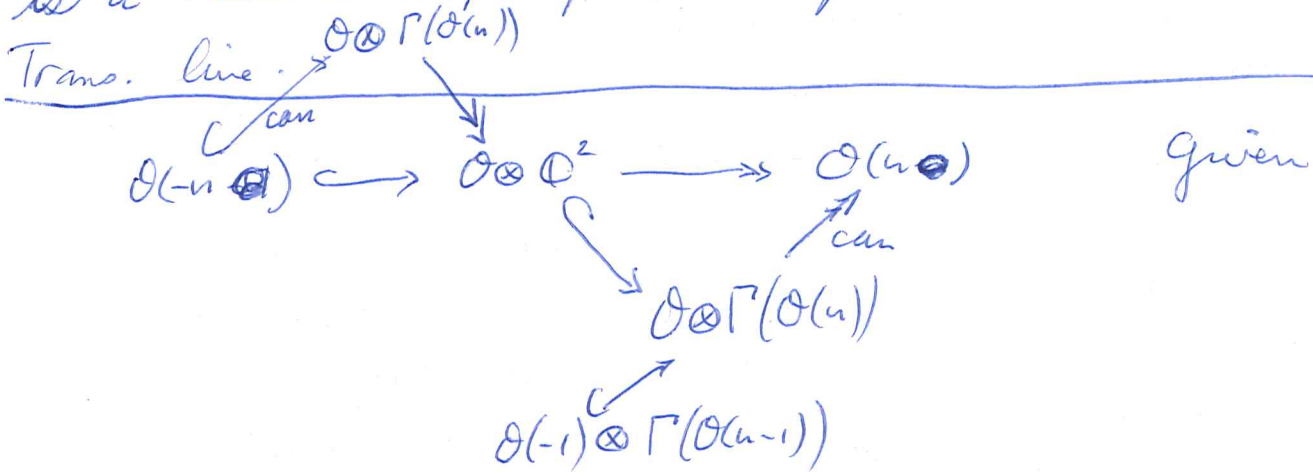
$$E - \rho I = B e^{s(x+t)}$$

so you get $\begin{pmatrix} E + \rho I \\ E - \rho I \end{pmatrix}_{x=0} = \begin{pmatrix} 1 & \rho \\ 1 & -\rho \end{pmatrix} \begin{pmatrix} E \\ I \end{pmatrix}_{x=0} = \begin{pmatrix} A \\ B \end{pmatrix} e^{st}$

so $\frac{A}{B} = \begin{pmatrix} 1 & \rho \\ 1 & -\rho \end{pmatrix} (-Z) = \frac{-Z + \rho}{-Z - \rho} = \frac{Z - \rho}{Z + \rho}$

Lesson seems to be that the ρ :

Structure of a 1-port, complex, 2 diml space
 equipped with a hermitian form of signature $(1, -1)$, also
 a line ω depending on $\Gamma(\mathcal{O}(n))$ - in finite case $\omega \mapsto \ell_\omega$
 is a rational maps from ω sphere to P^1 . ~~In Action~~



Seems strange but something might work. Go backwards,
 you have $0 \rightarrow \mathcal{O}(-1) \otimes Y \rightarrow \mathcal{O} \otimes T \otimes Y \rightarrow \mathcal{O}(1) \otimes Y \rightarrow 0$,
 and W isotropic in $T \otimes Y$

Go back to symplectic case T 2diml symplectic
 Y n diml quadratic $T \otimes Y$ symp. W isotropic
 in $T \otimes Y$. Assume $W \cap \mathcal{O}(-1) \otimes Y = 0$ $\forall \omega$, then

$$111 \quad W^\circ + \mathcal{O}(-1)_\omega \otimes Y = 0 \quad \forall \omega \quad \text{so}$$

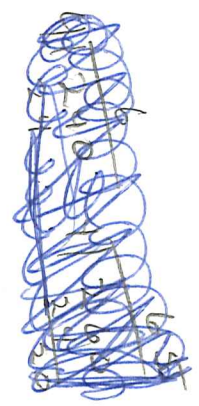
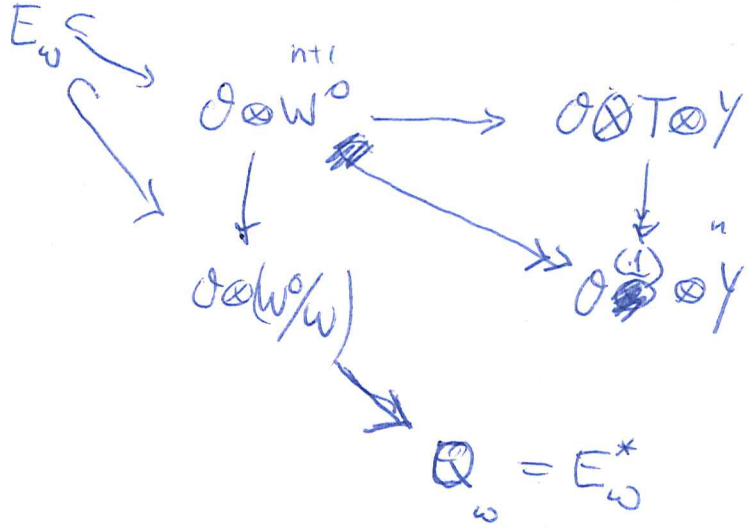
$$\text{get } W^\circ \cap \mathcal{O}(-1)_\omega \otimes Y \hookrightarrow W^\circ/W$$

nice intersection

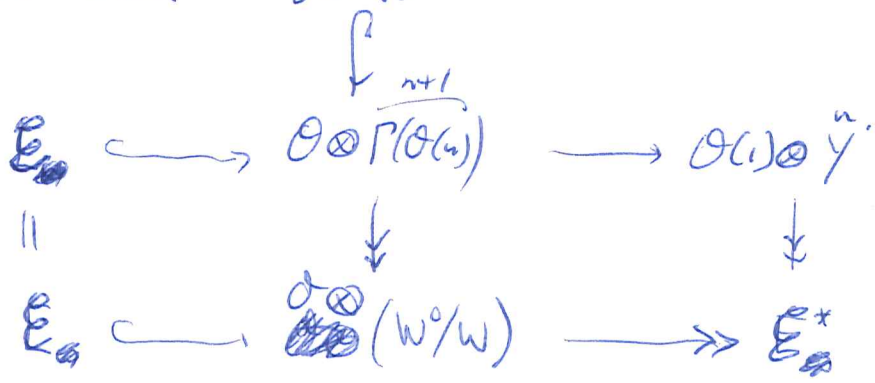
$$\mathcal{O} \otimes W^\circ \longrightarrow$$

Use $l_\omega^{\subset T}$ for $\mathcal{O}(-1)_\omega$, and

$$E_\omega = W^\circ \cap (l_\omega \otimes Y)$$



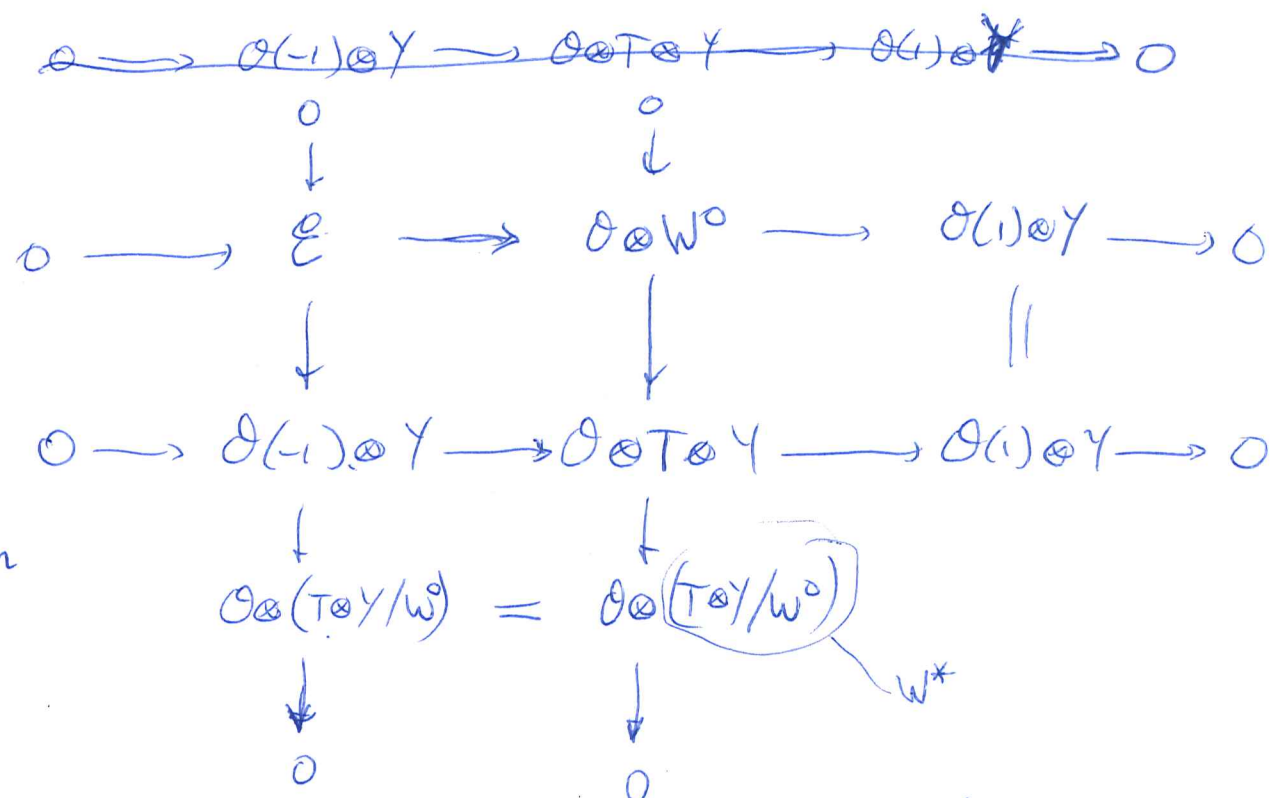
Suppose Y n dim W $n-1$ dim
 W° $n+1$ dim $E_\omega \simeq \mathcal{O}(-n)$. Can you
 reverse the process, namely start from W°/W 2dim
 symplectic and $\mathcal{O} \otimes W$



Try to reverse the symplectic version.

T 2dim symp. Y n dim ^{nondeg} quadratic $T \otimes Y$ symp.
 W isotropic in $T \otimes Y$, assume $\mathcal{O} \otimes W$ transversal to
 $\mathcal{O}(-1) \otimes Y$ over $P_1 = P_1, T$, i.e. $W \cap l_\omega \otimes Y = 0 \quad \forall \omega$
 Then $W^\circ + l_\omega \otimes Y = T \otimes Y \quad \forall \omega$ so get vector bundle

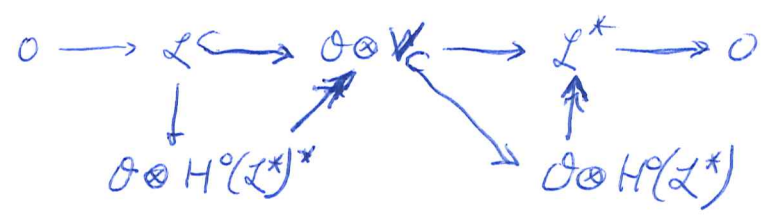
112 E $E_\omega = W^0 \cap (L_\omega \otimes Y)$. E should be Lagrangian inside $\mathcal{O} \otimes W^0/W$.



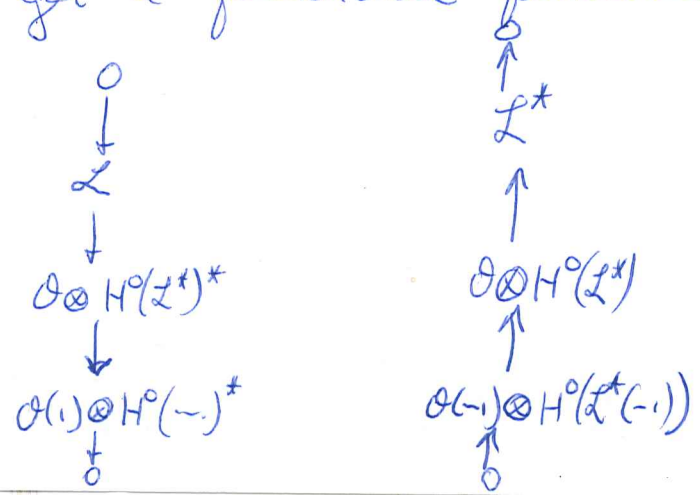
$n-1 = \dim W$
 $n = \dim Y$
 $\deg E = -n$
 $\text{rank } E = 1$

You want to reverse this process. So what do we have? How to proceed? To start with $E \hookrightarrow \mathcal{O} \otimes W^0/W \rightarrow E^*$ E Lagrangian

over \mathbb{P}_1 Problem: Classify Lagrangian subbundles L of $\mathcal{O} \otimes V$ where V is a symplectic vector space. First case $\dim V = 2$. Then $L = \mathcal{O}(n)$ for some $n \geq 0$. So we have a line bundle L^* with 2 independent sections.



Do we get a quadratic function on $H^0(L^*)^*$? deg.



113 Somehow you can fit together the canon. res. of $\mathcal{O}(n)$ and its dual

$$\begin{array}{ccccccc}
 0 & \longrightarrow & \mathcal{O}(-1) \otimes S_{n-1} & \longrightarrow & \mathcal{O} \otimes S_{n-1} & \longrightarrow & \mathcal{O}(n) \longrightarrow 0 \\
 & & & & \uparrow \mathcal{O} \otimes V & & \uparrow \mathcal{O} \otimes V \\
 0 & \longleftarrow & \mathcal{O}(1) \otimes S_{n-1}^* & \longleftarrow & \mathcal{O} \otimes S_{n-1}^* & \longleftarrow & \mathcal{O}(-n) \longleftarrow 0
 \end{array}$$

This seems fairly clear. **NO**, You probably need to use two disjoint divisors of degree n .

$$\begin{array}{ccccccc}
 0 & \longrightarrow & \mathcal{L} & \longrightarrow & \mathcal{O} \otimes W^0 & \longrightarrow & \mathcal{O}(1) \otimes Y \longrightarrow 0 \\
 & & \downarrow & & \downarrow & & \downarrow \\
 0 & \longrightarrow & \mathcal{O}(-1) \otimes Y & \longrightarrow & \mathcal{O} \otimes T \otimes Y & \longrightarrow & \mathcal{O}(1) \otimes Y \longrightarrow 0 \\
 & & \downarrow & & \downarrow & & \\
 & & \mathcal{O} \otimes W^* & = & \mathcal{O} \otimes (W^*) & &
 \end{array}$$

Maybe you should begin with $\mathcal{L} \hookrightarrow \mathcal{O} \otimes \overset{V}{W^0/W}$ and construct W^0 . But you observe that $\mathcal{L} \hookrightarrow \mathcal{O} \otimes W^0 \longrightarrow \mathcal{O}(1) \otimes Y$ must be the canonical resolution of \mathcal{L} , and then $\Gamma(\mathcal{O} \otimes W^0) \longrightarrow \Gamma(\mathcal{O}(1) \otimes Y)$ will be the

corresp K -modules. Now use ~~the~~ V symplectic and \mathcal{L} Lagrangian to get

$$\begin{array}{ccccccc}
 0 & \longrightarrow & \mathcal{L} & \longrightarrow & \mathcal{O} \otimes V & \longrightarrow & \mathcal{L}^* \longrightarrow 0 \\
 & & \downarrow & & \downarrow & & \downarrow \\
 0 & \longrightarrow & \mathcal{O}(-1) \otimes Y^* & \longrightarrow & \mathcal{O} \otimes (W^0)^* & \longrightarrow & \mathcal{L}^* \longrightarrow 0 \\
 & & \downarrow & & \downarrow & & \\
 & & \mathcal{O} \otimes (W^0)^* & = & \mathcal{O} \otimes (W^0)^* & &
 \end{array}$$

$$W^0 \subset T \otimes Y$$

Thus get canon. isom. of ~~the~~ the K -modules.

114 ~~Start~~ What happens is that $W^0 \subset T \otimes Y$
 is the K -module for \mathcal{L} and $Y^* \rightarrow T \otimes (W^0)^*$
 is the K -module for $\mathcal{L}(1)$. Other ways

$$0 \rightarrow \mathcal{O}(-1) \otimes Y^* \rightarrow \mathcal{O} \otimes (W^0)^* \rightarrow \mathcal{L}^* \rightarrow 0$$

$$(W^0)^* = H^0(\mathcal{L}^*) \quad Y^* = H^0(\mathcal{L}^*(-1))$$

$$0 \rightarrow \mathcal{L} \rightarrow \mathcal{O} \otimes W^0 \rightarrow \mathcal{O}(1) \otimes Y \rightarrow 0$$

$$Y = H^0(\mathcal{L}(-1)) \quad H^1(\mathcal{L}(-2)) = W^0$$

natural duality, but

$$0 \rightarrow \mathcal{L}(-1) \rightarrow \mathcal{O}(-1) \otimes V \rightarrow \mathcal{L}^*(-1) \rightarrow 0$$

$$\text{gives } H^0(\mathcal{L}^*(-1)) \xrightarrow{\sim} H^0(\mathcal{L}(-1)).$$

$$\begin{array}{ccc} \text{"} & & \text{"} \\ Y^* & & Y \end{array}$$

~~Basic~~ Basic data - symplectic space V and
 Lagrangian subbundle \mathcal{L} of $\mathcal{O} \otimes V$ over \mathbb{P}^1 . $H^0(\mathcal{L}) = 0$
 Wrong direction. Start with ~~nondegenerate~~ nondegenerate quadratic
 form on Y , T 2dim symplectic, ~~$T \otimes Y$~~ then
 symplectic, $W \subset T \otimes Y$ isotropic, assume $W \cap (L_W \otimes Y) = 0$
 all $\omega \in \mathbb{P}^1$, where $W^\omega + L_\omega \otimes Y = T \otimes Y \quad \forall \omega$
 where ~~\mathcal{L}_ω~~ get $\mathcal{L}_\omega = W^\omega \cap L_\omega \otimes Y \hookrightarrow W^\omega/W$

$$0 \rightarrow \mathcal{L} \rightarrow \mathcal{O} \otimes W^0 \rightarrow \mathcal{O}(1) \otimes Y \rightarrow 0$$

must be

$$\text{canonical resolution so } Y \xrightarrow{\sim} H^0(\mathcal{L}(-1))$$

$$H^1(\mathcal{L}(-2)) \xrightarrow{\sim} W^0$$

$$0 \rightarrow \mathcal{O}(-1) \otimes Y^* \rightarrow \mathcal{O} \otimes (W^0)^* \rightarrow \mathcal{L}^* \rightarrow 0$$

$$(W^0)^* \xrightarrow{\sim} H^0(\mathcal{L}^*) \quad H^0(\mathcal{L}^*(-1)) \xrightarrow{\sim} Y^*$$

$$0 \rightarrow \mathcal{L} \rightarrow \mathcal{O} \otimes V \rightarrow \mathcal{L}^* \rightarrow 0$$

$$\underbrace{H^0(\mathcal{L}^*(-1))}_{Y^*} \xrightarrow{\sim} \underbrace{H^0(\mathcal{L}(-1))}_Y$$

115 So ~~that~~ now that this is clear you ~~let~~ want to work in the real line. ~~Let~~ ~~is considered~~ details of the alg. situation. Use coord z .

$$T = \mathbb{C}^2 \quad l_z = \begin{pmatrix} 1 \\ z \end{pmatrix} \mathbb{C}. \quad T \otimes Y = \begin{matrix} Y \\ \oplus \\ Y \end{matrix} \quad \text{You}$$

need an isom $Y^* \cong Y$, non degenerate sym pairing

Naturally $\begin{matrix} Y \\ \oplus \\ Y^* \end{matrix}$ is symplectic and those subspaces

which are graphs $\begin{pmatrix} 1 \\ F \end{pmatrix} Y$ have form $\begin{pmatrix} 1 \\ F \end{pmatrix} Y \quad F = F^*$

$F: Y \rightarrow Y^*$ symmetric. But to make sense of $\begin{pmatrix} 1 \\ z \end{pmatrix}$ you need

a fixed g.f. So $T \otimes Y \quad l_z$ have a standard

form. ~~As is not the case~~

Real case $T = \mathbb{R}^2$ skew form $\begin{pmatrix} x_1 \\ x_2 \end{pmatrix}^t \begin{pmatrix} 0 & 1 \\ -1 & 0 \end{pmatrix} \begin{pmatrix} x'_1 \\ x'_2 \end{pmatrix}$

Y ~~is~~ Euclidean space

$$= x_1 x'_2 - x_2 x'_1 = \begin{vmatrix} x_1 & x'_1 \\ x_2 & x'_2 \end{vmatrix}$$

$$T \otimes Y = \begin{matrix} Y \\ \oplus \\ Y \end{matrix} \quad \text{with} \quad \begin{pmatrix} y_1 \\ y_2 \end{pmatrix}^t \begin{pmatrix} 0 & 1 \\ -1 & 0 \end{pmatrix} \begin{pmatrix} y'_1 \\ y'_2 \end{pmatrix} = y_1^t y'_2 - y_2^t y'_1$$

$\Gamma_\alpha = \begin{pmatrix} 1 \\ \alpha \end{pmatrix} Y$ is isotropic means $\begin{pmatrix} 1 \\ \alpha \end{pmatrix}^t \begin{pmatrix} 0 & 1 \\ -1 & 0 \end{pmatrix} \begin{pmatrix} 1 \\ \alpha \end{pmatrix} = \alpha - \alpha^t = 0$

$$\text{In fact } \Gamma_\alpha^0 = \left\{ \begin{pmatrix} y_1 \\ y_2 \end{pmatrix} \mid \begin{pmatrix} y_1 \\ y_2 \end{pmatrix}^t \begin{pmatrix} 0 & 1 \\ -1 & 0 \end{pmatrix} \begin{pmatrix} y_1 \\ y_2 \end{pmatrix} = 0 \right\} = \Gamma_{\alpha^t}$$

$$y_1^t y_2 - y_2^t \alpha^t y_1 = 0 \quad \therefore y_2 = \alpha^t y_1$$

So now consider ~~W~~ W isotropic in $\begin{matrix} Y \\ \oplus \\ Y \end{matrix}$

$W = \begin{pmatrix} \varepsilon \\ A \end{pmatrix} X$ If you want W is subspace of $\begin{matrix} Y \\ \oplus \\ Y \end{matrix}$

$$\varepsilon = p_1|_W, \quad A = p_2|_W \quad W^0 = \left\{ \begin{pmatrix} y_1 \\ y_2 \end{pmatrix} \mid \begin{pmatrix} \varepsilon x \\ Ax \end{pmatrix}^t \begin{pmatrix} 0 & 1 \\ -1 & 0 \end{pmatrix} \begin{pmatrix} y_1 \\ y_2 \end{pmatrix} = 0 \right\}$$

$$(\varepsilon x)^t y_2 = (Ax)^t y_1 \quad \text{or} \quad \varepsilon^t y_2 = A^t y_1 \quad \varepsilon^t A = A^t \varepsilon$$

t denotes $*$ ~~with~~ wrt some scalar prod on X

$$W \cap \begin{pmatrix} 1 \\ \alpha \end{pmatrix} Y = \left\{ \begin{pmatrix} \varepsilon \\ A \end{pmatrix} x \mid \lambda \varepsilon x = Ax \right\} \quad \text{I mean}$$

116 You need to complexify - choose orthonormal basis for Y and X , then have solution of $(\lambda \varepsilon - A)x = 0 \quad x \in X_c$.

$$\lambda x = Ax$$

$$(x, \lambda x) = (x, Ax) \quad \bar{\lambda} \|x\|^2$$

$$\lambda \|x\|^2 \quad (Ax, x) = \overline{(x, Ax)}$$

$$0 = (\varepsilon x, (\lambda \varepsilon - A)x) = \lambda \|\varepsilon x\|^2 - (\varepsilon x, Ax)$$

$$0 = (A\varepsilon - A)x, \varepsilon x) = \bar{\lambda} \|\varepsilon x\|^2 - (Ax, \varepsilon x) \quad \therefore \lambda = \bar{\lambda}$$

Point $(\varepsilon x, Ax) \in \mathbb{R}$

So make assumption that no bound states, this is something you test ~~over \mathbb{R}~~ in the real setting. In particular you want ε inj ($\lambda = \infty$) A inj ($\lambda = 0$).

More review $T = \mathbb{R}^2$ skew form $\begin{pmatrix} x_1 \\ x_2 \end{pmatrix} \begin{pmatrix} 0 & 1 \\ -1 & 0 \end{pmatrix} \begin{pmatrix} x'_1 \\ x'_2 \end{pmatrix} = \begin{vmatrix} x_1 & x'_1 \\ x_2 & x'_2 \end{vmatrix}$
 Y Euclidean space, $T \otimes Y = \begin{matrix} Y \\ \oplus \\ Y \end{matrix}$, ~~with~~ skew form

$$\begin{pmatrix} y_1 \\ y_2 \end{pmatrix}^* \begin{pmatrix} 0 & 1 \\ -1 & 0 \end{pmatrix} \begin{pmatrix} y'_1 \\ y'_2 \end{pmatrix} = y_1^* y'_2 - y_2^* y'_1$$

$$\Gamma_\alpha = \begin{pmatrix} 1 \\ \alpha \end{pmatrix} Y \quad \Gamma_\alpha^0 = \begin{pmatrix} 1 \\ \alpha^t \end{pmatrix} Y \quad (\varepsilon x)^* y_2 = (Ax)^* y_1$$

$$W \text{ isot. in } \begin{matrix} Y \\ \oplus \\ Y \end{matrix} \quad W = \begin{pmatrix} \varepsilon \\ A \end{pmatrix} X \quad W^0 = \left\{ \begin{pmatrix} y_1 \\ y_2 \end{pmatrix} \mid \varepsilon^* y_2 = A^* y_1 \right\}$$

$$W_c \cap \begin{pmatrix} 1 \\ \lambda \end{pmatrix} Y_c \cong \left\{ \begin{pmatrix} \varepsilon x \\ Ax \end{pmatrix} \mid (\lambda \varepsilon - A)x = 0 \right\} \quad \boxed{\varepsilon^* A = A^* \varepsilon \text{ if } W \subset W^0}$$

$$0 = (\varepsilon x, (\lambda \varepsilon - A)x) = \lambda \|\varepsilon x\|^2 - (\varepsilon x, Ax) \in \mathbb{R} \quad \therefore \lambda \in \mathbb{R}$$

~~$0 = (\varepsilon x, (\lambda \varepsilon - A)x) = \lambda \|\varepsilon x\|^2 - (\varepsilon x, Ax)$~~

Continue + calculate W^0 , ε inj so can arrange $\varepsilon^* \varepsilon = I$. $Y = \varepsilon X \oplus \text{Ker}(\varepsilon^*)$. $\text{Ker}(\varepsilon^*) = \{y \mid \varepsilon^* y = 0\}$

$$\begin{matrix} Y \\ \oplus \\ Y \end{matrix} \cong \begin{pmatrix} \varepsilon \\ A \end{pmatrix} X \oplus \begin{pmatrix} 0 \\ \oplus \\ \text{Ker} \varepsilon^* \end{pmatrix}$$

Assume $\varepsilon^* y_2 = A^* y_1$

$$\begin{pmatrix} \varepsilon x \\ y_1 \end{pmatrix} = \begin{pmatrix} A^* y_2 \\ y_2 \end{pmatrix}$$

$$\begin{pmatrix} \varepsilon x \\ y_2 \end{pmatrix} = \begin{pmatrix} A x \\ y_1 \end{pmatrix}$$

117 Given $y_1 \quad x \mapsto (Ax, y_1)$ can be represented as $(\varepsilon x, y_2)$ uniquely with $y_2 \in \varepsilon X$.

Define a map $Y \xrightarrow{\theta} X$ by requiring $(Ax, y) = (\varepsilon x, \varepsilon \theta y) = (x, \theta y) \quad \therefore \theta = A^*$

W^0 seems to consist of $\begin{pmatrix} \varepsilon \\ A \end{pmatrix}$

Start with $\begin{pmatrix} y_1 \\ y_2 \end{pmatrix} \in W^0$ i.e. $\varepsilon^* y_2 = A^* y_1$

or $(\varepsilon x, y_2) = (Ax, y_1) \quad \forall x$ But

$$(Ax, y_1) = (x, A^* y_1) = (\varepsilon x, \varepsilon A^* y_1)$$

~~$(\varepsilon x, \varepsilon \varepsilon^* y_2 + (1 - \varepsilon \varepsilon^*) y_2) = (x, \varepsilon y_2)$~~

Claim y_1 so $y_2 \equiv \varepsilon A^* y_1 \pmod{\text{Ker } \varepsilon^*}$

$$W^0 = \begin{pmatrix} 1 \\ \varepsilon A^* \end{pmatrix} Y + \begin{pmatrix} 0 \\ \oplus \\ \text{Ker } \varepsilon^* \end{pmatrix}$$

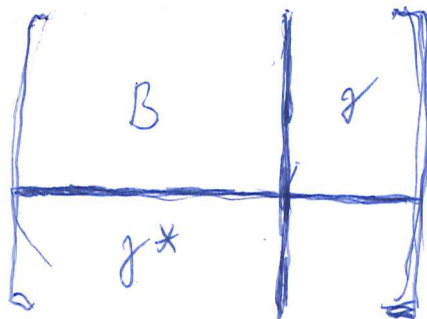
Proof: Given $\begin{pmatrix} y_1 \\ y_2 \end{pmatrix} \in W^0$ i.e. $(\varepsilon x, y_2) = (Ax, y_1) \quad \forall x$

Then $(\varepsilon x, y_2) = (x, A^* y_1) = (\varepsilon x, \varepsilon A^* y_1) \Rightarrow y_2 - \varepsilon A^* y_1 \in \text{Ker } \varepsilon^*$

$$\begin{pmatrix} y_1 \\ y_2 \end{pmatrix} = \begin{pmatrix} y_1 \\ \varepsilon A^* y_1 \end{pmatrix} \oplus \begin{pmatrix} 0 \\ y_2 - \varepsilon A^* y_1 \end{pmatrix}$$

εx
 $\varepsilon A^* \varepsilon x$
 $= \varepsilon \varepsilon^* A x = A x$

Note $(\varepsilon A^*)^* = A \varepsilon^*$



$$\pi = 1 - \varepsilon \varepsilon^*$$

$$A \varepsilon^* = \underbrace{\varepsilon \varepsilon^* A \varepsilon^*}_{\varepsilon A^* \varepsilon \varepsilon^*} + \pi A \varepsilon^*$$

$$\varepsilon A^* = \varepsilon A^* \varepsilon \varepsilon^* + \varepsilon A^* \pi$$

So it should be possible to ~~uniquely~~ uniquely extend the partial ~~symm.~~ symm. op. $\begin{pmatrix} \varepsilon \\ A \end{pmatrix}$ to a symm. operator \tilde{A} on $Y \ni \pi(\tilde{A})\pi = 0$. This gives a kind of canonical extension. ~~uniquely~~!!!

~~On the end you seem to get~~

W^0/W is symplectic and you have ~~constructed~~ found a canonical Lagrangian subspace. In fact we have $W^0/W \cong \begin{matrix} \text{Ker } \varepsilon^* \\ \oplus \\ \text{Ker } \varepsilon \end{matrix}$

What is the answer? You have $W = \begin{pmatrix} \varepsilon \\ A \end{pmatrix} X \subseteq \begin{matrix} Y \\ \oplus \\ Y \end{matrix}$ and $W \subseteq \begin{pmatrix} 1 \\ \tilde{A} \end{pmatrix} Y \subseteq W^0$
 $\varepsilon \varepsilon^* A \varepsilon^*$

$$\tilde{A} = A \varepsilon^* + \varepsilon A^* - \underbrace{\varepsilon A^* \varepsilon \varepsilon^*}_{\varepsilon A^* \varepsilon \varepsilon^*}$$

$$= A \varepsilon^* + \varepsilon A^* \pi = \pi A \varepsilon^* + \varepsilon A^*$$

Now we have ~~a simple problem~~ to find $W^0 \cap \begin{pmatrix} 1 \\ \lambda \end{pmatrix} Y$,

$$W^0 = \left\{ \begin{pmatrix} y_1 \\ y_2 \end{pmatrix} \mid A^* y_1 = \varepsilon^* y_2 \right\} \quad W^0 \cap \begin{pmatrix} 1 \\ \lambda \end{pmatrix} Y \cong \begin{pmatrix} 1 \\ \lambda \end{pmatrix} \text{Ker}(\lambda \varepsilon^* - A^*)$$

You guess that the response function should have a simple form, like what you found for an LC network.

~~The basic problem here~~ The idea here is that the response is ~~subbundle~~ Lagrangian subbundle $L \subset \mathcal{O} \otimes (W^0/W)$ and since $W^0/W = \begin{pmatrix} \text{Ker } \varepsilon^* \\ \text{Ker } \varepsilon \end{pmatrix}$, L_ω should be the graph of a ~~symmetric~~ symmetric operator on $\text{Ker}(\varepsilon^*)$.

Idea resolvent of $\left(\begin{array}{c|c} \varepsilon^* A = A^* \varepsilon & \varepsilon A^* \pi \\ \hline \pi A \varepsilon^* & 0 \end{array} \right)$ - Something like this

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$$\begin{pmatrix} a & b \\ c & d \end{pmatrix} = \begin{pmatrix} \lambda - \beta & -\gamma \\ -\gamma^* & \lambda \end{pmatrix}$$

$$\begin{pmatrix} a & b \\ c & d \end{pmatrix}^{-1} = \frac{1}{ad-bc} \begin{pmatrix} d & -b \\ -c & a \end{pmatrix}$$

$$(d - \epsilon a^* b)^{-1} = \left(\lambda - \gamma^* \frac{1}{\lambda - \beta} \gamma \right)^{-1}$$

You are trying for ~~something~~ something like ~~inducing~~ inducing a quadratic form on a subspace or quotient space. ~~So how~~

So you run into a ~~familiar~~ familiar situation namely you take the resolvent and ~~project~~ project into ~~something~~ compress into a generating subspaces. The answer is very easy. The problem is to fit it into something ~~like~~.

Now what. $\epsilon^* \epsilon = I$. ~~EA~~ Can write

$$Y = \epsilon X \oplus \text{Ker } \epsilon^*$$

$$\gamma =$$

You need to organize all this stuff. ~~What~~ How. Go back To construct $L_\omega = W^\circ \cap (\omega) Y$

$$\tilde{A} = \left[\begin{array}{c|c} \epsilon \epsilon^* A \epsilon^* & \epsilon A^* \pi \\ \hline \pi A \epsilon^* & 0 \end{array} \right]$$

$$\xrightarrow{\delta \log} \det \begin{pmatrix} \lambda - \alpha & -\gamma \\ -\gamma^* & \lambda - \beta \end{pmatrix} = \text{tr} \begin{pmatrix} \lambda - \alpha & -\gamma \\ -\gamma^* & \lambda - \beta \end{pmatrix}^{-1} \begin{pmatrix} 0 & \sigma \\ 0 & \delta \beta \end{pmatrix}$$

$$= \text{tr} \left(\frac{1}{\lambda - \beta - \gamma^* \frac{1}{\lambda - \alpha} \gamma} \right) \delta \beta$$

Note that

(is hermitian for λ real.

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~~What's going on?~~

Given $\varepsilon^* y_1 = 0$, then $A y_1$

You want to calculate W^0 where $W = \begin{pmatrix} \varepsilon x \\ A x \end{pmatrix}$
 Let $\begin{pmatrix} y_1 \\ y_2 \end{pmatrix} \in W^0$, i.e. $A^* y_1 = \varepsilon^* y_2$ removes

$\begin{pmatrix} \varepsilon \varepsilon^* y_1 \\ A \varepsilon^* y_1 \end{pmatrix}$ from $\begin{pmatrix} y_1 \\ y_2 \end{pmatrix}$ to assume $\varepsilon^* y_1 = 0$. Now

we have $\forall x \quad \begin{pmatrix} y_1 \\ y_2 \end{pmatrix}, Ax = (x_1, x)$ for ~~some~~ x_1

i.e. $A \begin{pmatrix} y_1 \\ y_2 \end{pmatrix} = \varepsilon x \quad (A^* y_1, x) = (x_1, x)$.

so it seems that if $\varepsilon^* y_1 = 0$ then $\begin{pmatrix} y_1 \\ \varepsilon A^* y_1 \end{pmatrix}$

Wait given $\begin{pmatrix} y_1 \\ y_2 \end{pmatrix} \in W^0$ i.e. $A^* y_1 = \varepsilon^* y_2$ then

$\begin{pmatrix} y_1 \\ \varepsilon A^* y_1 \end{pmatrix}$ satisfies $A^* y_1 = \varepsilon^* (\varepsilon A^* y_1)$, so $\Gamma_{\varepsilon A^*} \subset W^0$
 also $(\Gamma_{\varepsilon A^*})^0 = \Gamma_{A \varepsilon^*}$?

The point is that ~~the set~~ $\begin{pmatrix} 1 \\ \varepsilon A^* \end{pmatrix} Y \subset W^0$

because $\varepsilon A^* (\varepsilon x) = \varepsilon \varepsilon^* A x$ NO

$$\begin{pmatrix} y_1 \\ y_2 \end{pmatrix} = \begin{pmatrix} y_1 \\ \varepsilon A^* y_1 \end{pmatrix} + \begin{pmatrix} 0 \\ y_2 - \varepsilon A^* y_1 \end{pmatrix}$$

$$\begin{pmatrix} \varepsilon \varepsilon^* y_1 \\ \varepsilon A^* \varepsilon \varepsilon^* y_1 \end{pmatrix} + \begin{pmatrix} y_1 - \varepsilon \varepsilon^* y_1 \\ y_2 - \varepsilon A^* y_1 \end{pmatrix}$$

Given $y_1 \exists! x_1$
 $\exists \begin{pmatrix} y_1 \\ Ax \end{pmatrix} = (x_1, \varepsilon x) \quad \forall x$
 i.e. $x_1 = A^* y_1$

$$\therefore \begin{pmatrix} y_1 \\ \varepsilon A^* y_1 \end{pmatrix} \in W^0$$

$A \varepsilon^*$

$\begin{pmatrix} y_1 \\ A \varepsilon^* y_1 \end{pmatrix} \notin W^0$

$A^* y_1 \stackrel{?}{=} \varepsilon^* A \varepsilon^* y_1$
 \parallel
 $A^* \varepsilon \varepsilon^* y_1$
 NO